

# LOW SKEW, 1-TO-2, DIFFERENTIAL/LVCMOS-TO-0.7V HCSL FANOUT BUFFER

ICS85102I

## GENERAL DESCRIPTION



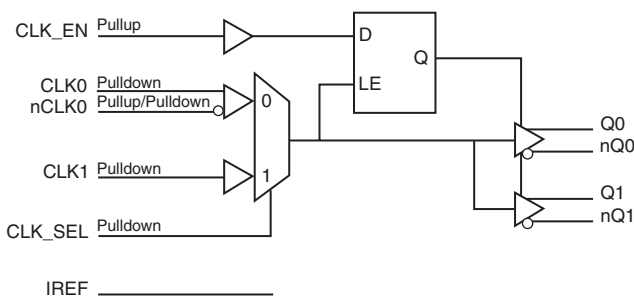
The ICS85102I is a low skew, high performance 1-to-2 Differential-to-HCSL fanout buffer and a member of the HiPerClockS™ family of High Performance Clock Solutions from IDT. The ICS85102I has a differential clock input. The CLK0, nCLK0 input pair can accept most standard differential input levels. The clock enable is internally synchronized to eliminate runt clock pulses on the output during asynchronous assertion/deassertion of the clock enable pin.

Guaranteed output and part-to-part skew characteristics make the ICS85102I ideal for those applications demanding well defined performance and repeatability.

## FEATURES

- Two 0.7V differential HCSL outputs
- Selectable differential CLK0, nCLK0 or LVCMOS inputs
- CLK0, nCLK0 pair can accept the following differential input levels: LVPECL, LVDS, LVHSTL, HCSL, SSTL
- CLK1 can accept the following input levels: LVCMOS or LVTTTL
- Maximum output frequency: 500MHz
- Translates any single-ended input signal to 3.3V HCSL levels with resistor bias on nCLK input
- Output skew: 65ps (maximum)
- Part-to-part skew: 600ps (maximum)
- Propagation delay: 3.2ns (maximum)
- Additive phase jitter, RMS: 0.14ps typical @ 250MHz
- 3.3V operating supply
- -40°C to 85°C ambient operating temperature
- Available in both standard (RoHS 5) and lead-free (RoHS 6) packages

## BLOCK DIAGRAM



## PIN ASSIGNMENT

CLK_EN	1	16	GND
CLK_SEL	2	15	V <sub>DD</sub>
CLK0	3	14	Q0
nCLK0	4	13	nQ0
CLK1	5	12	Q1
nc	6	11	nQ1
nc	7	10	V <sub>DD</sub>
IREF	8	9	V <sub>DD</sub>

### ICS85102I 16-Lead TSSOP

4.4mm x 5.0mm x 0.925mm body package  
G Package  
Top View

TABLE 1. PIN DESCRIPTIONS

Number	Name	Type		Description
1	CLK_EN	Input	Pullup	Synchronizing clock enable. When HIGH, clock outputs follow clock input. When LOW, Qx outputs are forced low, nQx outputs are forced high. LVTTTL / LVCMOS interface levels.
2	CLK_SEL	Input	Pulldown	Clock select input. When HIGH, selects CLK1 input. When LOW, selects CLK0, nCLK0 inputs. LVTTTL / LVCMOS interface levels.
3	CLK0	Input	Pulldown	Non-inverting differential clock input.
4	nCLK0	Input	Pullup/ Pulldown	Inverting differential clock input.
5	CLK1		Pulldown	Single-ended clock input. LVTTTL / LVCMOS interface levels.
6, 7	nc	Unused		No connect.
8	IREF	Input		An external fixed resistor (475Ω) from this pin to ground provides a reference current used for differential current-mode Qx/nQx clock outputs.
9, 10, 15	V <sub>DD</sub>	Power		Positive supply pins.
11, 12	nQ1, Q1	Output		Differential output pair. HCSSL interface levels.
13, 14	nQ0, Q0	Output		Differential output pair. HCSSL interface levels.
16	GND	Power		Power supply ground.

NOTE: *Pullup* and *Pulldown* refer to internal input resistors. See Table 2, Pin Characteristics, for typical values.

TABLE 2. PIN CHARACTERISTICS

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
C <sub>IN</sub>	Input Capacitance			4		pF
R <sub>PULLUP</sub>	Input Pullup Resistor			51		kΩ
R <sub>PULLDOWN</sub>	Input Pulldown Resistor			51		kΩ

TABLE 3A. CONTROL INPUT FUNCTION TABLE

Inputs			Outputs	
CLK_EN	CLK_SEL	Selected Source	Q0:Q1	nQ0:nQ1
0	0	CLK0, nCLK0	Disabled; LOW	Disabled; HIGH
0	1	CLK1	Disabled; LOW	Disabled; HIGH
1	0	CLK0, nCLK0	Enabled	Enabled
1	1	CLK1	Enabled	Enabled

After CLK\_EN switches, the clock outputs are disabled or enabled following a falling input clock edge as shown in Figure 1.

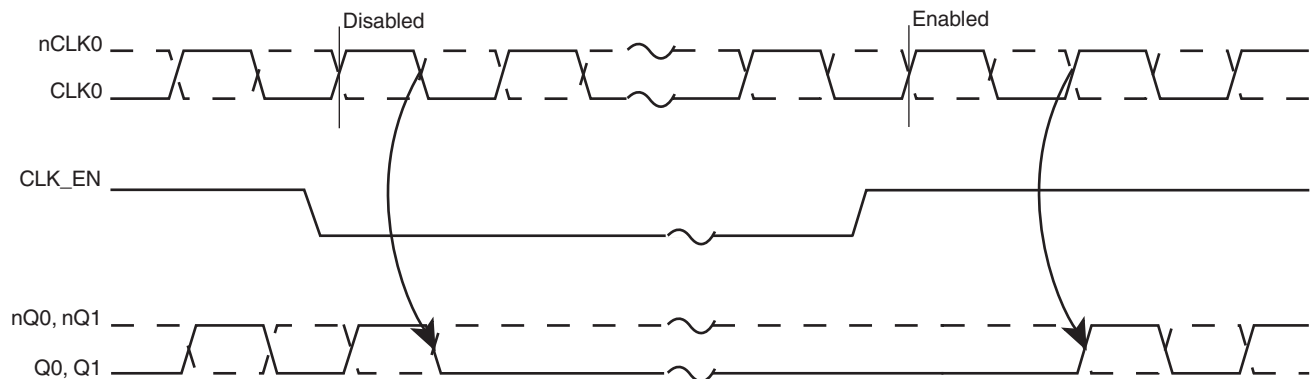


FIGURE 1. CLK\_EN TIMING DIAGRAM

## ABSOLUTE MAXIMUM RATINGS

Supply Voltage, $V_{CC}$	4.6V
Inputs, $V_I$	-0.5V to $V_{DD} + 0.5V$
Outputs, $I_O$	
Continuous Current	50mA
Surge Current	100mA
Package Thermal Impedance, $\theta_{JA}$	100.3°C/W (0 mps)
Storage Temperature, $T_{STG}$	-65°C to 150°C

NOTE: Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These ratings are stress specifications only. Functional operation of product at these conditions or any conditions beyond those listed in the *DC Characteristics* or *AC Characteristics* is not implied. Exposure to absolute maximum rating conditions for extended periods may affect product reliability.

