

FOD2742A FOD2742B FOD2742C

DESCRIPTION

The FOD2742 Optically Isolated Amplifier consists of the popular KA431 precision programmable shunt reference and an optocoupler. The optocoupler is a gallium arsenide (GaAs) light emitting diode optically coupled to a silicon phototransistor. It comes in 3 grades of reference voltage tolerance = 2%, 1%, and 0.5%.

The Current Transfer Ratio (CTR) ranges from 100% to 200%. It also has an outstanding temperature coefficient of 50 ppm/°C. It is primarily intended for use as the error amplifier/reference voltage/optocoupler function in isolated ac to dc power supplies and dc/dc converters.

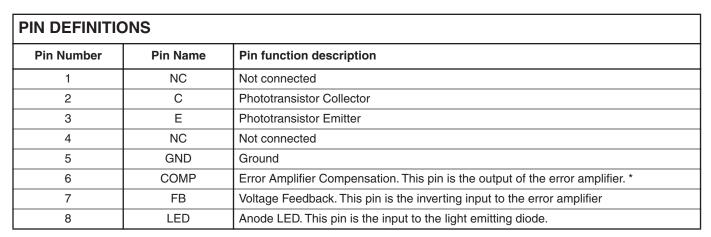
When using the FOD2742, power supply designers can reduce the component count and save space in tightly packaged designs. The tight tolerance reference eliminates the need for adjustments in many applications. The device comes in a 8-pin small outline package.

FEATURES

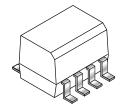
- · Optocoupler, precision reference and error amplifier in single package
- · 2.5V reference
- CTR 100% to 200%
- · 2,500V RMS isolation
- UL approval E90700, Volume 2
 BSI approval 8661, 8662
- VDE approval 136616
- CSA approval 1113643
- Low temperature coefficient 50 ppm/°C max
- FOD2742A: tolerance 0.5%
 FOD2742B: tolerance 1%
 FOD2742C: tolerance 2%

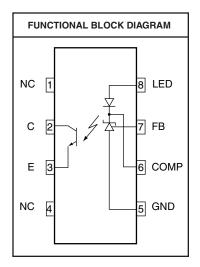
APPLICATIONS

- · Power supplies regulation
- DC to DC converters



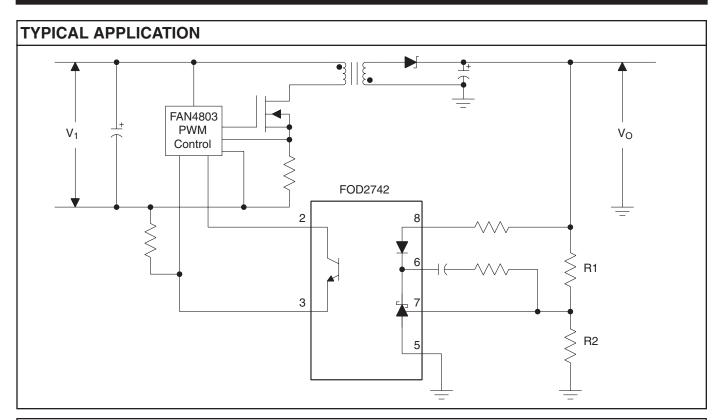
^{*} The compensation network must be attached between pins 6 and 7.







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ABSOLUTE MAXIMUM RATINGS (T _A = 25°C Unless otherwise specified.)					
Parameter	Symbol	Value	Units		
Storage Temperature	T _{STG}	-40 to +125	°C		
Operating Temperature	T _{OPR}	-25 to +85	°C		
Reflow Temperature Profile (refer to fig. 21)					
Input Voltage	V_{LED}	37	V		
Input DC Current	I _{LED}	20	mA		
Collector-Emitter Voltage	V _{CEO}	70	V		
Emitter-Collector Voltage	V _{ECO}	7	V		
Collector Current	I _C	50	mA		
Input Power Dissipation (note 1)	PD1	145	mW		
Transistor Power Dissipation (note 2)	PD2	85	mW		
Total Power Dissipation (note 3)	PD3	145	mW		

Notes

- 1. Derate linearly from 25°C at a rate of 2.42 mW/ °C
- 2. Derate linearly from 25°C at a rate of 1.42 mW/ °C.
- 3. Derate linearly from 25°C at a rate of 2.42 mW/ °C.
- 4. Functional operation under these conditions is not implied. Permanent damage may occur if the device is subjected to conditions outside these ratings.



