

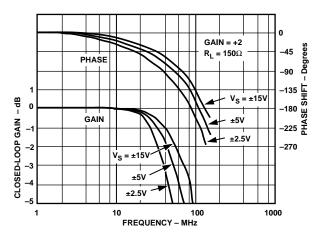
Low Power Video Op Amp with Disable

AD810

FEATURES

High Speed 80 MHz Bandwidth (3 dB, G = +1) 75 MHz Bandwidth (3 dB, G = +2) 1000 V/µs Slew Rate 50 ns Settling Time to 0.1% (Vo = 10 V Step) **Ideal for Video Applications** 30 MHz Bandwidth (0.1 dB, G = +2) 0.02% Differential Gain 0.04° Differential Phase Low Noise 2.9 nV/ $\sqrt{\text{Hz}}$ Input Voltage Noise 13 pA/ $\sqrt{\text{Hz}}$ Inverting Input Current Noise Low Power 8.0 mA Supply Current max 2.1 mA Supply Current (Power-Down Mode) **High Performance Disable Function** Turn-Off Time 100 ns **Break Before Make Guaranteed** Input to Output Isolation of 64 dB (OFF State) **Flexible Operation** Specified for ±5 V and ±15 V Operation ± 2.9 V Output Swing Into a 150 Ω Load (V_s = ± 5 V)

APPLICATIONS Professional Video Cameras Multimedia Systems NTSC, PAL & SECAM Compatible Systems Video Line Driver ADC/DAC Buffer DC Restoration Circuits

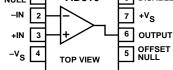


Closed-Loop Gain and Phase vs. Frequency, G = +2, $R_L = 150$, $R_F = 715 \Omega$

REV. A

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OFFSET 1 • AD810 8 DISABLE



CONNECTION DIAGRAM

8-Pin Plastic Mini-DIP (N), SOIC (R)

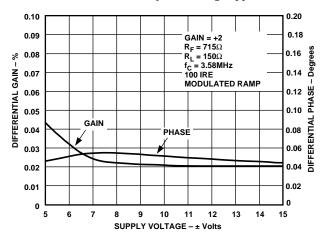
and Cerdip (Q) Packages

PRODUCT DESCRIPTION

The AD810 is a composite and HDTV compatible, current feedback, video operational amplifier, ideal for use in systems such as multimedia, digital tape recorders and video cameras. The 0.1 dB flatness specification at bandwidth of 30 MHz (G = +2) and the differential gain and phase of 0.02% and 0.04° (NTSC) make the AD810 ideal for any broadcast quality video system. All these specifications are under load conditions of 150 Ω (one 75 Ω back terminated cable).

The AD810 is ideal for power sensitive applications such as video cameras, offering a low power supply current of 8.0 mA max. The disable feature reduces the power supply current to only 2.1 mA, while the amplifier is not in use, to conserve power. Furthermore the AD810 is specified over a power supply range of ± 5 V to ± 15 V.

The AD810 works well as an ADC or DAC buffer in video systems due to its unity gain bandwidth of 80 MHz. Because the AD810 is a transimpedance amplifier, this bandwidth can be maintained over a wide range of gains while featuring a low noise of 2.9 nV/ $\sqrt{\text{Hz}}$ for wide dynamic range applications.



Differential Gain and Phase vs. Supply Voltage

AD810–SPECIFICATIONS (@ $T_A = +25$ °C and $V_S = \pm 15$ V dc, $R_L = 150 \Omega$ unless otherwise noted)