TABLE 4A. POWER SUPPLY DC CHARACTERISTICS,  $V_{DD} = 3.3V \pm 10\%$ ,  $T_A = -40^\circ C$  TO  $85^\circ C$ 

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
$V_{DD}$	Positive Supply Voltage		2.97	3.3	3.63	V
$I_{DD}$	Power Supply Current	Unterminated			27	mA

TABLE 4B. LVCMOS / LVTTTL DC CHARACTERISTICS,  $V_{DD} = 3.3V \pm 10\%$ ,  $T_A = -40^\circ C$  TO  $85^\circ C$ 

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
$V_{IH}$	Input High Voltage		2		$V_{DD} + 0.3$	V
$V_{IL}$	Input Low Voltage		-0.3		0.8	V
$I_{IH}$	Input High Current	CLK1, CLK_SEL	$V_{IN} = V_{DD} = 3.63V$		150	$\mu A$
		CLK_EN	$V_{IN} = V_{DD} = 3.63V$		5	$\mu A$
$I_{IL}$	Input Low Current	CLK1, CLK_SEL	$V_{IN} = 0V, V_{DD} = 3.63V$	-5		$\mu A$
		CLK_EN	$V_{IN} = 0V, V_{DD} = 3.63V$	-150		$\mu A$

TABLE 4C. DIFFERENTIAL DC CHARACTERISTICS,  $V_{DD} = 3.3V \pm 10\%$ ,  $T_A = -40^\circ C$  TO  $85^\circ C$ 

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
$I_{IH}$	Input High Current	CLK0/nCLK0	$V_{DD} = V_{IN} = 3.63V$		150	$\mu A$
$I_{IL}$	Input Low Current	CLK0	$V_{DD} = 3.63V, V_{IN} = 0V$	-5		$\mu A$
		nCLK0	$V_{DD} = 3.63V, V_{IN} = 0V$	-150		$\mu A$
$V_{PP}$	Peak-to-Peak Input Voltage; NOTE 1		0.15		1.3	V
$V_{CMR}$	Common Mode Input Voltage; NOTE 1, 2		GND + 0.5		$V_{DD} - 0.85$	V

NOTE 1:  $V_{IL}$  should not be less than -0.3V.

NOTE 2: Common mode voltage is defined as  $V_{IH}$ .

TABLE 5. AC CHARACTERISTICS,  $V_{DD} = 3.3V \pm 10\%$ ,  $T_A = -40^\circ C$  TO  $85^\circ C$ 

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
$f_{MAX}$	Output Frequency	CLK_SEL = 0			500	MHz
		CLK_SEL = 1			250	MHz
$t_{PD}$	Propagation Delay; NOTE 1	CLK_SEL = 0	2.0		3.2	ns
		CLK_SEL = 1	2.0		2.8	ns
$t_{sk(o)}$	Output Skew; NOTE 2, 4				65	ps
$t_{sk(pp)}$	Part-to-Part Skew; NOTE 3, 4				600	ps
$t_{jit}$	Buffer Additive Phase Jitter, RMS; refer to Additive Phase Jitter Section	100MHz (12kHz - 20MHz)		0.22		ps
		250MHz (12kHz - 20MHz)		0.14		ps
$V_{MAX}$	Absolute Maximum Output Voltage; NOTE 5, 10				1150	mV
$V_{MIN}$	Absolute Minimum Output Voltage; NOTE 5, 11		-300			mV
$V_{RB}$	Ringback Voltage; NOTE 6, 13		-100		100	V
$t_{STABLE}$	Time before $V_{RB}$ is allowed; NOTE 6, 13		500			ps
$V_{CROSS}$	Absolute Crossing Voltage; NOTE 5, 8, 9		250		550	mV
$\Delta V_{CROSS}$	Total Variation of $V_{CROSS}$ over all edges; NOTE 5, 8, 12				140	mV
	Rise/Fall Edge Rate; NOTE 6, 7	Measured between -150mV to +150mV	0.6		5.5	V/ns
odc	Output Duty Cycle; NOTE 14		45		55	%

All parameters measured at  $f \leq 250\text{MHz}$  unless noted otherwise.

NOTE 1: Measured from the  $V_{DD}/2$  of the input to the differential output crossing point.

NOTE 2: Defined as skew between outputs at the same supply voltage and with equal load conditions. Measured at the output differential cross points.

NOTE 3: Defined as skew between outputs on different devices operating at the same supply voltage, same temperature, and with equal load conditions. Using the same type of inputs on each device, the outputs are measured at the differential cross points.

NOTE 4: This parameter is defined in accordance with JEDEC Standard 65.

NOTE 5: Measurement taken from single-ended waveform.

NOTE 6: Measurement taken from differential waveform.

NOTE 7: Measured from -150mV to +150mV on the differential waveform (derived from Qx minus nQx). The signal must be monotonic through the measurement region for rise and fall time. The 300mV measurement window is centered on the differential zero crossing. See Parameter Measurement Information Section.

NOTE 8: Measured at crossing point where the instantaneous voltage value of the rising edge of Qx equals the falling edge of nQx. See Parameter Measurement Information Section.

NOTE 9: Refers to the total variation from the lowest crossing point to the highest, regardless of which edge is crossing. Refers to all crossing points for this measurement. See Parameter Measurement Information Section.

NOTE 10: Defined as the maximum instantaneous voltage including overshoot. See Parameter Measurement Information Section.

NOTE 11: Defined as the minimum instantaneous voltage including undershoot. See Parameter Measurement Information Section.

NOTE 12: Defined as the total variation of all crossing voltage of Rising Qx and Falling nQx. This is the maximum allowed variance in the  $V_{CROSS}$  for any particular system. See Parameter Measurement Information Section.

