Preferred Device

Power MOSFET 12 Amps, 60 Volts

P-Channel TO-220

This Power MOSFET is designed to withstand high energy in the avalanche and commutation modes. Designed for low voltage, high speed switching applications in power supplies, converters and power motor controls, these devices are particularly well suited for bridge circuits where diode speed and commutating safe operating areas are critical and offer additional safety margin against unexpected voltage transients.

Features

- Avalanche Energy Specified
- I_{DSS} and V_{DS(on)} Specified at Elevated Temperature
- Pb-Free Package is Available*

MAXIMUM RATINGS (T_C = 25°C unless otherwise noted)

Rating	Symbol	Value	Unit					
Drain-to-Source Voltage	V _{DSS}	60	Vdc					
Drain-to-Gate Voltage (R_{GS} = 1.0 M Ω)	VDGR	60	Vdc					
Gate–to–Source Voltage – Continuous – Non–Repetitive (t _p ≤ 10 μs)	V _{GS} V _{GSM}	± 15 ± 25	Vdc Vpk					
Drain Current – Continuous – Continuous @ 100°C – Single Pulse (t _p ≤ 10 μs)	I _D I _D I _{DM}	12 8.0 42	Adc Apk					
Total Power Dissipation Derate above 25°C	PD	60 0.40	W W/°C					
Operating and Storage Temperature Range	T _J , T _{stg}	-55 to 175	°C					
$ \begin{array}{l} \mbox{Single Pulse Drain-to-Source Avalanche} \\ \mbox{Energy} - \mbox{Starting } T_J = 25^\circ C \\ \mbox{(V}_{DD} = 25 \mbox{ Vdc}, \mbox{V}_{GS} = 10 \mbox{ Vdc}, \mbox{Peak} \\ \mbox{I}_L = 12 \mbox{ Apk}, \mbox{L} = 3.0 \mbox{ mH}, \mbox{R}_G = 25 \Omega) \end{array} $	Eas	216	mJ					
Thermal Resistance – Junction-to-Case – Junction-to-Ambient	$R_{\theta JC}$ $R_{\theta JA}$	2.5 62.5	°C/W					
Maximum Lead Temperature for Soldering Purposes, 1/8" from case for 10 seconds	ΤL	260	°C					

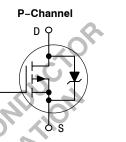
Stresses exceeding Maximum Ratings may damage the device. Maximum Ratings are stress ratings only. Functional operation above the Recommended Operating Conditions is not implied. Extended exposure to stresses above the Recommended Operating Conditions may affect device reliability.

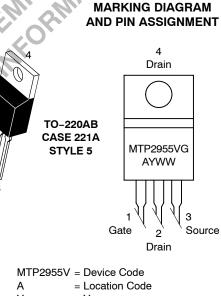


ON Semiconductor®

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12 AMPERES, 60 VOLTS $R_{DS(on)} = 230 \text{ m}\Omega$





A = Location Code Y = Year WW = Work Week G = Pb-Free Package

ORDERING INFORMATION

Device	Package	Shipping
MTP2955V	TO-220AB	50 Units/Rail
MTP2955VG	TO-220AB (Pb-Free)	50 Units/Rail

Preferred devices are recommended choices for future use

and best overall value.

*For additional information on our Pb-Free strategy and soldering details, please download the ON Semiconductor Soldering and Mounting Techniques Reference Manual, SOLDERRM/D.