Parameter DYNAMIC PERFORMANCE 3 dB Bandwidth 0.1 dB Bandwidth Full Power Bandwidth Slew Rate ²	$\begin{tabular}{ c c c c } \hline Conditions \\ \hline (G = +2) & R_{FB} = 715 \\ \hline (G = +2) & R_{FB} = 715 \\ \hline (G = +1) & R_{FB} = 1000 \\ \hline (G = +10) & R_{FB} = 270 \\ \hline (G = +2) & R_{FB} = 715 \\ \hline (G = +2) & R_{FB} = 715 \\ \hline V_O = 20 & V & P-P, \\ \hline \end{tabular}$	$\begin{array}{c} \pm 5 \ V \\ \pm 15 \ V \\ \pm 5 \ V \end{array}$	Min 40 55	Тур 50 75	Max	Min 40	Тур	Max	Units
3 dB Bandwidth 0.1 dB Bandwidth Full Power Bandwidth	$ \begin{array}{l} (G = +2) \ R_{FB} = 715 \\ (G = +1) \ R_{FB} = 1000 \\ (G = +10) \ R_{FB} = 270 \\ (G = +2) \ R_{FB} = 715 \\ (G = +2) \ R_{FB} = 715 \end{array} $	±15 V ±15 V ±15 V	55			40	50		1
0.1 dB Bandwidth Full Power Bandwidth	$ \begin{array}{l} (G = +2) \ R_{FB} = 715 \\ (G = +1) \ R_{FB} = 1000 \\ (G = +10) \ R_{FB} = 270 \\ (G = +2) \ R_{FB} = 715 \\ (G = +2) \ R_{FB} = 715 \end{array} $	±15 V ±15 V ±15 V	55				50		MHz
Full Power Bandwidth	$ \begin{array}{l} (G = +1) \ R_{FB} = 1000 \\ (G = +10) \ R_{FB} = 270 \\ (G = +2) \ R_{FB} = 715 \\ (G = +2) \ R_{FB} = 715 \end{array} $	±15 V ±15 V		/0		55	75		MHz
Full Power Bandwidth	$\begin{array}{l} (G = +10) \ R_{FB} = 270 \\ (G = +2) \ R_{FB} = 715 \\ (G = +2) \ R_{FB} = 715 \end{array}$	±15 V	40	80		40	80		MHz
Full Power Bandwidth	$ \begin{array}{l} (G = +2) \ R_{FB} = 715 \\ (G = +2) \ R_{FB} = 715 \end{array} $		50	65		50	65		MHz
Full Power Bandwidth	$(G = +2) R_{FB} = 715$		13	22		13	22		MHz
		±15 V	15	30		15	30		MHz
Slew Rate ²	$V \cap - \lambda U V P^{-} P_{1}$								
Slew Rate ²	$R_{\rm L} = 400 \Omega$	±15 V		16			16		MHz
	$R_{\rm L} = 150 \Omega$	±5 V		350			350		V/µs
	$R_{\rm L} = 400 \ \Omega$	±15 V		1000			1000		V/µs
Settling Time to 0.1%	10° V Step, G = -1	±15 V		50			50		ns
Settling Time to 0.01%	10 V Step, G = -1	±15 V		125			125		ns
Differential Gain	f = 3.58 MHz	±15 V		0.02	0.05		0.02	0.05	%
	f - 3.58 MHz	±5 V		0.04	0.07		0.04	0.07	%
Differential Phase	f = 3.58 MHz	±15 V		0.04	0.07		0.04	0.07	Degrees
	f = 3.58 MHz	±5 V		0.045	0.08		0.045	0.08	Degrees
Total Harmonic Distortion	$f = 10 \text{ MHz}, V_0 = 2 \text{ V } p-p$								
	$R_{L} = 400 \Omega, G = +2$	±15 V		-61			-61		dBc