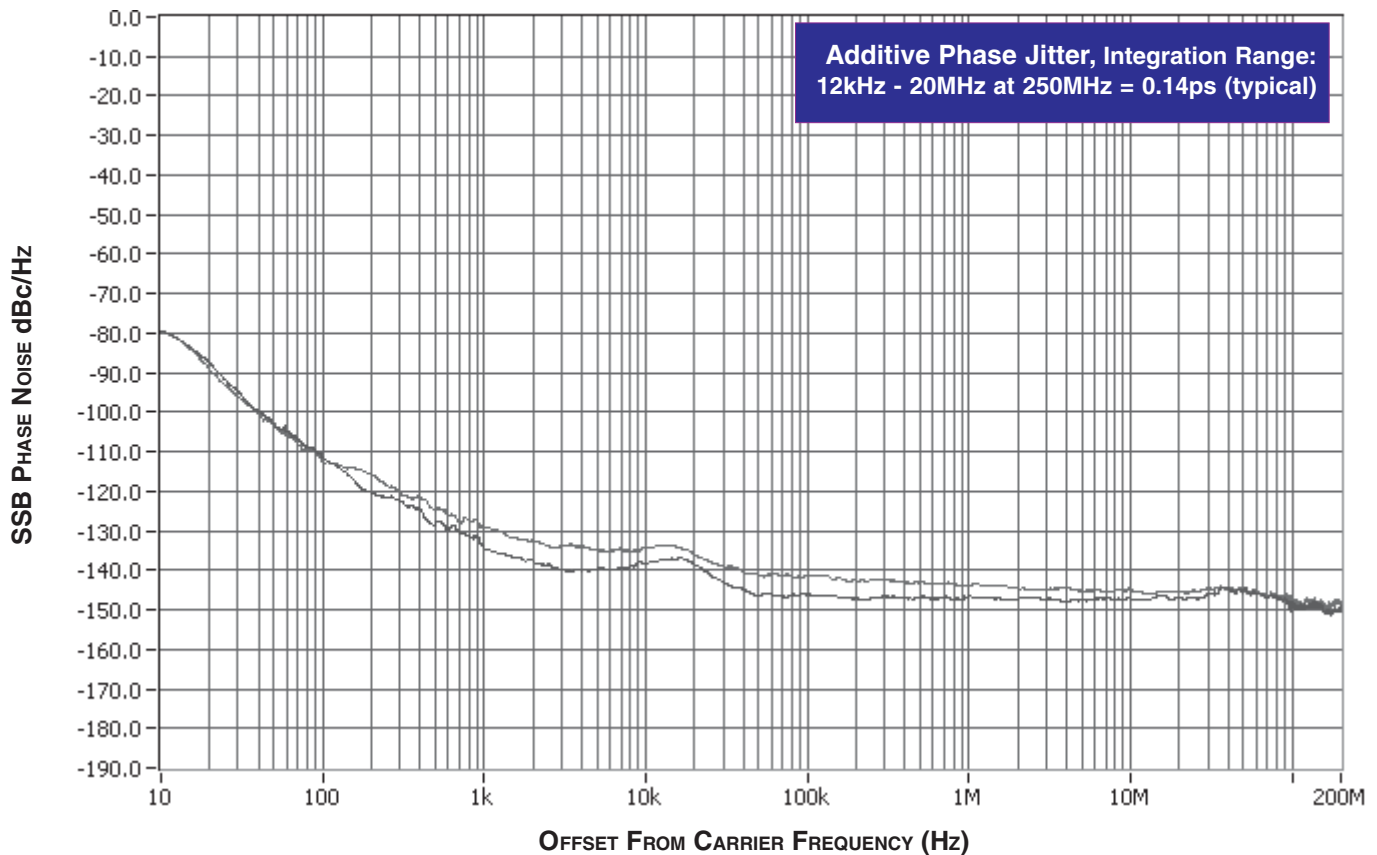
NOTE 13:  $T_{STABLE}$  is the time the differential clock must maintain a minimum  $\pm 150\text{mV}$  differential voltage after rising/falling edges before it is allowed to droop back into the  $V_{RB} \pm 100\text{mV}$  differential range. See Parameter Measurement Information Section.

NOTE 14: Input duty cycle must be 50%.

## ADDITIVE PHASE JITTER

The spectral purity in a band at a specific offset from the fundamental compared to the power of the fundamental is called the ***dBc Phase Noise***. This value is normally expressed using a Phase noise plot and is most often the specified plot in many applications. Phase noise is defined as the ratio of the noise power present in a 1Hz band at a specified offset from the fundamental frequency to the power value of the fundamental. This ratio is expressed in decibels (dBm) or a ratio of the power in the 1Hz

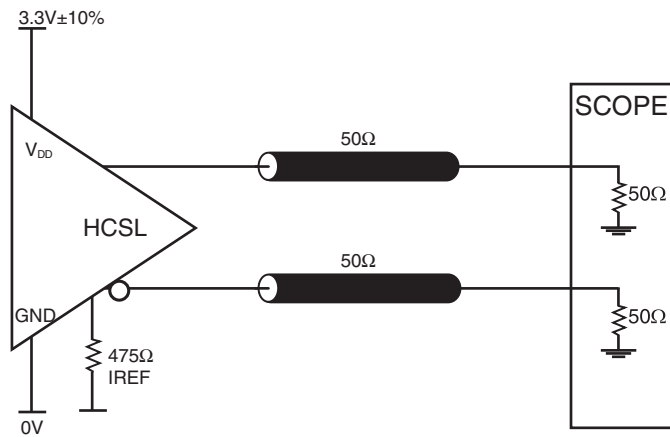
band to the power in the fundamental. When the required offset is specified, the phase noise is called a ***dBc*** value, which simply means dBm at a specified offset from the fundamental. By investigating jitter in the frequency domain, we get a better understanding of its effects on the desired application over the entire time record of the signal. It is mathematically possible to calculate an expected bit error rate given a phase noise plot.



