# 31W TRIPLE OUTPUT STEP-DOWN SWITCHING REGULATORS

Туре	V <sub>i</sub>	V <sub>o</sub>	lo
GS-R51212S	15 to 40 V	+ 5,1 V	4,5 A
G5-R512125	15 to 40 V	± 12 V	0,35 A
GS-R51515S	15 to 40 V	+ 5,1 V	4,5 A
GS-K515155	15 to 40 V	± 15 V	0,3 A

#### **FEATURES**

- 5.1V/4.5A and ±12V/0.35A or ±15V/0.3A output voltages
- ±12 or ±15V externally adjustable
- High efficiency (81% typ.)
- Short-circuit protection
- Reset output
- Power Fail programmable input
- Inhibit/Enable control input
- Soft-start
- PCB or chassis mounting



#### **DESCRIPTION**

The GS-R51212S and GS-R51515S are versatile triple output, high current step-down switching regulators that provide +5.1V/4.5A output voltage and an isolated  $\pm 12V/0.35A$  or  $\pm 15V/0.3A$  dual output voltage.

They are ideal for microprocessor based boards because power the logic and the communication ports and have Reset output and Power Fail programmable input for the correct system start-up.

The Inhibit/Enable pin allows the ON/OFF logic function with TTL/CMOS compatible input signal. The auxiliary outputs ( $\pm 12V$  or  $\pm 15V$ ) are externally adjustable in a very wide range, i.e. from  $\pm 4.25V$  to  $\pm 12.45V$  on GS-R51212S and from  $\pm 4.50V$  to  $\pm 15.25V$  (typical values) on GS-R51515S.

#### **ABSOLUTE MAXIMUM RATINGS**

Symbol	Parameter	Value	Unit
Vi	DC Input Voltage	44	V
Irs	Reset Output Sink Current	20	mA

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# **ELECTRICAL CHARACTERISTICS** (T<sub>amb</sub> = 25°C unless otherwise specified)