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ELECTRICAL CHARACTERISTICS (T_A = 25°C Unless otherwise specified.)

INPUT CHARACTERISTI	CS							
Parameter		Test Conditions	Symbol	Device	Min.	Тур.	Max.	Unit
LED Forward Voltage	(I _{LED} = 10 m	$A, V_{COMP} = V_{FB}$ (fig. 1)	V _F	ALL		1.20	1.5	٧
				Α	2.482	2.495	2.508	V
Reference Voltage	I _{LED} = 10 r	mA , $V_{COMP} = V_{FB}$ (fig. 1)	V_{REF}	В	2.470	2.495	2.520	V
				С	2.450	2.500	2.550	V
Deviation of V _{REF} over temperature	T _A :	= -25°C to +85°C (fig. 1)	V _{REF (DEV)}	ALL		3.5	17	mV
Ratio of V _{REF} variation to the	I _{LED} = 10 mA	$\Delta V_{COMP} = 10V \text{ to } V_{REF}$	ΔV _{REF} /	A1.1		-0.5	-2.7	mV/
output of the error amplifier		ΔV_{COMP}	ALL		-0.3	-2.0	V	
Feedback Input Current	I _{LED} = 1	0mA, $R_1 = 10$ KΩ (fig. 3)	I _{REF}	ALL		2.2	4	μΑ
Deviation of I _{REF} over temperature	T _A :	= -25°C to +85°C (fig. 3)	I _{REF (DEV)}	ALL		1.0	1.2	μA
Minimum Drive Current		$V_{COMP} = V_{FB}$ (fig. 1)	I _{LED (MIN)}	ALL		0.45	1.0	mA
Off-state error amplifier current	V _{LE}	$ED = 37V, V_{FB} = 0 \text{ (fig. 4)}$	I _(OFF)	ALL		0.01	1.0	μΑ
Error amplifier output impedance (see note 2)	$V_{COMP} = V_{R}$	$I_{LED} = 1 \text{mA to } 20 \text{mA},$ $f \ge 1.0 \text{ kHz}$	IZ _{OUT} I	ALL		0.15	0.5	Ω

 The deviation parameters V_{REF(DEV)} and I_{REF(DEV)} are defined as the differences between the maximum and minimum values obtained over the rated temperature range. The average full-range temperature coefficient of the reference input voltage, ΔV_{REF} is defined as:

$$\left|\Delta V_{REF}\right|(ppm/^{\circ}C) \ = \ \frac{\left\{V_{REF(DEV)}/V_{REF}(T_{A}=25^{\circ}C)\right\} \times 10^{6}}{\Delta T_{A}}$$

where ΔT_A is the rated operating free-air temperature range of the device.

2. The dynamic impedance is defined as $|Z_{OUT}| = \Delta V_{COMP}/\Delta I_{LED}$. When the device is operating with two external resistors (see Figure 2), the total dynamic impedance of the circuit is given by:

$$\left|Z_{OUT,\;TOT}\right| = \frac{\Delta V}{\Delta I} \approx \left|Z_{OUT}\right| \times \left\lceil 1 + \frac{R1}{R2} \right\rceil$$