As with most timing specifications, phase noise measurements has issues relating to the limitations of the equipment. Often the noise floor of the equipment is higher than the noise floor of the

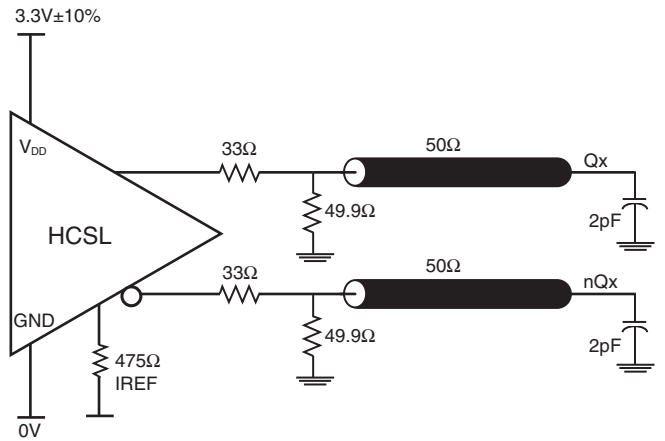
device. This is illustrated above. The device meets the noise floor of what is shown, but can actually be lower. The phase noise is dependent on the input source and measurement equipment.

## PARAMETER MEASUREMENT INFORMATION

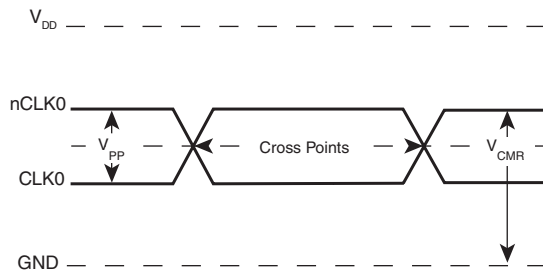


This load condition is used for  $I_{DD}$ ,  $t_{sk}(o)$ ,  $t_{PD}$ , and  $f_{jit}$  measurements.