INPUT OFFSET VOLTAGE		±5 V, ±15 V	1	1.5	6		1.5	6	mV
INPUT OFFSET VOLTAGE	T _{MIN} -T _{MAX}				0 7.5		1.5 4	o 15	mV
Offset Voltage Drift	I MIN ^{-I} MAX	±5 V, ±15 V		2 7	7.5		4 15	15	μV/°C
Offset Voltage Drift				1			15		μν/ C
INPUT BIAS CURRENT									
–Input	T _{MIN} -T _{MAX}	±5 V, ±15 V		0.7	5		0.8	5	μA
+Input	T _{MIN} -T _{MAX}	±5 V, ±15 V		2	7.5		2	10	μA
OPEN-LOOP	T _{MIN} -T _{MAX}								
TRANSRESISTANCE	$V_{\rm O} = \pm 10 \text{ V}, \text{ R}_{\rm L} = 400 \Omega$	±15 V	1.0	3.5		1.0	3.5		MΩ
	$V_0 = \pm 2.5 \text{ V}, \text{ R}_L = 100 \Omega$	±5 V	0.3	1.2		0.2	1.0		MΩ
OPEN-LOOP	$T_{MIN}-T_{MAX}$	1 4 1 7 17		100		00	100		10
DC VOLTAGE GAIN	$V_0 = \pm 10 \text{ V}, \text{ R}_L = 400 \Omega$	$\pm 15 \text{ V}$	86	100		80	100		dB
	$V_0 = \pm 2.5 \text{ V}, \text{ R}_L = 100 \Omega$	±5 V	76	88		72	88		dB
COMMON-MODE REJECTION	T _{MIN} -T _{MAX}								
V _{OS}	$V_{CM} = \pm 12 V$	±15 V	56	64		56	64		dB
	$V_{CM} = \pm 2.5 V$	±5 V	52	60		50	60		dB
±Input Current	T _{MIN} -T _{MAX}	±5 V, ±15 V		0.1	0.4		0.1	0.4	μA/V
POWER SUPPLY REJECTION		±4.5 V to ±18 V							
V _{os}	T _{MIN} -T _{MAX}	14.5 V 10 110 V	65	72		60	72		dB
±Input Current	$T_{MIN} - T_{MAX}$ $T_{MIN} - T_{MAX}$		05	0.05	0.3	00	0.05	0.3	μA/V
			_		0.5			0.5	
INPUT VOLTAGE NOISE	f = 1 kHz	±5 V, ±15 V		2.9			2.9		nV/√Hz
INPUT CURRENT NOISE	$-I_{IN}$, f = 1 kHz	±5 V, ±15 V		13			13		pA/√Hz
	$+I_{IN}, f = 1 \text{ kHz}$	±5 V, ±15 V		1.5			1.5		pA/\sqrt{Hz}
			105			105			•
INPUT COMMON-MODE		±5 V	±2.5	±3.0		±2.5	±3		V
VOLTAGE RANGE		±15 V	±12	±13		±12	±13		V
OUTPUT CHARACTERISTICS									
Output Voltage Swing ³	$R_L = 150 \Omega, T_{MIN} - T_{MAX}$	±5 V	± 2.5	± 2.9		± 2.5	± 2.9		V
	$R_L = 400 \Omega$	±15 V	±12.5	± 12.9		±12.5	± 12.9		V
	$R_{L}^{-} = 400 \Omega, T_{MIN}^{-}T_{MAX}$	±15 V	±12			±12			V
Short-Circuit Current		±15 V	1	150			150		mA
Output Current	T _{MIN} -T _{MAX}	±5 V, ±15 V	40	60		30	60		mA
OUTPUT RESISTANCE	Open Loop (5 MHz)			15			15		Ω
				10			10		
INPUT CHARACTERISTICS									
Input Resistance	+Input	±15 V	2.5	10		2.5	10		MΩ
	-Input	±15 V		40			40		Ω
Input Capacitance	+Input	±15 V		2			2		pF
DISABLE CHARACTERISTICS ⁴									
OFF Isolation	f = 5 MHz, See Figure 43		1	64			64		dB
OFF Output Impedance	See Figure 43		(R	$+ R_{G}) \ 13$	рF	(R+	$R_{\rm G} = R_{\rm G}$	рF	