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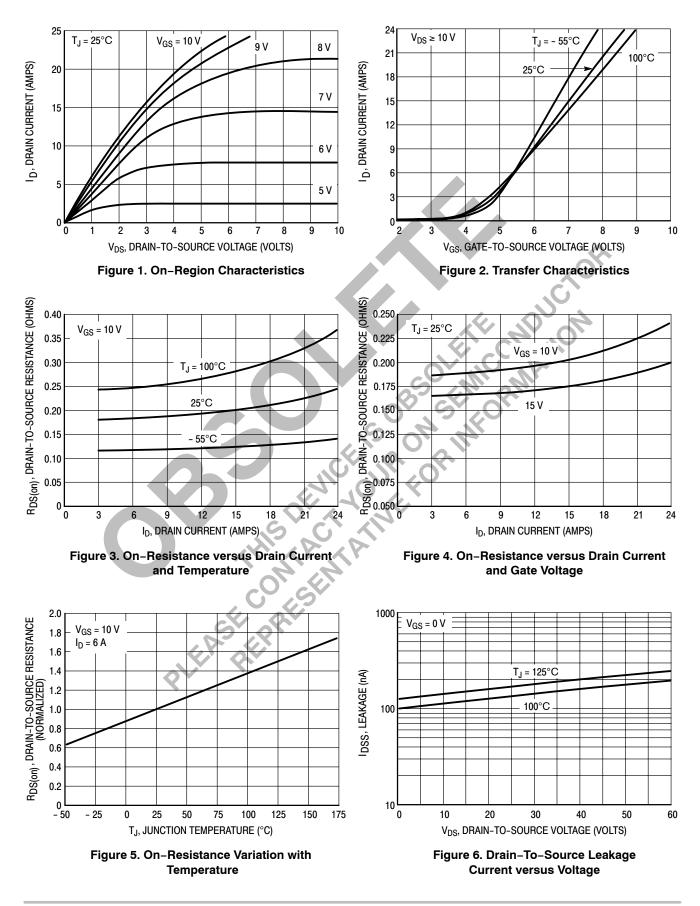
ELECTRICAL CHARACTERISTICS (T_J = $25^{\circ}C$ unless otherwise noted)

	Characteristic	Symbol	Min	Тур	Max	Unit
OFF CHARACTERISTICS						
Drain-to-Source Breakdown Volta (V _{GS} = 0 Vdc, I _D = 0.25 mAdc) Temperature Coefficient (Positiv		V _{(BR)DSS}	60 -	_ 58	-	Vdc mV/°C
Zero Gate Voltage Drain Current ($V_{DS} = 60 \text{ Vdc}, V_{GS} = 0 \text{ Vdc}$) ($V_{DS} = 60 \text{ Vdc}, V_{GS} = 0 \text{ Vdc}, T_{CS}$	J = 150°C)	I _{DSS}	-	-	10 100	μAdc
Gate-Body Leakage Current (V _{GS}	= ± 15 Vdc, V _{DS} = 0 Vdc)	I _{GSS}	-	-	100	nAdc
ON CHARACTERISTICS (Note 1)						
Gate Threshold Voltage $(V_{DS} = V_{GS}, I_D = 250 \ \mu Adc)$ Threshold Temperature Coeffici	(Cpk ≥ 2.0) (Note 3) ient (Negative)	V _{GS(th)}	2.0	2.8 5.0	4.0	Vdc mV/°C
Static Drain-to-Source On-Resist (V_{GS} = 10 Vdc, I_D = 6.0 Adc)	ance (Cpk ≥ 1.5) (Note 3)	R _{DS(on)}	-	0.185	0.230	Ω
$\begin{array}{l} \mbox{Drain-to-Source On-Voltage} \\ (V_{GS} = 10 \mbox{ Vdc}, \mbox{ I}_{D} = 12 \mbox{ Adc}) \\ (V_{GS} = 10 \mbox{ Vdc}, \mbox{ I}_{D} = 6.0 \mbox{ Adc}, \mbox{ T}_{D} \end{array}$	₁ = 150°C)	V _{DS(on)}	-	20	2.9 2.5	Vdc
Forward Transconductance (V _{DS} =	= 10 Vdc, I _D = 6.0 Adc)	9 _{FS}	3.0	5.0	-	mhos
DYNAMIC CHARACTERISTICS		4.	0	4		
Input Capacitance		C _{iss}	<u> </u>	550	700	pF
Output Capacitance	(V _{DS} = 25 Vdc, V _{GS} = 0 Vdc, f = 1.0 MHz)	C _{oss}		200	280	
Reverse Transfer Capacitance		C _{rss}	-	50	100	
SWITCHING CHARACTERISTICS	(Note 2)					
Turn-On Delay Time		t _{d(on)}	-	15	30	ns
Rise Time	(V _{DD} = 30 Vdc, I _D = 12 Adc,	tr	-	50	100	
Turn-Off Delay Time	$V_{GS} = 10 \text{ Vdc}, R_G = 9.1 \Omega)$	t _{d(off)}	-	24	50	
Fall Time		t _f	-	39	80	
Gate Charge		QT	-	19	30	nC
		Q ₁	-	4.0	-	
	$(V_{DS} = 48 \text{ Vdc}, I_D = 12 \text{ Adc}, V_{GS} = 10 \text{ Vdc})$	Q ₂	-	9.0	_	
		Q ₃	-	7.0	_	
SOURCE-DRAIN DIODE CHARAC	TERISTICS					
Forward On–Voltage (Note 1)	$(I_{S} = 12 \text{ Adc}, V_{GS} = 0 \text{ Vdc})$ $(I_{S} = 12 \text{ Adc}, V_{GS} = 0 \text{ Vdc}, T_{J} = 150^{\circ}\text{C})$	V _{SD}		1.8 1.5	3.0 _	Vdc
Reverse Recovery Time	N 18	t _{rr}	-	115	-	ns
	(I _S = 12 Adc, V _{GS} = 0 Vdc,	ta	-	90	-	
Q	$dI_{S}/dt = 100 \text{ A/}\mu\text{s}$	t _b	_	25	_	
Reverse Recovery Stored Charge	1	Q _{RR}	-	0.53	_	μC
NTERNAL PACKAGE INDUCTAN	CE	I				ı
Internal Drain Inductance (Measured from the drain lead 0.25	5" from package to center of die)	L _D	_	4.5	_	nH
Internal Source Inductance		L _S				nH