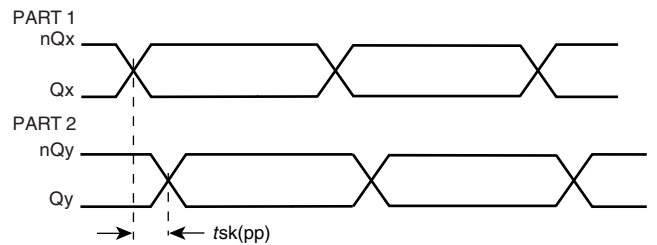
**HCSL OUTPUT LOAD AC TEST CIRCUIT**



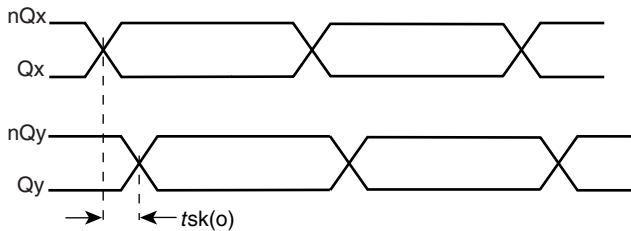
**HCSL OUTPUT LOAD AC TEST CIRCUIT**



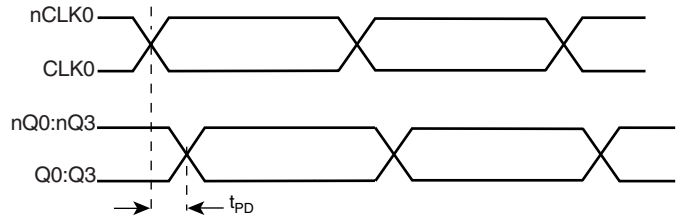
**DIFFERENTIAL INPUT LEVELS**



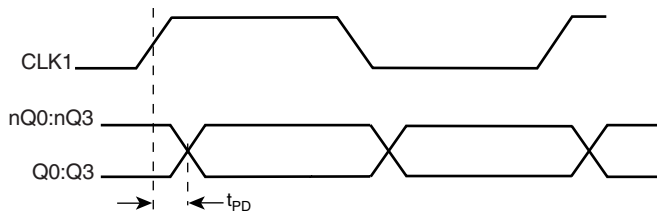
**PART-TO-PART SKEW**



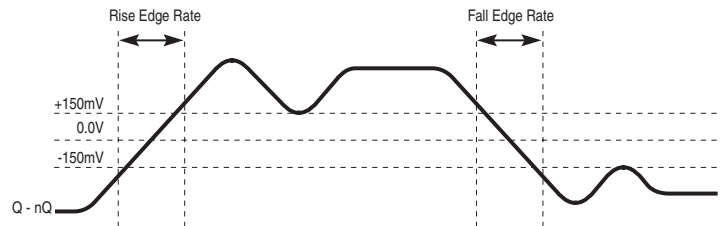
**OUTPUT SKEW (DIFFERENTIAL INPUT)**



**PROPAGATION DELAY (DIFFERENTIAL INPUTS)**

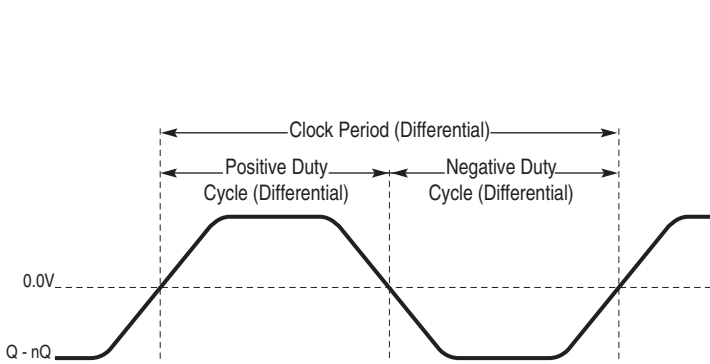


**PROPAGATION DELAY (LVCMOS INPUT)**

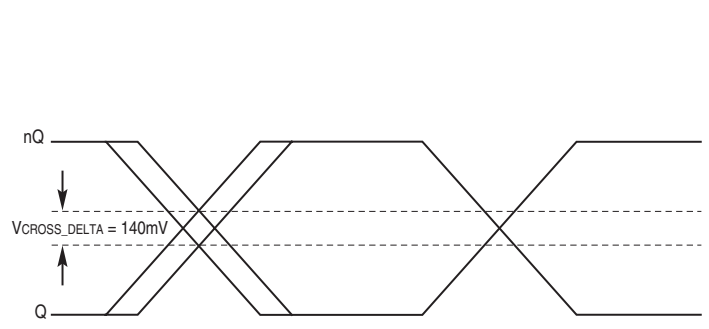


**DIFFERENTIAL MEASUREMENT POINTS FOR RISE/FALL TIME**

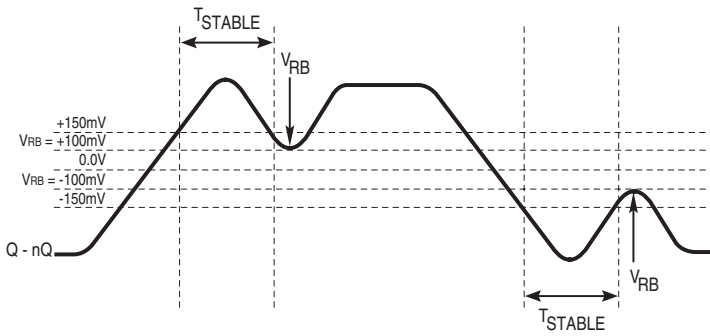
## PARAMETER MEASUREMENT INFORMATION, CONTINUED



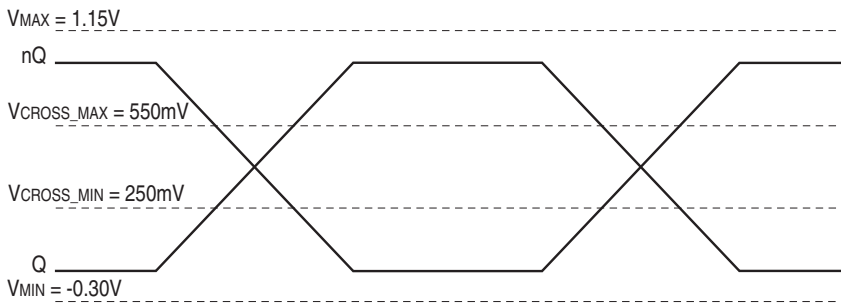
**DIFFERENTIAL MEASUREMENT POINTS FOR DUTY CYCLE/PERIOD**



**SE MEASUREMENT POINTS FOR DELTA CROSS POINT**



**DIFFERENTIAL MEASUREMENT POINTS FOR RINGBACK**



**SE MEASUREMENT POINTS FOR ABSOLUTE CROSS POINT/SWING**



## APPLICATION INFORMATION

### RECOMMENDATIONS FOR UNUSED INPUT AND OUTPUT PINS

#### INPUTS:

##### CLK INPUT

For applications not requiring the use of a clock input, it can be left floating. Though not required, but for additional protection, a 1k $\Omega$  resistor can be tied from the CLK input to ground.

##### CLK/nCLK INPUTS

For applications not requiring the use of the differential input, both CLK and nCLK can be left floating. Though not required, but for additional protection, a 1k $\Omega$  resistor can be tied from CLK to ground.

#### LVCMOS CONTROL PINS

All control pins have internal pull-ups or pull-downs; additional resistance is not required but can be added for additional protection. A 1k $\Omega$  resistor can be used.

#### OUTPUTS:

##### DIFFERENTIAL OUTPUTS

All unused differential outputs can be left floating. We recommend that there is no trace attached. Both sides of the differential output pair should either be left floating or terminated.

### WIRING THE DIFFERENTIAL INPUT TO ACCEPT SINGLE ENDED LEVELS

Figure 2 shows how the differential input can be wired to accept single ended levels. The reference voltage  $V_{BIAS} = V_{DD}/2$  is generated by the bias resistors R1, R2 and C1. This bias circuit should be located as close as possible to the input pin. The ratio

of R1 and R2 might need to be adjusted to position the  $V_{BIAS}$  in the center of the input voltage swing. For example, if the input clock swing is only 2.5V and  $V_{DD} = 3.3V$ ,  $V_{BIAS}$  should be 1.25V and  $R2/R1 = 0.609$ .

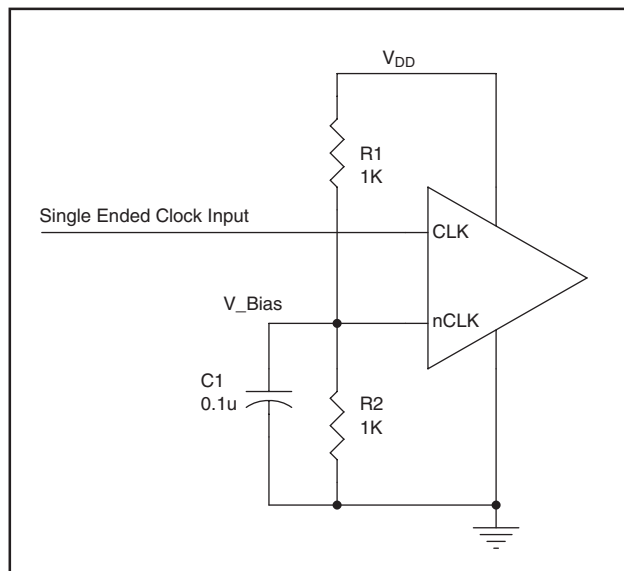
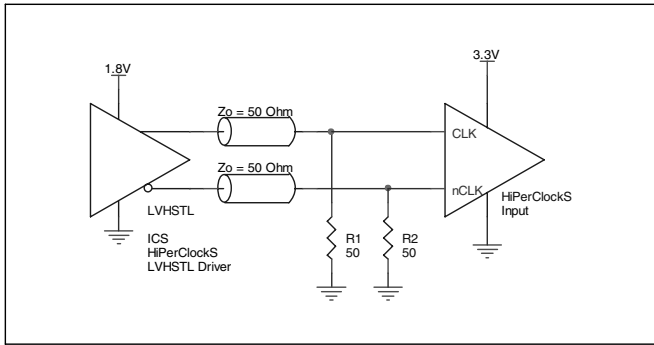


FIGURE 2. SINGLE ENDED SIGNAL DRIVING DIFFERENTIAL INPUT

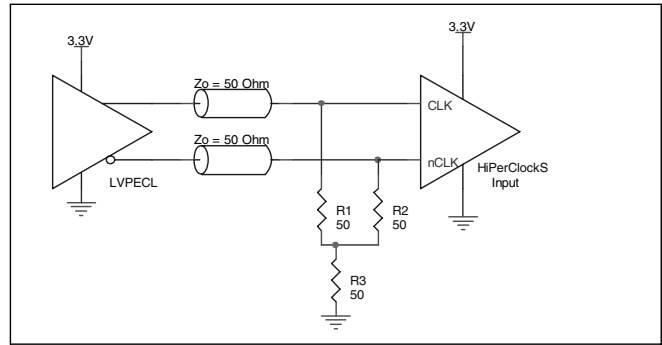
**DIFFERENTIAL CLOCK INPUT INTERFACE**

The CLK /nCLK accepts LVDS, LVPECL, LVHSTL, SSTL, HC SL and other differential signals. Both  $V_{SWING}$  and  $V_{OH}$  must meet the  $V_{PP}$  and  $V_{CMR}$  input requirements. Figures 3A to 3F show interface examples for the HiPerClockS CLK/nCLK input driven by the most common driver types. The input interfaces suggested here are

