

Complies with Directive 2002/95/EC (RoHS)(Ph)

TXC100 Product Overview

TXC100 is a rugged single-chip ASK/FSK transmitter IC designed for operation in the 300-450 MHz frequency range. The highly integrated TXC100 includes a complete PLL frequency synthesizer plus an RF power amplifier that require very few external components. The TXC100 is very small in size and features high output power and low current consumption making it ideal for a wide range of short-range wireless applications in the industrial, automotive and consumer markets.

Key Features

Operating Frequency Range: 300-450 MHz

Modulation Types: ASK/FSK

Operation supply voltage: 2.1 - 3.6V

High Date rate:

OOK/ASK: 100 kbps FSK: 20 kbps

Low current consumption:

OOK/ASK mode: 7 mA typical FSK mode: 10 mA typical Low Stand by current: < 1 nA

Adjustable Output power: -10 to +10 dBm

Adjustable FSK Shift

Programmable Clock Output

Very Low external component count

Extended temperature range: -40 to +125 °C.

Small Package: 3x3 mm 16-pin TQFN package

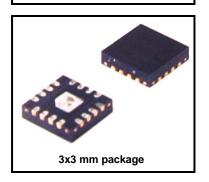
Standard 13 inch reel, 2500 pieces

Typical Applications

- Active RFID tags
- Automated Meter reading
- Wireless sensor nodes
- Home Automation
- Security systems
- Tire pressure monitoring
- Remote keyless entry
- Automobile Immobilizers
- Sports & Performance monitoring
- Wireless Toys
- Medical equipment
- Command & Control systems

TXC100

300-450 MHz **Transmitter**



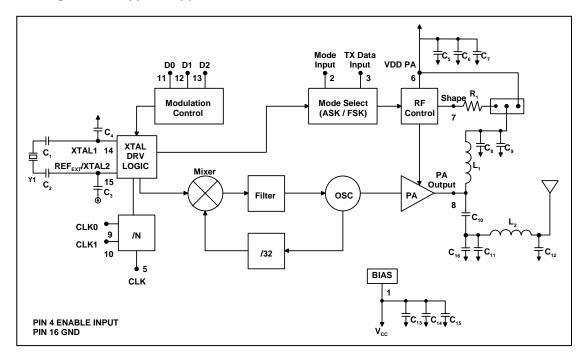
Electrical Characteristics

Characteristics	Sym	Min	Typical	Max	Units
Operating Frequency	fo	300		450	MHz
Modulation Types			OOK/ASK/ FSK		
ASK Data Rate				100	kbps
FSK Data Rate				20	kbps
Peak RF Output Power			+10		dBm
Standby Current				1	nA
Supply Voltage Range	Vdd	2.1		3.6	Vdc
Operating Temperature	Та	-40		+125	°C

Reference Crystal Parameters

Characteristics	Sym	Min	Typical	Max	Units
Crystal Frequency	fc		fo/32		MHz
Load Capacitance	CI			3	pF
Motional Capacitance	Cm	9		10	fF
Tolerance	Tol		±30		ppm

TXC100 Block Diagram and Typical Application Circuit



Component Values for Typical Application Circuit

Comp	315 MHz Band	433 MHz Band			
C1	100 pF	100 pF			
C2	100 pF	100 pF			
C3 ³	DNP	DNP			
C4 ³	DNP	DNP			
C5	1.0 uF	1.0 uF			
C6	0.01uF	0.01 uF			
C7	220 pF	220 pF			
C8	100 pF	100 pF			
C9	680 pF	680 pF			
C10 ¹	15 pF	3.3 pF			
C11 ¹	22 pF	5.6 pF			
C12 ¹	15 pF	12 pF			
C13	1.0 uF	1.0 uF			
C14	0.01uF	0.01uF			
C15	220 pF	220 pF			
C16 ¹	DNP	DNP			
L1 ¹²	27nH	22 nH			
L2 ¹²	22 nH	18 nH			
Y1	9.84375 MHz*	13.5600 MHz**			