2. Switching characteristics are independent of operating junction temperature. 3. Reflects typical values. $C_{pk} = \left| \frac{Max limit - Typ}{3 \times SIGMA} \right|$

$$_{\rm ok} = \left| \frac{\text{Max limit} - 1y}{2x \text{ SIGMA}} \right|$$

TYPICAL ELECTRICAL CHARACTERISTICS



POWER MOSFET SWITCHING

Switching behavior is most easily modeled and predicted by recognizing that the power MOSFET is charge controlled. The lengths of various switching intervals (Δt) are determined by how fast the FET input capacitance can be charged by current from the generator.

The published capacitance data is difficult to use for calculating rise and fall because drain–gate capacitance varies greatly with applied voltage. Accordingly, gate charge data is used. In most cases, a satisfactory estimate of average input current ($I_{G(AV)}$) can be made from a rudimentary analysis of the drive circuit so that

 $t = Q/I_{G(AV)}$

During the rise and fall time interval when switching a resistive load, V_{GS} remains virtually constant at a level known as the plateau voltage, V_{SGP} . Therefore, rise and fall times may be approximated by the following:

 $t_r = Q_2 x R_G / (V_{GG} - V_{GSP})$

 $t_f = Q_2 x R_G / V_{GSP}$

where

 V_{GG} = the gate drive voltage, which varies from zero to V_{GG}

 R_G = the gate drive resistance

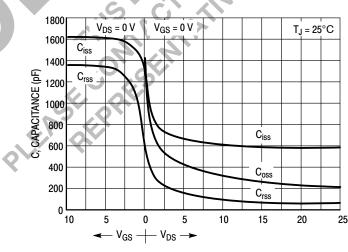
and Q_2 and V_{GSP} are read from the gate charge curve.