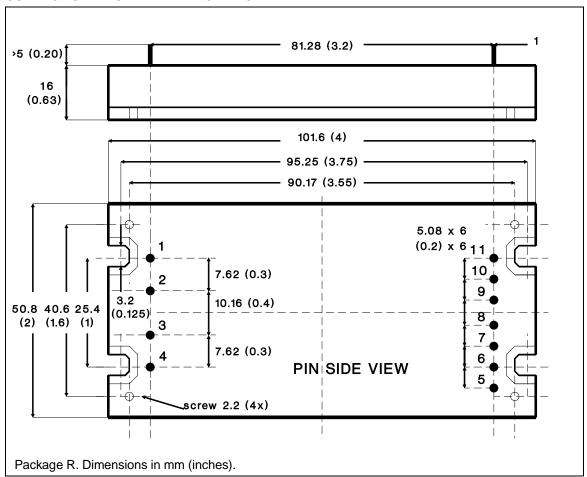
Symbol	Parameter	Test Conditions		Min	Тур	Max	Unit
Vi	Input Voltage GS-R51212S	V <sub>01</sub> = +5.1V V <sub>02</sub> = +12V V <sub>03</sub> = -12V	lo1 = 4.5A lo2 = 0.35A lo3 = -0.35A	15		40	V
Vi	Input Voltage GS-R51515S	V <sub>01</sub> = +5.1V V <sub>02</sub> = +15V V <sub>03</sub> = -15V	Io1 = 4.5A Io2 = 0.3A Io3 = -0.3A	15		40	V
lir	Input Reflected Current	Vi = 24V I <sub>01,2</sub> No external input	,3 = Full Load capacitor		0.5		Арр
lir	Input Reflected Current	Vi = 24V lo1,2 Ci (external) = 10	,3 = Full Load 0μF/50V		0.15		Арр
Vien	Enable Input Voltage	Vi = 15 to 40V		0		0.8	V
lien	Enable Input Current	Vi = 15 to 40V				<b>– 1</b>	mA
Viinh	Inhibit Input Voltage	V <sub>i</sub> = 15 to 40V		1.2		+Vi	V
V <sub>0</sub> 1	Output Voltage 1	V <sub>i</sub> = 15 to 40V lo2 = 0 to 0.35/0.3 lo3 = 0 to - 0.35/-		+5	+5.1	+5.2	V
V <sub>0</sub> 2	Output Voltage 2 GS-R51212S	Vi = 15 to 40V lo2 = 0 to 0.35A	$l_{01} = 0 \text{ to } 4.5A$ $l_{03} = 0 \text{ to } -0.35A$	+11.5	+12	+12.5	V
V <sub>o2</sub>	Output Voltage 2 GS-R51515S	Vi = 15 to 40V I <sub>02</sub> = 0 to 0.3A	$l_{01} = 0 \text{ to } 4.5A$ $l_{03} = 0 \text{ to } -0.3A$	+14.5	+15	+15.5	V
Vo3	Output Voltage 3 GS-R51212S	Vi = 15 to 40V lo2 = 0 to 0.35A	$l_{01} = 0 \text{ to } 4.5A$ $l_{03} = 0 \text{ to } -0.35A$	- 11.5	- 12	- 12.5	<b>V</b>
Vo3	Output Voltage 3 GS-R51515S	Vi = 15 to 40V lo2 = 0 to 0.3A	$l_{01} = 0 \text{ to } 4.5A$ $l_{03} = 0 \text{ to } -0.3A$	- 14.5	- 15	- 15.5	V
Vor1	Output Ripple Voltage 1	Vi = 24V	lo1 = 4.5A		30	50	mVpp
Vor2,3	Output Ripple Voltage 2,3	Vi = 24V	lo2,3 = 0.35/0.3A		50	100	mVpp
δVol1	Line Regulation 1	Vi = 15 to 40V lo2,3 = 0.35/0.3A	lo1 = 2.5A		0.5		mV/V
δV <sub>OL2,3</sub>	Line Regulation 2,3	V <sub>i</sub> = 15 to 40V lo2,3 = 0.35/0.3A	Io1 = 2.5A		1		mV/V
δV001	Load Regulation 1	Vi = 24V I <sub>02,3</sub> = 0.35/0.3A	lo1 = 0.5 to 4.5A		2		mV/A
δV <sub>OO2,3</sub>	Load Regulation 2,3	V <sub>i</sub> = 24V lo2,lo3 = 0.05 to 0	l <sub>01</sub> = 2.5A 0.35/0.3A		500		mV/A
lo1	Output Current 1	Vi = 15 to 40V l <sub>02,3</sub> = 0 to 0.35/0	V <sub>0</sub> 1 = 5.1V 0.3A	0		4.5	А
l <sub>02</sub>	Output Current 2* GS-R51212S	V <sub>i</sub> = 15 to 40V V <sub>0</sub> 2 = +12V	$l_{01} = 0 \text{ to } 4.5A$ $l_{03} = 0 \text{ to } -0.35A$	0		0.35	Α
lo2	Output Current 2* GS-R51515S	Vi = 15 to 40V Vo2 = +15V	$l_{01} = 0 \text{ to } 4.5A$ $l_{03} = 0 \text{ to } -0.3A$	0		0.3	Α
l <sub>02</sub>	Output Current 2* GS-R51212S	V <sub>i</sub> = 15 to 40V V <sub>0</sub> 2 = +12V	lo1 = 0 to 4.5A lo3 = 0A	0		0.7	Α
lo2	Output Current 2* GS-R51515S	Vi = 15 to 40V Vo2 = +15V	lo1 = 0 to 4.5A lo3 = 0A	0		0.6	А
lo3	Output Current 3* GS-R51212S	Vi = 15 to 40V Vo3 = - 12V	lo1 = 0 to 4.5A lo2 = 0 to 0.35A	0		- 0.35	А

# **ELECTRICAL CHARACTERISTICS** (T<sub>amb</sub> = 25°C unless otherwise specified) (cont'd)

Symbol	Parameter	Test C	onditions	Min	Тур	Max	Unit
103	Output Current 3* GS-R51515S	V <sub>i</sub> = 15 to 40V V <sub>0</sub> 3 = -15V	l <sub>01</sub> = 0 to 4.5A l <sub>02</sub> = 0 to 0.3A	0		- 0.3	Α
lo3	Output Current 3* GS-R51212S	Vi = 15 to 40V V <sub>0</sub> 3 = - 12V	lo1 = 0 to 4.5A l <sub>02</sub> = 0A	0		- 0.7	Α
103	Output Current 3* GS-R51515S	V <sub>i</sub> = 15 to 40V V <sub>0</sub> 3 = -15V	l <sub>01</sub> = 0 to 4.5A l <sub>02</sub> = 0A	0		- 0.6	Α
losck1	Output Current Limit 1	Vi = 15 to 40V	Overload		5.5		Α
losc1	Output Short-circuit Current 1	Vi = 15 to 40V			3		Α
losc2,3	Output Short-circuit Current 2,3	Vi = 15 to 40V			8.0		Α
tss	Soft-start time				10		ms
tdr	Reset Time Delay				100		ms
fs	Switching Frequency	$V_i = 15 \text{ to } 40V$ $I_{01} = 0.5 \text{ to } 4.5A$ $V_{02} = +12/+15V$ $I_{02}$ , $I_{03} = -0.05$	V <sub>0</sub> 3 = - 12/- 15V		100		kHz
η	Efficiency	Vi = 24V	I <sub>01,2,3</sub> = Full Load	78	81		%
Rth	Thermal Resistance				7.5		°C/W
Тсор	Operating Case Temperature Range			0		+85	°C
T <sub>stg</sub>	Storage Temperature Range			- 40		+105	°C

<sup>\*</sup> Note: when output current is less than 50mA, output ripple voltage increases due to discontinuous operation.