¹Matched to 50 ohms

²Use wirewound inductors only

³Use for External Reference Input

DNP - Do Not Populate

^{*}Hong Kong Crystal P/N SSL9843750E03FAFR800

^{**}Hong Kong Crystal P/N SSM1356000E03FAFR800

Pin Configuration

BOTTOM VIEW 3x3mm Xtal2/REF_{IN} Xtal1 Dev2 GND 15 13 16 Dev1 12 1 $V_{\scriptscriptstyle DD}$ Dev0 $\mathsf{Mode}_{\mathsf{Sel}}$ 11 2 Clk1 10 3 Data_{IN} Clk0 9 4 Stdby CIR_{Out}

Pin Description

Pin	Name	Description
1	V _{DD}	V_{DD} is the supply voltage for the PLL and logic circuitry. Bypass this pin as close as possible with parallel 1.0 μ F, 0.01 μ F, 220 pF capacitors.
2	Mode _{Sel}	The Mode Select Pin allows the TXC100 to be set in either OOK/ASK or FSK mode: Logic Low: OOK/ASK mode Logic High: FSK mode In OOK/ASK mode, data input to the Data _{IN} Pin (3) gates the internal power amplifier. A logic high turns the power amplifier on. A logic low turns off the power amplifier. In FSK mode, data input to the Data _{IN} Pin shifts the carrier frequency by the amount programmed through the DEV[20] pins (11,12,13). A logic low performs no shift. The frequency of a logic low input in FSK mode is the same frequency as a logic high in OOK/ASK mode. The FSK deviation is achieved by pulling the crystal frequency. See <i>Crystal Reference</i> section (pin 15) for more details. The maximum FSK deviation for the 315 MHz band and 433 MHz band is approximately 55 kHz and 80 kHz, respectively.
3	Data _{IN}	The Data Input Pin turns the Power Amplifier on/off in OOK/ASK mode, or high/low frequency in FSK mode: Low (OOK/ASK mode): Power Amplifier off High (OOK/ASK mode): Power Amplifier on Low (FSK mode): Low frequency High (FSK mode): High frequency
4	Stdby	The Standby Input Pin selects active or low power shutdown/standby mode: If this pin unconnected or logic low, the TXC100 is placed in low current standby mode. If this pin is logic high, the TXC100 is in active mode and ready to transmit Note: Lowest current consumption is achieved when all configuration pins are logic low. When this pin sets the device in low power shutdown, the device draws nominally 0.2 nA. When the device is brought out of standby with a logic high input, it is ready for operation within 200 µs. This pin has an internal pull-down resistor so it can be pulled low or left unconnected. The 200 µs turn-on time is due to crystal start-up. An optimally matched crystal will minimize this turn-on time. See the Crystal Reference section (Pin 15) for details on crystal load matching.