During the turn–on and turn–off delay times, gate current is not constant. The simplest calculation uses appropriate values from the capacitance curves in a standard equation for voltage change in an RC network. The equations are:

 $t_{d(on)} = R_G C_{iss} In \left[V_{GG} / (V_{GG} - V_{GSP}) \right]$ $t_{d(off)} = R_G C_{iss} In \left(V_{GG} / V_{GSP} \right)$ The capacitance (C_{iss}) is read from the capacitance curve at a voltage corresponding to the off-state condition when calculating $t_{d(on)}$ and is read at a voltage corresponding to the on-state when calculating $t_{d(off)}$.

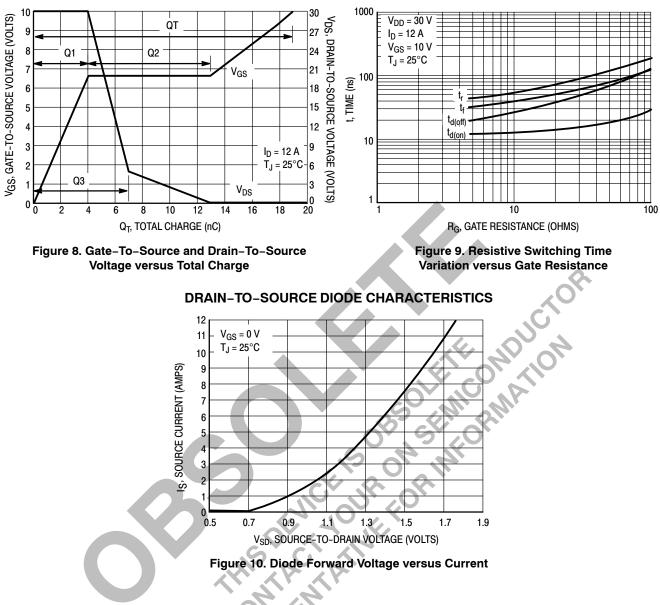
At high switching speeds, parasitic circuit elements complicate the analysis. The inductance of the MOSFET source lead, inside the package and in the circuit wiring which is common to both the drain and gate current paths, produces a voltage at the source which reduces the gate drive current. The voltage is determined by Ldi/dt, but since di/dt is a function of drain current, the mathematical solution is complex. The MOSFET output capacitance also complicates the mathematics. And finally, MOSFETs have finite internal gate resistance which effectively adds to the resistance of the driving source, but the internal resistance is difficult to measure and, consequently, is not specified.

The resistive switching time variation versus gate resistance (Figure 9) shows how typical switching performance is affected by the parasitic circuit elements. If the parasitics were not present, the slope of the curves would maintain a value of unity regardless of the switching speed. The circuit used to obtain the data is constructed to minimize common inductance in the drain and gate circuit loops and is believed readily achievable with board mounted components. Most power electronic loads are inductive; the data in the figure is taken with a resistive load, which approximates an optimally snubbed inductive load. Power MOSFETs may be safely operated into an inductive load; however, snubbing reduces switching losses.



GATE-TO-SOURCE OR DRAIN-TO-SOURCE VOLTAGE (VOLTS)

Figure 7. Capacitance Variation





The Forward Biased Safe Operating Area curves define the maximum simultaneous drain-to-source voltage and drain current that a transistor can handle safely when it is forward biased. Curves are based upon maximum peak junction temperature and a case temperature (T_C) of 25°C. Peak repetitive pulsed power limits are determined by using the thermal response data in conjunction with the procedures discussed in AN569, "Transient Thermal Resistance–General Data and Its Use."

Switching between the off-state and the on-state may traverse any load line provided neither rated peak current (I_{DM}) nor rated voltage (V_{DSS}) is exceeded and the transition time (t_r, t_f) do not exceed 10 µs. In addition the total power averaged over a complete switching cycle must not exceed ($T_{J(MAX)} - T_C$)/($R_{\theta JC}$).