## **CONNECTION DIAGRAM AND MECHANICAL DATA**



## **PIN DESCRIPTION**

Pin	Function	Description
1	GND IN	Return for input voltage source. Internally connected to pin 7.
2	ĒN.	Inhibit/Enable control input. The converter is ON (ENABLE) when the voltage applied to this pin is lower than 0.8V. The converter is OFF (INHIBIT) when this pin is unconnected or the input voltage is in the range of 1.2 to Vi.
3	P.F.	Power Fail programmable input. If unconnected the Power Fail threshold voltage is 11V with 1V hysteresis (factory setting).
4	+ Vin	DC input voltage. Recommended maximum voltage is 40V.
5	RT	Reset output (active high). When the supply voltage +Vin and the regulated output voltage +Vo1 are in the correct range this signal is generated after a delay time of 100ms typical.
6	Vo + 5V	Regulated +5.1V output voltage.
7	GND 1	Return for output 1 current path. Internally connected to pin 1.
8	Vo + 12/15V	Regulated +12 or +15V output.
9	Vo – 12/15V	Regulated – 12 or – 15V output.
10	ADJ.	External adjustment for output voltages ±12 and ±15V.
11	GND Aux.	Return for ±12 and ±15V output current path.



#### **USER NOTES**

#### **Input Voltage**

The recommended operating maximum DC input voltage is 40V inclusive of the ripple voltage. The use of an external low ESR, high ripple current capacitor located as close the module as possible is recommended; suggested value is  $100\mu F/50V$ .

### Soft-start

To avoid heavy inrush current the output voltage rise time is typically 10ms in any condition of load.

#### **Power Fail-Reset Circuit**

The module include a voltage sensing circuit that may be used to generate a power-on/power-off reset signal for a microprocessor system.

The circuit sense the input supply voltage and the output generated voltage Vo1 (+5V) and will generate the required reset signal only when both the sensed voltages have reached the required value for correct system operation.

When both the supply voltage and the regulated voltage are in the correct range the output Reset signal is generated after a delay time  $t_{\text{DR}}$  of 100ms typical.

A latch assures that if a spike is present on the sensed voltage the delay time circuit discharges completely before initialization of a new reset cycle.

Reset output has internal pull-up resistor of 10kOhm connected to Vo +5V pin.

Maximum sink output current is 20mA at VRESET(sat) = 200mV.

Fig. 1 and fig. 2 show reset waveforms.

### **Power Fail Programmable Input**

This pin is internally connected via a divider to the +Vin pin for Power Fail function.

The factory setting is for a value of 11V with 1V hysteresis.

It is possible to program a different value of Power Fail threshold by connecting a resistor (Rpf) between pin 3 (Power Fail Input) and pin 1 (GND Input). The value of Rpf must be calculated according to the following formula:

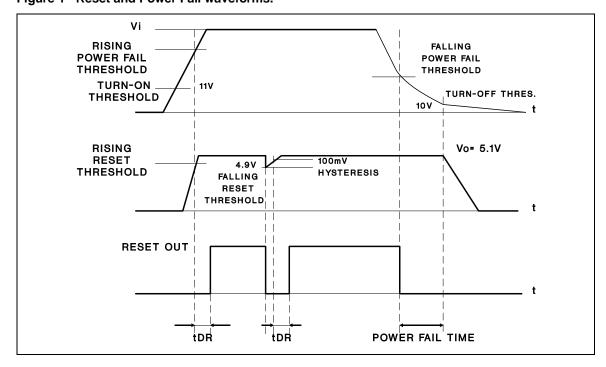
$$R_{pf} = \frac{5.1}{\frac{V_{pf} - 5.1}{34} - 0.191} = (k\Omega)$$

where Vpf is the desired value of Power Fail threshold voltage.

Exampe: Vpf = 24V (must not be lower than 12V):

$$R_{pf} = \frac{5.1}{\frac{24 - 5.1}{34} - 0.191} = 14k\Omega$$

Figure 1 - Reset and Power Fail waveforms.



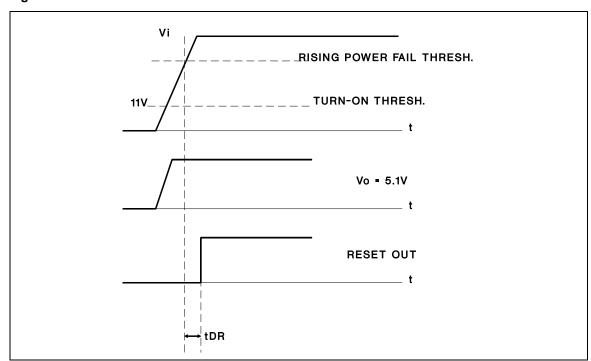


Figure 2 - Reset and Power Fail waveforms.