		<u> </u>
		The clock output is a scaled and buffered version of the crystal frequency, which may be used to drive external logic or a microprocessor. The frequency is programmable through Pins 9 (Clk0) and 10(Clk1) as below:
5	Clk _{Out}	Clk0 Clk1 Clk _{Out} 0 0 0
3	CinQut	1 0 fc/4 0 1 fc/8
		1 1 fc/16
		fc = Crystal Frequency This girls a payor supply welters a course for the transmitter payor amplifier. Purpose this girls a close as possible with a 0.04 µF and 220.
6	V_{DDPA}	This pin is a power supply voltage source for the transmitter power amplifier. Bypass this pin as close as possible with a 0.01 µF and 220 pF capacitor. ES _{Out} is the alternate power amplifier voltage source as discussed below.
7	ES _{Out}	This pin can be used to supply voltage to the power amplifier instead of V _{DDPA} , allowing the on/off rise and fall times of the power amplifier to be adjusted in ASK mode. Reducing the rise and fall times reduces the spectral bandwidth of the modulated output signal, and also reduces average transmitter power. Rise and fall times are adjusted by placing a resistor in the range of 1K to 5K in series with this output, as close as possible to the TXC100 IC to minimize circuit parasitics. The power amplifier side of the resistor should be bypassed with 680 pF and 220 pF capacitors in parallel, as close to the resistor as possible.
8	PA _{Out}	Power Amplifier - the power amplifier is an open-drain, Class C amplifier designed for a load impedance at PA _{OUT} (pin 8) of about 250 ohms. The power amplifier requires a DC path to the supply voltage through a series inductor, which can be part of the output matching network. A 50 ohm antenna matching network is shown in the <i>Typical Application Circuit</i> section. The matching network also suppresses carrier harmonics to aid in compliance testing.
10, 9	Clk[10]	See description for Pin 5
13, 12, 11	FreqDev[20]	The Frequency Deviation Pins set the amount of deviation between data logic states in FSK mode. Frequency deviation is programmable through pins 11, 12, 13 as shown in the table below: DEV
14	Xtal1	External Crystal Input 1 presents a capacitance of 3 pF to GND in ASK and FSK (Data _{IN} logic low) modes. Circuit parasitics add to this package capacitance, presenting a total load of about 4.5 pF.
15	Xtal2/REF _{IN}	External Crystal Input 2 presents a capacitance of 3 pF to GND in ASK and FSK(Data _{IN} logic low) mode. Circuit parasitics add to this package capacitance, presenting a total load of about 4.5 pF. The External Ref Input allows an external frequency source to be used to obtain the desired transmit frequency. In this case, the Xtal1 input must be bypassed with a 0.01µF capacitor, and a 0.01µF series capacitance should be added into External Reference input. Crystal Reference - the Xtal1 and Xtal2 inputs are designed to present a 3 pF load to GND to each reference crystal connection. Including PCB parasitic capacitances, this increases to about 4.5 pFat each connection. In ASK mode, the full 3 pF load is applied to the crystal allowing it to oscillate at the desired frequency. In FSK mode, a portion of the 3 pF load is removed in response to a logic high being applied to the Data _{IN} (pin 2) as set by the frequency deviation pins DEV[02] (11,12,13). To achieve larger frequency deviations, use a crystal with large motional capacitance or reduce PCB parasitic capacitance to the extent possible.
		Connect to system ground.
16	GND	Note: The ground pad in the middle of the package is the power amplifier ground. It must be connected to system ground thru a low inductance path.

Absolute Maximum Ratings

To	Min	Max	
To			
•	-40	+125	°C
T _J	-40	+150	°C
Ts	-60	+150	°C
Vs	-0.3	+4	V
	-0.3	Vdd + 0.3	V
	T _S	T _s -60 V _s -0.3	T _S -60 +150 V _S -0.3 +4 -0.3 Vdd + 0.3

Note: Maximum ratings must not be exceeded under any circumstances and can cause permanent damage to the IC

DC Electrical Characteristic

(Typical values taken at V_{DD} = +3.0 V, T_A = +25 °C, unless otherwise noted)

Characteristic	Sym	Notes	L	imit Values	;	Unit	Test Condition	•	
Cital acteristic	Sylli	Notes	min	typ	max	Offic	rest conditions		
Supply Voltage	V_{DD}		2.1		3.6	V			
Current Consumption									
				0.2	1		T _A = +25 °C		
Standby	I _{STDBY}			120	300	nA	T _A = +85 °C		
				700	1600		T _A = +125 °C		
		1,4		2.9	4.3		PA off, Data=0V (ASK)		
Supply	I _{DD}			7	10.7		50% duty cycle (ASK)	315 MHz Band	
				10.5	17.1	mA	Data = V _{DD} (FSK and ASK)		
				3.3	4.8		PA off, Data=0V (ASK)		
				7.3	11.4		50% duty cycle (ASK)	433 MHz Band	
				10	18.1		Data = V _{DD} (FSK and ASK)		
Digital Inputs									
Data Input Low	V _{IL}				0.25	V			
Data Input High	V _{IH}		V _{DD} -0.25			V			
Max Input Current	I _I			15.5	20	μΑ			
Digital Outputs									
Output Voltage Low	V _{OL}				0.25	V	Clkout, Load = 10 pF		
Output Voltage High	V _{OH}		V _{DD} -0.25			V	Clkout, Load = 10 pF		