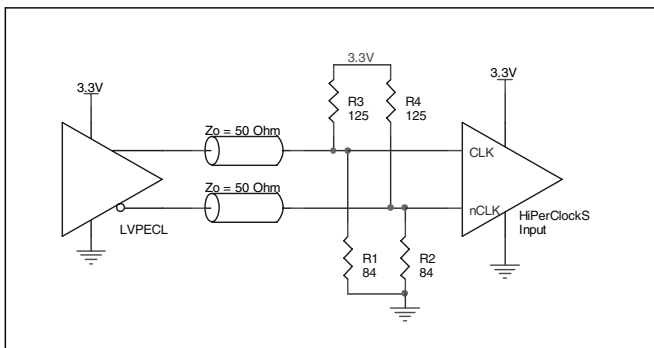
examples only. Please consult with the vendor of the driver component to confirm the driver termination requirements. For example in Figure 3A, the input termination applies for IDT HiPerClockS open emitter LVHSTL drivers. If you are using an LVHSTL driver from another vendor, use their termination



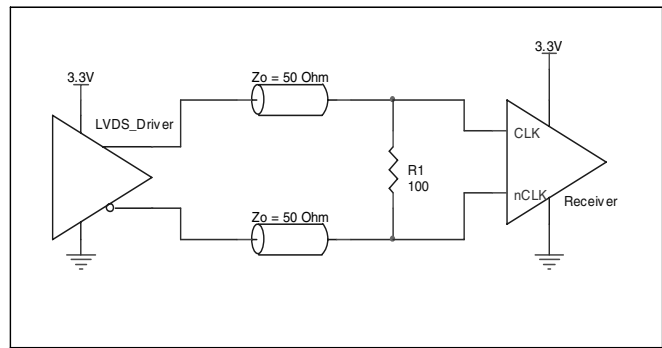
**FIGURE 3A. HiPerClockS CLK/nCLK INPUT DRIVEN BY AN IDT OPEN EMITTER HiPerClockS LVHSTL DRIVER**



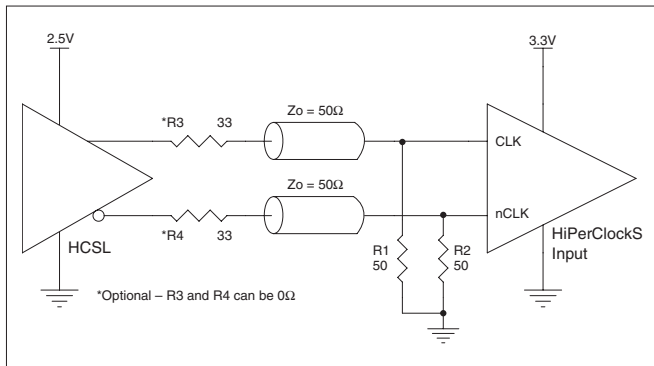
**FIGURE 3B. HiPerClockS CLK/nCLK INPUT DRIVEN BY A 3.3V LVPECL DRIVER**



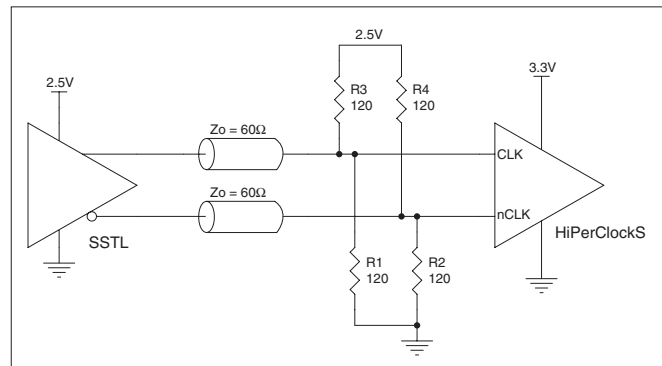
**FIGURE 3C. HiPerClockS CLK/nCLK INPUT DRIVEN BY A 3.3V LVPECL DRIVER**



**FIGURE 3D. HiPerClockS CLK/nCLK INPUT DRIVEN BY A 3.3V LVDS DRIVER**



**FIGURE 3E. HiPerClockS CLK/nCLK INPUT DRIVEN BY A 3.3V HC SL DRIVER**



**FIGURE 3F. HiPerClockS CLK/nCLK INPUT DRIVEN BY A 2.5V SSTL DRIVER**

**RECOMMENDED TERMINATION**

Figure 4A is the recommended termination for applications which require the receiver and driver to be on a separate PCB. All traces should be 50Ω impedance.