# **Auxiliary Outputs**

The auxiliary outputs ( $\pm 12V$  or  $\pm 15V$ ) are externally adjustable in symmetric way by connecting a resistor Ra between pin 10 (ADJ.) and pin 8 (Vo + 12/+15V), according to the following formula:

$$\begin{aligned} &\text{GS-R51212S} & &R_a = 32.66 \, \times \, \frac{V_o - 4.229}{12.485 - V_o} \\ &\text{GS-R5151SS} & &R_a = 38.66 \, \times \, \frac{V_o - 4.39}{15.252 - V_o} \end{aligned}$$

where  $V_0$  is the desired dual output voltage. Example:  $V_0 = \pm 5V$ .

 $R_a$  (GS-R51212S) = 3.36k $\Omega$ 

 $R_a (GS-R51515S) = 2.3k\Omega$ 

Example:  $V_0 = \pm 10V$ .

 $R_a (GS-R51212S) = 75.8k\Omega$  $R_a (GS-R51515S) = 41.3k\Omega$ 

**RESET Out** ENABLE ()-Rr ANALOG & DIGITAL 2 **BOARD** 5 6 +5.1V +Vin Cext GS-R GND 1 7 GND IN O 10 9 8 11 Rpf **POWER FAIL** GND AUX. +Vo +12/15V -Vo -12/15V ADJ.

Figure 3 - Typical Application.

## Inhibit/Enable Input

The Inhibit/Enable function allows the ON/OFF logic control of the module.

The converter is ON (Enable) when the voltage applied to pin 2 (EN.) and referred to pin 1 (GND IN) is lower than 0.8V (TTL, CMOS, open collector compatible level).

The converter is OFF (Inhibit) when pin 2 is unconnected or the voltage applied is in the range of 1.2V to +Vin. Maximum sinking current is 1mA.

#### **Module Protection**

The module is protected against occasional and permanent short-circuits of the output pins to ground, as well as against output current overload. The main output (+5.1V) uses a foldback current limiting; the output current decreases with increasing overload, reaching a minimum at short-circuit condition.

This solution minimizes internal power dissipation. The auxiliary outputs ( $\pm 12V$  or  $\pm 15V$ ) use a current limiting protection circuitry.

## Thermal characteristics

Sometimes the GS-R51212S and GS-R51515S require an external heat-sink depending on both operating temperature conditions and power.

Before entering into calculations details, some basic concepts will be explained to better understand the problem.

The thermal resistance between two points is represented by their temperature difference in front of a specified dissipated power, and it is expressed in Degree Centigrade per Watt (°C/W).

For the modules the thermal resistance case to ambient is 7.5°C/W. This means that an internal power dissipation of 1W will bring the case temperature at 7.5°C above the ambient temperature.

The maximum case temperature is 85°C.

Let's suppose to have a GS-R51515S that delivers the maximum output power of 31.4W at an ambient temperature of 40°C.



The dissipated power in this operating condition is about 7.4W (at typical efficiency of 81%), and the case temperature of the module will be:

$$T_{case} = T_{amb} + P_d \times R_{th} = 40 + 7.4 \times 7.5 = 95.5$$
 °C

This value exceeds the maximum allowed temperature and an external heat-sink must be added. To this purpose four holes (see mechanical drawing) are provided on the metal surface of the module. To calculate this heat-sink, let's first determine what the total thermal resistance should be:

$$R_{th} \, = \, \frac{T_{case(max)} \, - \, T_{amb}}{P_d} \, = \, \frac{85 - 40}{7.4} \, = \, 5.40 \, ^{\circ}\text{C/W}$$

This value is the resulting value of the parallel connection of GS-R thermal resistance and of the additional heatsink thermal resistance.

$$\frac{R_{th} (GSR) \times R_{th} (Heatsink)}{R_{th} (GSR) + R_{th} (Heatsink)} = 5.40^{\circ}C / W$$

To calculate the thermal resistance of the additional heat-sink the following equation may be used:

$$R_{th} \left( \text{Heatsink} \right) = \frac{5.40 \times R_{th} \left( \text{GSR} \right)}{R_{th \left( \text{GSR} \right)} - 5.40} = \frac{5.40 \times 7.5}{7.5 - 5.40} = 19.3 \ ^{\circ}\text{C} \ / \ \text{W}$$

In instead of or in addition to the external heatsink, a forced ventilation with an air speed of about 200 linear feet/minute can be used reducing the thermal resistance of the module at the specified value.

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