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OUTPUT CHARACTERISTICS (T _A = 25°C Unless otherwise specified.)						
Parameter	Test Conditions	Symbol	Min	Тур	Max	Unit
Collector dark current	(V _{CE} = 10 V) (Fig. 5)	I _{CEO}		1	50	nA
Emitter-collector voltage breakdown	(I _E = 100 μA)	BV _{ECO}	7	10		V
Collector-emitter voltage breakdown	$(I_C = 1.0 \text{mA})$	BV _{CEO}	70	120		V

TRANSFER CHARACTERISTICS (T _A = 25°C Unless otherwise specified.)						
Parameter	Test Conditions	Symbol	Min	Тур	Max	Unit
Current transfer ratio	$(I_{LED} = 10 \text{ mA}, V_{COMP} = V_{FB},$ $V_{CE} = 5 \text{ V}) \text{ (Fig. 6)}$	CTR	100	140	200	%
Collector-emitter saturation voltage	$(I_{LED} = 10 \text{ mA}, V_{COMP} = V_{FB,}$ $I_{C} = 2.5 \text{ mA}) \text{ (Fig. 6)}$	V _{CE (SAT)}		0.16	0.4	V

ISOLATION CHARACTERISTICS (T _A = 25°C Unless otherwise specified.)						
Parameter	Test Conditions	Symbol	Min	Тур	Max	Unit
Input-output insulation leakage current	(RH = 45%, T_A = 25°C, t = 5s, V_{I-O} = 3000 VDC) (note. 1)	I _{I-O}			1.0	μΑ
Withstand insulation voltage	(RH <= 50%, T _A = 25°C, t = 1 min) (note 1)	V _{ISO}	2500			Vrms
Resistance (input to output)	V _{I-O} = 500 VDC (note 1)	R _{I-O}		10 ¹²		Ohm

SWITCHING CHARACTERISTICS (T _A = 25°C Unless otherwise specified.)						
Parameter	Test Conditions	Symbol	Min	Тур	Max	Unit
Bandwidth	(Fig. 7)	BW		50		kHZ
Common mode transient immunity at output high	$(I_{LED} = 0 \text{ mA}, Vcm = 10 V_{PP}$ RL = 2.2 kV (Fig. 8) (note 2)			1.0		kV/μs
Common mode transient immunity at output low	$(I_{LED} = 10 \text{ mA}, Vcm = 10 V_{PP}$ RL = 2.2 kV (Fig. 8) (note 2)	CML		1.0		kV/μs

Notes

- 1. Device is considered as a two terminal device: Pins 1,2 3 and 4 are shorted together and Pins 5,6,7 and 8 are shorted together.
- 2. Common mode transient immunity at output high is the maximum tolerable (positive) dVcm/dt on the leading edge of the common mode impulse signal, Vcm, to assure that the output will remain high. Common mode transient immunity at output low is the maximum tolerable (negative) dVcm/dt on the trailing edge of the common pulse signal, Vcm, to assure that the output will remain low.



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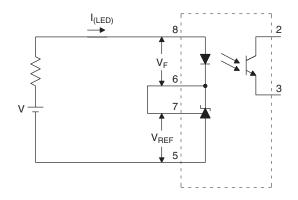