A Power MOSFET designated E-FET can be safely used in switching circuits with unclamped inductive loads. For reliable operation, the stored energy from circuit inductance dissipated in the transistor while in avalanche must be less than the rated limit and adjusted for operating conditions differing from those specified. Although industry practice is to rate in terms of energy, avalanche energy capability is not a constant. The energy rating decreases non–linearly with an increase of peak current in avalanche and peak junction temperature.

Although many E–FETs can withstand the stress of drain–to–source avalanche at currents up to rated pulsed current (I_{DM}), the energy rating is specified at rated continuous current (I_D), in accordance with industry custom. The energy rating must be derated for temperature as shown in the accompanying graph (Figure 13). Maximum energy at currents below rated continuous I_D can safely be assumed to equal the values indicated.

SAFE OPERATING AREA

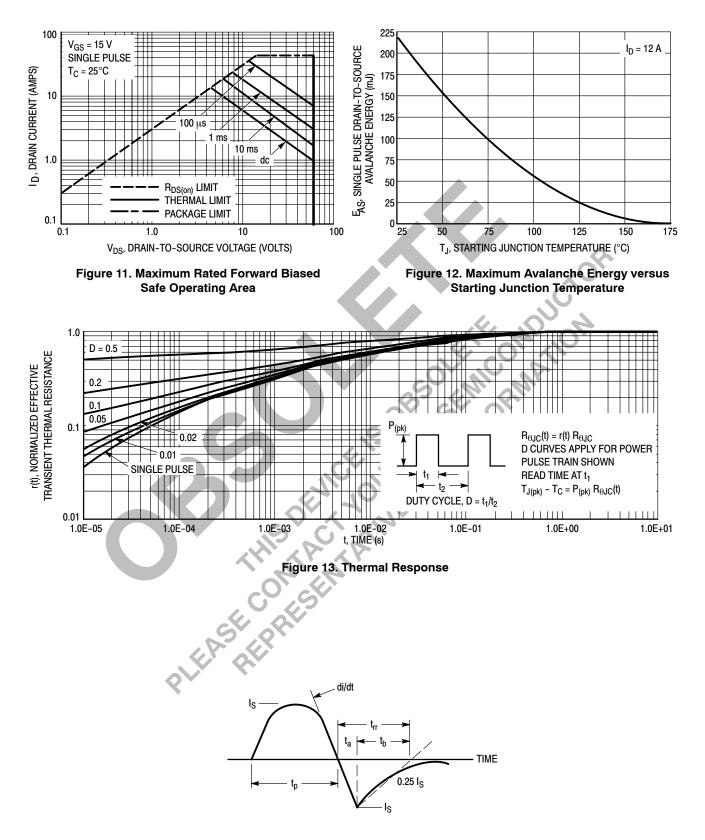
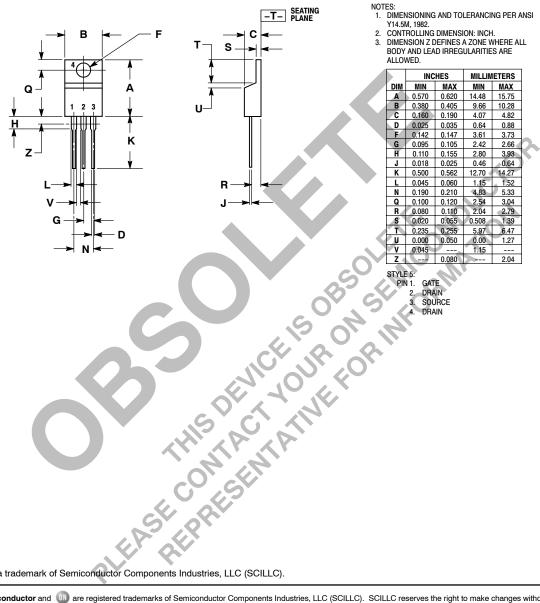


Figure 14. Diode Reverse Recovery Waveform

PACKAGE DIMENSIONS





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