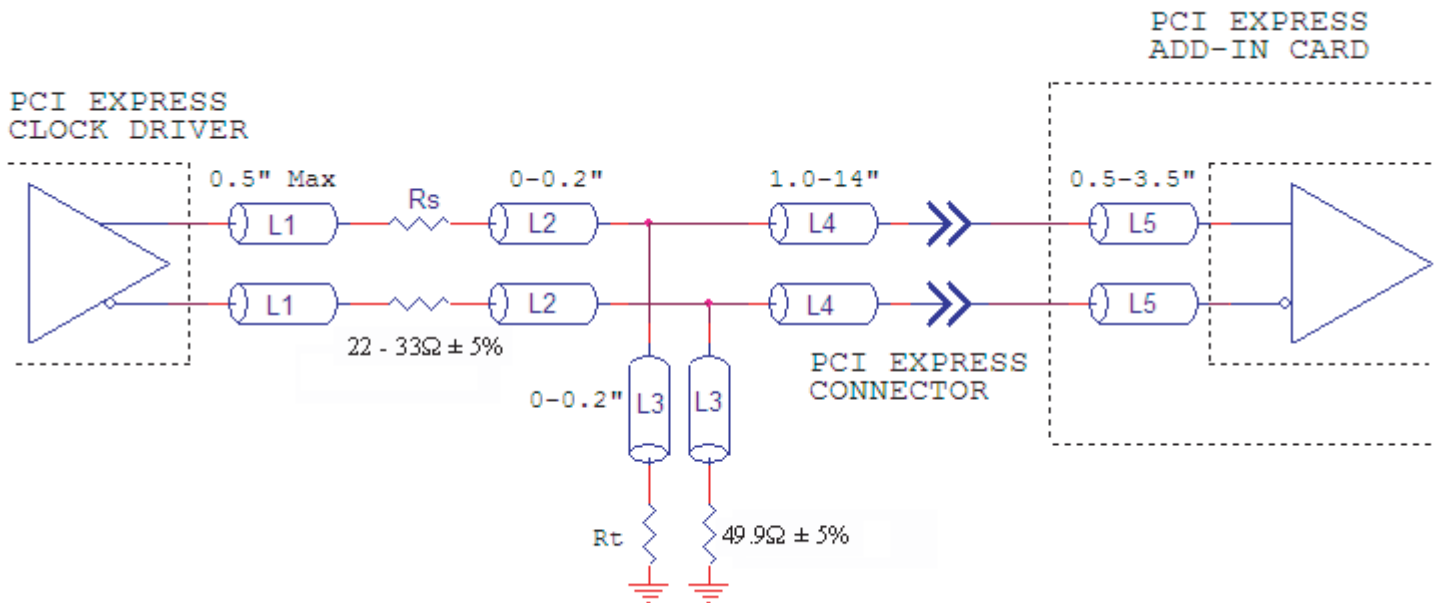
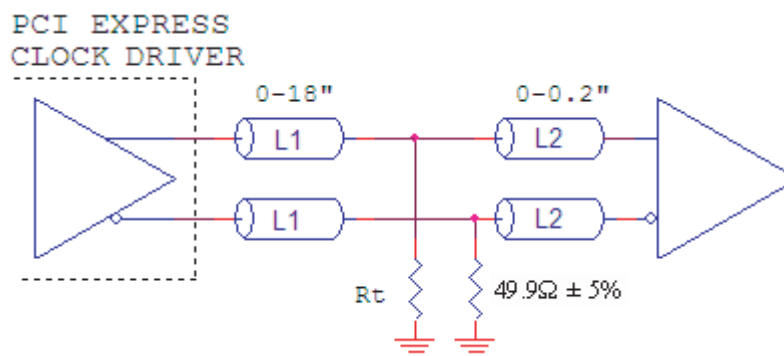
**FIGURE 4A. RECOMMENDED TERMINATION**

Figure 4B is the recommended termination for applications which require a point to point connection and contain the driver

and receiver on the same PCB. All traces should all be 50Ω impedance.

**FIGURE 4B. RECOMMENDED TERMINATION**

## POWER CONSIDERATIONS

This section provides information on power dissipation and junction temperature for the ICS85102I. Equations and example calculations are also provided.

### 1. Power Dissipation.

The total power dissipation for the ICS85102I is the sum of the core power plus the power dissipated in the load(s). The following is the power dissipation for  $V_{DD} = 3.3V + 10\% = 3.63V$ , which gives worst case results.

**NOTE:** Please refer to Section 3 for details on calculating power dissipated in the load.

- Power (core)<sub>MAX</sub> =  $V_{DD\_MAX} * I_{DD\_MAX} = 3.63V * 27mA = 98.01mW$
- Power (outputs)<sub>MAX</sub> = **47.3mW/Loaded Output pair**  
If all outputs are loaded, the total power is  $2 * 47.3mW = 94.6mW$

$$\text{Total Power}_{\_MAX} \text{ (3.63V, with all outputs switching)} = 98.01mW + 94.6mW = 192.61mW$$

### 2. Junction Temperature.

Junction temperature,  $T_j$ , is the temperature at the junction of the bond wire and bond pad and directly affects the reliability of the device. The maximum recommended junction temperature for HiPerClockS™ devices is 125°C.

The equation for  $T_j$  is as follows:  $T_j = \theta_{JA} * Pd\_total + T_A$

$T_j$  = Junction Temperature

$\theta_{JA}$  = Junction-to-Ambient Thermal Resistance

$Pd\_total$  = Total Device Power Dissipation (example calculation is in Section 1 above)

$T_A$  = Ambient Temperature

In order to calculate junction temperature, the appropriate junction-to-ambient thermal resistance  $\theta_{JA}$  must be used. Assuming no air flow and a multi-layer board, the appropriate value is 100.3°C/W per Table 6 below.

Therefore,  $T_j$  for an ambient temperature of 85°C with all outputs switching is:

$$85^\circ C + 0.193W * 100.3^\circ C/W = 104.4^\circ C. \text{ This is well below the limit of } 125^\circ C.$$

This calculation is only an example.  $T_j$  will obviously vary depending on the number of loaded outputs, supply voltage, air flow, and the type of board (single layer or multi-layer).

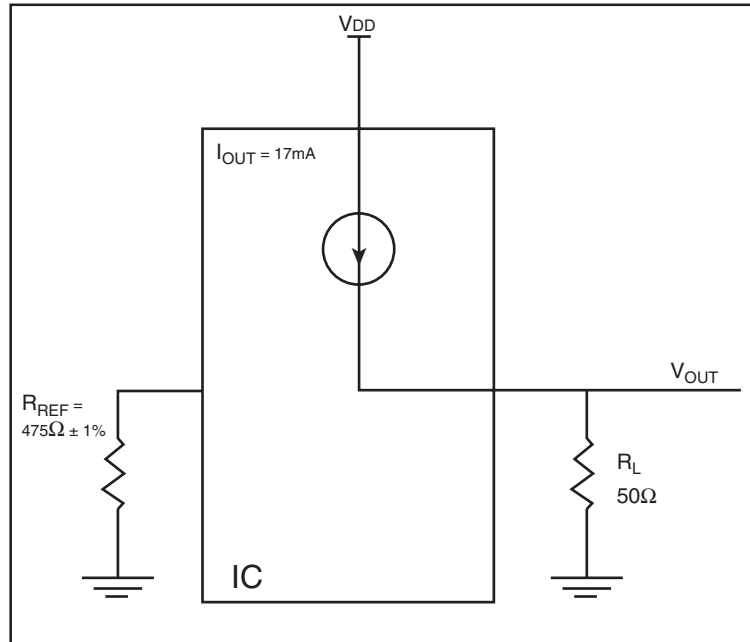
**TABLE 6. THERMAL RESISTANCE  $\theta_{JA}$  FOR 16-LEAD TSSOP, FORCED CONVECTION**

	$\theta_{JA}$ by Velocity (Meters per Second)		
	0	1	2.5
Multi-Layer PCB, JEDEC Standard Test Boards	100.3°C/W	96.0°C/W	93.9°C/W

### 3. Calculations and Equations.

The purpose of this section is to calculate power dissipation on the IC per HCSL output pair.