FIG. 1. V_{REF}, V_F, I_{LED} (min) TEST CIRCUIT

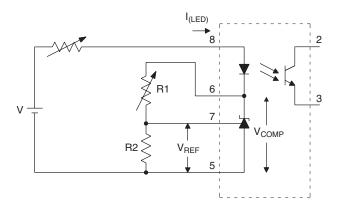


FIG. 2. $\Delta V_{REF}/\Delta V_{COMP}$ TEST CIRCUIT

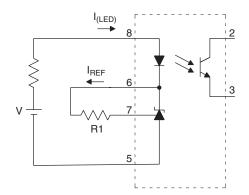


FIG. 3. I_{REF} TEST CIRCUIT

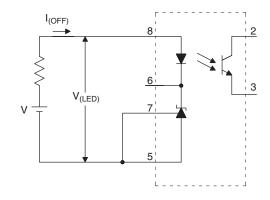


FIG. 4. I_(OFF) TEST CIRCUIT

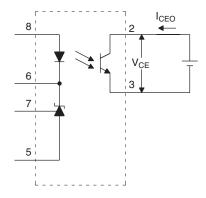


FIG. 5. I_{CEO} TEST CIRCUIT

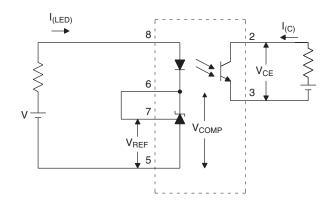


FIG. 6. CTR, V_{CE(sat)} TEST CIRCUIT



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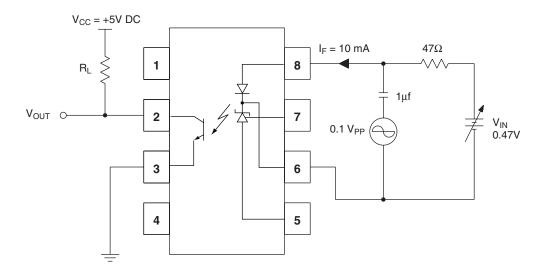


Fig. 7 Frequency Response Test Circuit

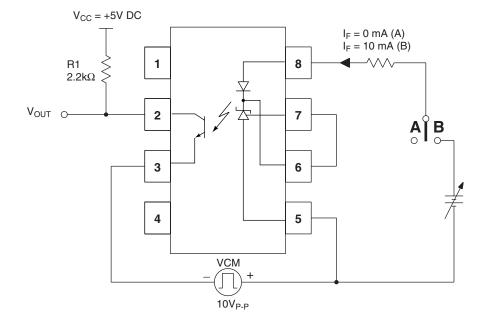
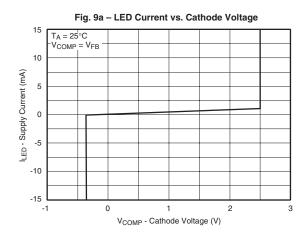


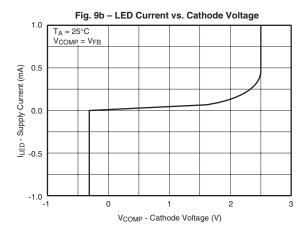
Fig. 8 CMH and CML Test Circuit

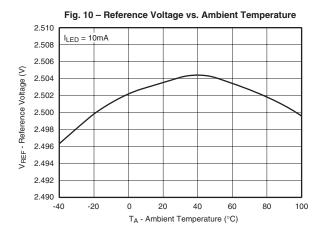


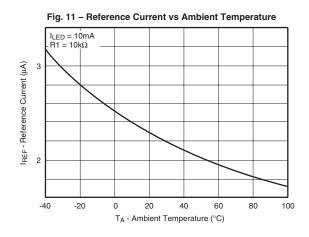
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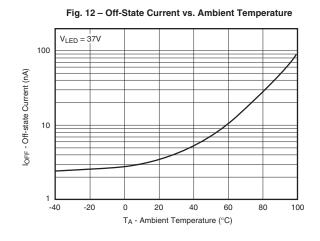
TYPICAL PERFORMANCE CURVES