			AD810A AD810S ¹		1				
Parameter	Conditions	Vs	Min	Тур	Max	Min	Тур	Max	Units
Turn On Time ⁵	Z_{OUT} = Low, See Figure 54			170			170		ns
Turn Off Time	$Z_{OUT} = High$			100			100		ns
Disable Pin Current	Disable $Pin = 0 V$	±5 V		50	75		50	75	μA
		±15 V		290	400		290	400	μA
Min Disable Pin Current to									
Disable	$T_{MIN}-T_{MAX}$	±5 V, ±15 V		30			30		μA
POWER SUPPLY									
Operating Range	+25°C to T _{MAX}		±2.5		± 18	± 2.5		± 18	V
	T _{MIN}		±3.0		± 18	± 3.5		± 18	V
Quiescent Current		±5 V		6.7	7.5		6.7	7.5	mA
-		±15 V		6.8	8.0		6.8	8.0	mA
	T _{MIN} -T _{MAX}	±5 V, ±15 V		8.3	10.0		9	11.0	mA
Power-Down Current		±5 V		1.8	2.3		1.8	2.3	mA
		±15 V		2.1	2.8		2.1	2.8	mA

NOTES

¹See Analog Devices Military Data Sheet for 883B Specifications.

²Slew rate measurement is based on 10% to 90% rise time with the amplifier configured for a gain of -10.

³Voltage Swing is defined as useful operating range, not the saturation range.

⁴Disable guaranteed break before make.

⁵Turn On Time is defined with ±5 V supplies using complementary output CMOS to drive the disable pin.

Specifications subject to change without notice.

ABSOLUTE MAXIMUM RATINGS¹

Supply Voltage±18 V
Internal Power Dissipation ² Observe Derating Curves
Output Short Circuit Duration Observe Derating Curves
Common-Mode Input Voltage $\ldots \ldots \pm V_S$
Differential Input Voltage±6 V
Storage Temperature Range
Plastic DIP65°C to +125°C
Cerdip65°C to +150°C
Small Outline IC
Operating Temperature Range
AD810A
AD810S55°C to +125°C
Lead Temperature Range (Soldering 60 sec) +300°C

NOTES

¹Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum raring conditions for extended periods may affect device reliability. ²8-Pin Plastic Package: $\theta_{IA} = 90^{\circ}$ C/Watt; 8-Pin Cerdip Package: $\theta_{IA} = 110^{\circ}$ C/Watt; 8-Pin SOIC Package: $\theta_{IA} = 150^{\circ}$ C/Watt.

ESD SUSCEPTIBILITY

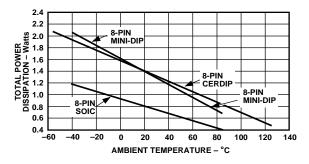
ESD (electrostatic discharge) sensitive device. Electrostatic charges as high as 4000 volts, which readily accumulate on the human body and on test equipment, can discharge without detection. Although the AD810 features ESD protection circuitry, permanent damage may still occur on these devices if they are subjected to high energy electrostatic discharges. Therefore, proper ESD precautions are recommended to avoid any performance degradation or loss of functionality.

ORDERING GUIDE

Model	Temperature Range	Package Description	Package Option	
AD810AN	-40°C to +85°C	8-Pin Plastic DIP	N-8	
AD810AR	-40°C to +85°C	8-Pin Plastic SOIC	R-8	
AD810AR-REEL	-40°C to +85°C	8-Pin Plastic SOIC	R-8	
5962-9313201MPA	-55°C to +125°C		Q-8	

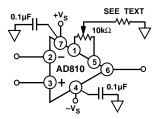
MAXIMUM POWER DISSIPATION

The maximum power that can be safely dissipated by the AD810 is limited by the associated rise in junction temperature. For the plastic packages, the maximum safe junction temperature is 145°C. For the cerdip package, the maximum junction temperature is 175°C. If these maximums are exceeded momentarily, proper circuit operation will be restored as soon as the die temperature is reduced. Leaving the device in the "overheated" condition for an extended period can result in device burnout. To ensure proper operation, it is important to observe the derating curves.



Maximum Power Dissipation vs. Temperature

While the AD810 is internally short circuit protected, this may not be sufficient to guarantee that the maximum junction temperature is not exceeded under all conditions.

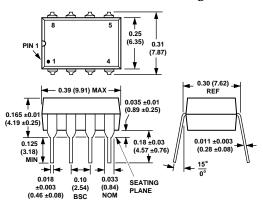


Offset Null Configuration

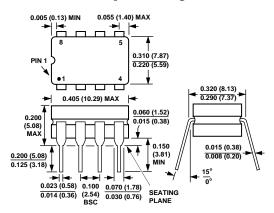
OUTLINE DIMENSIONS

Dimensions shown in inches and (mm).

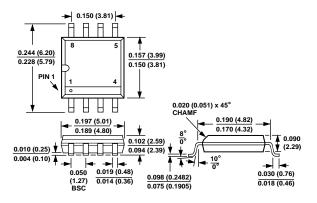
Plastic Mini-DIP (N) Package



Cerdip (Q) Package



8-Pin SOIC (R) Package



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