HCSL output driver circuit and termination are shown in *Figure 5*.



**FIGURE 5. HCSL DRIVER CIRCUIT AND TERMINATION**

HCSL is a current steering output which sources a maximum of 17mA of current per output. To calculate worst case on-chip power dissipation, use the following equations which assume a 50Ω load to ground.

The highest power dissipation occurs when  $V_{DD}$  is HIGH.

$$\begin{aligned} \text{Power} &= (V_{DD\_HIGH} - V_{OUT}) * I_{OUT}, \text{ since } V_{OUT} = I_{OUT} * R_L \\ &= (V_{DD\_HIGH} - I_{OUT} * R_L) * I_{OUT} \\ &= (3.63V - 17mA * 50\Omega) * 17mA \end{aligned}$$

Total Power Dissipation per output pair = **47.3mW**

## RELIABILITY INFORMATION

TABLE 7.  $\theta_{JA}$  VS. AIR FLOW TABLE FOR 16 LEAD TSSOP

$\theta_{JA}$ by Velocity (Meter per Second)			
	0	1	2.5
Multi-Layer PCB, JEDEC Standard Test Boards	100.3°C/W	96.0°C/W	93.9°C/W

### TRANSISTOR COUNT

The transistor count for ICS85102I is: 614

## PACKAGE OUTLINE AND DIMENSIONS

PACKAGE OUTLINE - G SUFFIX FOR 16 LEAD TSSOP

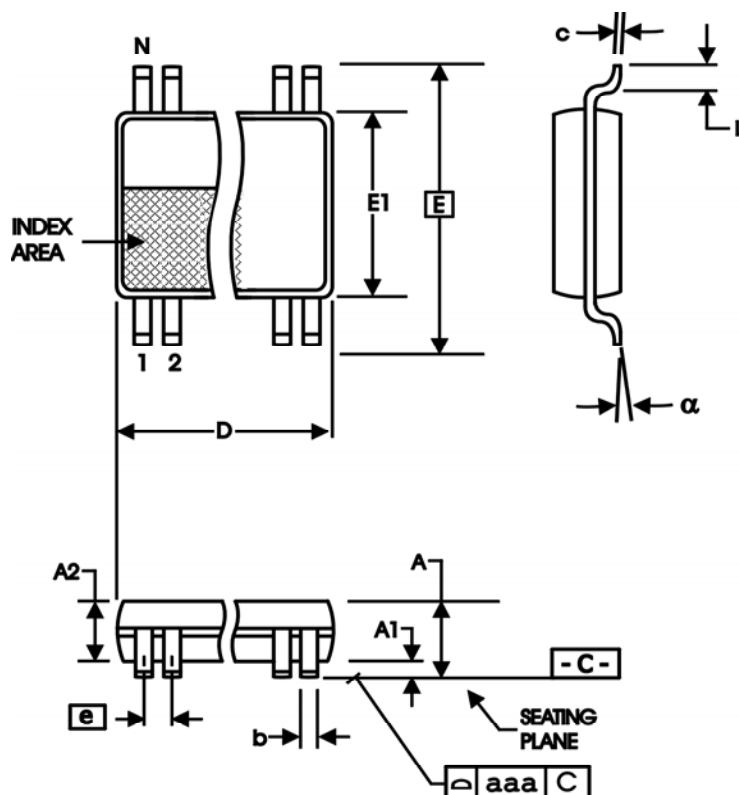


TABLE 8. PACKAGE DIMENSIONS

SYMBOL	Millimeters	
	Minimum	Maximum
N	16	
A	--	1.20
A1	0.05	0.15
A2	0.80	1.05
b	0.19	0.30
c	0.09	0.20
D	4.90	5.10
E	6.40 BASIC	
E1	4.30	4.50
e	0.65 BASIC	
L	0.45	0.75
$\alpha$	0°	8°
aaa	--	0.10

Reference Document: JEDEC Publication 95, MO-153

TABLE 9. ORDERING INFORMATION

Part/Order Number	Marking	Package	Shipping Packaging	Temperature
85102AGI	85102AGI	16 lead TSSOP	tube	-40°C to 85°C
85102AGIT	85102AGI	16 lead TSSOP	2500 tape & reel	-40°C to 85°C
85102AGILF	85102AIL	16 lead "Lead-Free" TSSOP	tube	-40°C to 85°C
85102AGILFT	85102AIL	16 lead "Lead-Free" TSSOP	2500 tape & reel	-40°C to 85°C

NOTE: Parts that are ordered with an "LF" suffix to the part number are the Pb-Free configuration and are RoHS compliant.

While the information presented herein has been checked for both accuracy and reliability, Integrated Device Technology, Incorporated (IDT) assumes no responsibility for either its use or for infringement of any patents or other rights of third parties, which would result from its use. No other circuits, patents, or licenses are implied. This product is intended for use in normal commercial and industrial applications. Any other applications such as those requiring high reliability or other extraordinary environmental requirements are not recommended without additional processing by IDT. IDT reserves the right to change any circuitry or specifications without notice. IDT does not authorize or warrant any IDT product for use in life support devices or critical medical instruments.

ICS85102I

LOW SKEW, 1-TO-2, DIFFERENTIAL/LVCMOS-TO-0.7V HCSL FANOUT BUFFER

**Innovate with IDT and accelerate your future networks. Contact:**

[www.IDT.com](http://www.IDT.com)

**For Sales**

800-345-7015 (inside USA)  
+408-284-8200 (outside USA)  
Fax: 408-284-2775  
[www.IDT.com/go/contactIDT](http://www.IDT.com/go/contactIDT)

**For Tech Support**

[netcom@idt.com](mailto:netcom@idt.com)  
+480-763-2056

**Corporate Headquarters**

Integrated Device Technology, Inc.  
6024 Silver Creek Valley Road  
San Jose, CA 95138  
United States  
800-345-7015 (inside USA)  
+408-284-8200 (outside USA)



© 2008 Integrated Device Technology, Inc. All rights reserved. Product specifications subject to change without notice. IDT, the IDT logo, ICS and HiPerClockS are trademarks of Integrated Device Technology, Inc. Accelerated Thinking is a service mark of Integrated Device Technology, Inc. All other brands, product names and marks are or may be trademarks or registered trademarks used to identify products or services of their respective owners.  
Printed in USA