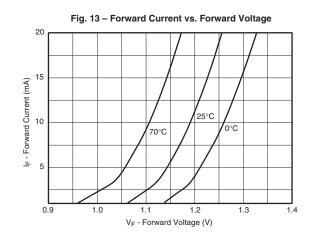








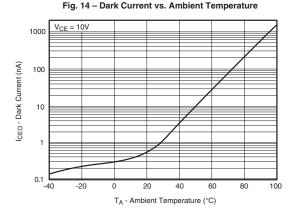


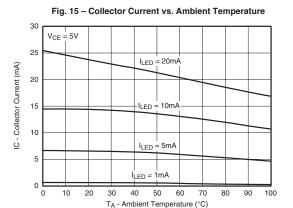


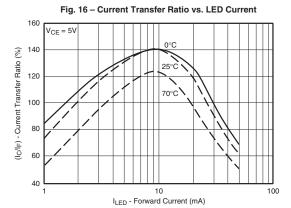


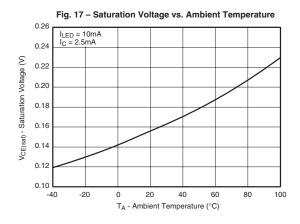
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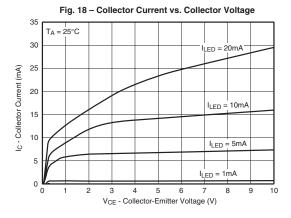
TYPICAL PERFORMANCE CURVES

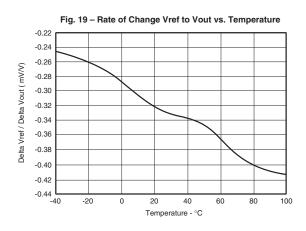








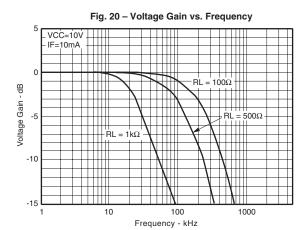






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TYPICAL PERFORMANCE CURVES





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The FOD2742

The FOD2742 is an optically isolated error amplifier. It incorporates three of the most common elements necessary to make an isolated power supply, a reference voltage, an error amplifier, and an optocoupler. It is functionally equivalent to the popular KA431 shunt voltage regulator plus the CNY17F-X optocoupler.

Powering the Secondary Side

The LED pin in the FOD2742 powers the secondary side, and in particular provides the current to run the LED. The actual structure of the FOD2742 dictates the minimum voltage that can be applied to the LED pin: The error amplifier output has a minimum of the reference voltage, and the LED is in series with that. Minimum voltage applied to the LED pin is thus 2.5V \pm 1.5V \pm 4.0V. This voltage can be generated either directly from the output of the converter, or else from a slaved secondary winding. The secondary winding will not affect regulation, as the input to the FB pin may still be taken from the output winding.

The LED pin needs to be fed through a current limiting resistor. The value of the resistor sets the amount of current through the LED, and thus must be carefully selected in conjunction with the selection of the primary side resistor.

Feedback

Output voltage of a converter is determined by selecting a resistor divider from the regulated output to the FB pin. The FOD2742 attempts to regulate its FB pin to the reference voltage, 2.5V. The ratio of the two resistors should thus be:

$$\frac{R_{TOP}}{R_{BOTTOM}} = \frac{V_{OUT}}{V_{BEE}} - 1$$

The absolute value of the top resistor is set by the input offset current of 5.2 μ A. To achieve 0.5% accuracy, the resistance of R_{TOP} should be:

$$\frac{V_{OUT}-2.5}{R_{TOP}} > 1040 \mu A$$

Compensation

The compensation pin of the FOD2742 provides the opportunity for the designer to design the frequency response of the converter. A compensation network may be placed between the COMP pin and the FB pin. In typical low-bandwidth systems, a 0.1µF capacitor may be used. For converters with more stringent requirements, a network should be designed based on measurements of the system's loop. An excellent reference for this process may be found in "Practical Design of Power Supplies" by Ron Lenk, IEEE Press, 1998.

Secondary Ground

The GND pin should be connected to the secondary ground of the converter.

No Connect Pins

The NC pins have no internal connection. They should not have any connection to the secondary side, as this may compromise the isolation structure.

Photo-Transistor

The Photo-transistor is the output of the FOD2742. In a normal configuration the collector will be attached to a pull-up resistor and the emitter grounded. There is no base connection necessary.

The value of the pull-up resistor, and the current limiting resistor feeding the LED, must be carefully selected to account for voltage range accepted by the PWM IC, and for the variation in current transfer ratio (CTR) of the opto-isolator itself.

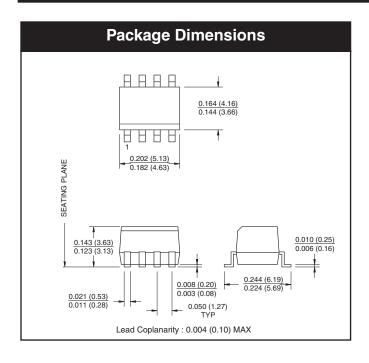
Example: The voltage feeding the LED pins is +12V, the voltage feeding the collector pull-up is +10V, and the PWM IC is the Fairchild KA1H0680, which has a 5V reference. If we select a 10KV resistor for the LED, the maximum current the LED can see is (12V-4V) /10K Ω = 800 μ A. The CTR of the opto-isolator is a minimum of 100%, so the minimum collector current of the photo-transistor when the diode is full on is also 800 μ A. The collector resistor must thus be such that:

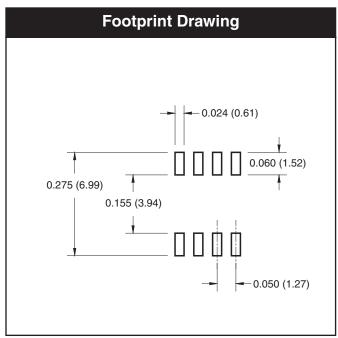
$$\frac{10V - 5V}{R_{COLLECTOR}}$$
 < 800 μ A or $R_{COLLECTOR}$ > 6.25 $K\Omega$;

select $12K\Omega$ to allow some margin.



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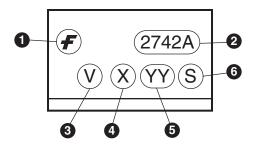




ORDERING INFORMATION

Option	Order Entry Identifier	Description
R1	R1	Tape and reel (500 units per reel)
R1V	R1V	VDE0884, Tape and reel (500 units per reel)
R2	R2	Tape and reel (2,500 units per reel)
R2V	R2V	VDE0884, Tape and reel (2,500 units per reel)

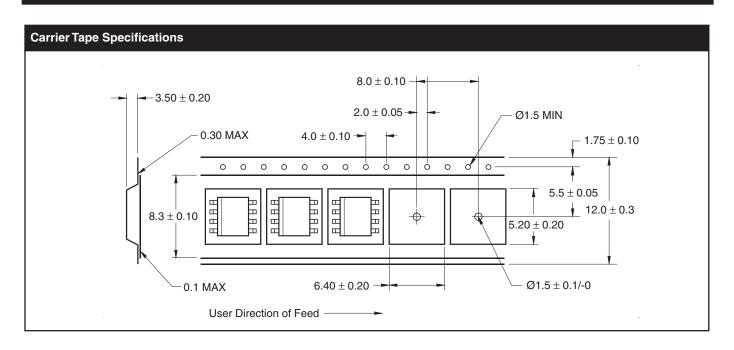
MARKING INFORMATION

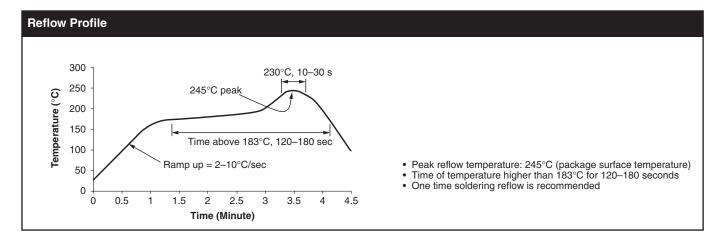


Definitions					
1	Fairchild logo				
2	Device number				
3	VDE mark (Note: Only appears on parts ordered with VDE option – See order entry table)				
4	One digit year code				
5	Two digit work week ranging from '01' to '53'				
6	Assembly package code				



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