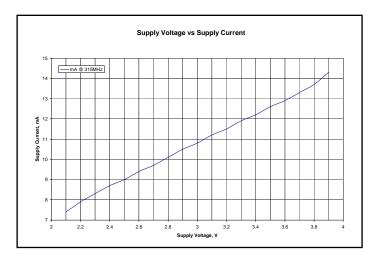
AC Electrical Characteristic

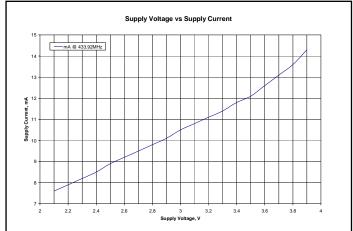
(Typical values taken at V_{DD} = +3.0 V, T_A = +25 °C, unless otherwise noted)

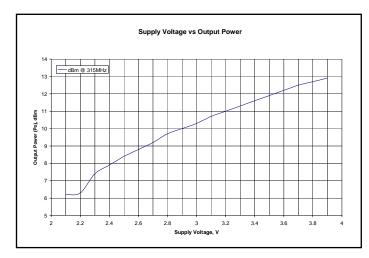
D	0	Notes	Limit Values			l lm!t	Total Constitution of		
Parameter	Sym		min	typ	max	Unit	Test Conditions		
			•						
PLL Performance									
VCO Gain	K _{vco}			280		MHz/V			
				-75			315 MHz Band	Freq Offset = 100 kHz	
Phase Noise				-74		dBc/Hz	433 MHz Band	rieq Oliset = 100 KHZ	
Filase Noise				-98		UDC/11Z	315 MHz Band	Freq Offset = 1 MHz	
				-98			433 MHz Band	rieq Oliset = Tiviliz	
Loop BW	BW			300		kHz			
Reference Spur				-40		dBc			
2nd Harmonic				-56		dBc	315 MHz Band		
ZIIU HaiiiiOiiiC				-52		UDC	433 MHz Band		
3rd Harmonic				-56		dPo	315 MHz Band		
Sid Haimonic				-50		dBc	433 MHz Band		
Crystal									
Frequency Range	f _{REF}			f _{RF} /32		MHz	fundamental mode, AT		
Tolerance		3		50		ppm			
Internal Load Capacitance		2		3		pF			
Clock Output Frequency	CLK _{OUT}			F _{XTAL} /N		MHz	Determined by CLK1 and CLK2		
System Characteristics									
Frequency Range			300		450	MHz			
	4			12.2	.2 16.1	dBm	$TA = -40 ^{\circ}C, V_{DD} = +3.6 V$	into 50Ω matched load	
Output Power		4	6.1	10	12.4		TA = +25 °C, VDD = +3.0 V		
			2.7	5.3			TA = +125 °C, VDD = +2.1 V	load	
Start-up time	t _{ON}	5		160		μs	STDBY to	TX	
Rise Time	tr	5		300		ns			
Max Data Rate		5		20		libna	FSK (50% Dut	y Cycle)	
Max Dala Rale		5		100		kbps	ASK (50% Duty Cycle)		
Fragues av Davieties (FCV)				55		kHz	315 MHz Band	DEV/(2, 0), 444	
Frequency Deviation (FSK)				80		KHZ	433 MHz Band	DEV[20]=111	
				35			315 MHz Band	CM	
Transmit Efficiency		,		31		0/	433 MHz Band	CW	
$\eta = P_{OUT}/(V_{DD}^*I_{DD})$		4		27		%	315 MHz Band	F00/ data and	
				25			433 MHz Band	50% duty cycle	
Power ON/OFF Ratio				-77		dB	ASK Mod	de	
Frequency Stability vs. VDD	Δdf_{VDD}			4		kHz			
Frequency Stability vs. Temp	Δdf_{TA}			TBD		kHz	-40 °C to +8	35 ℃	

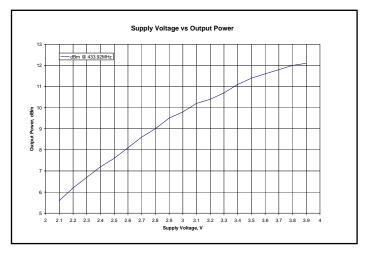
- 1. 10 kHz, 50% duty cycle
 2. Dependent on PCB parasitic trace capacitance and crystal parameters.
- 3. Dependent on crystal parameters.
- 4. Transmit Efficiency, RF Output Power, and Supply Current are heavily dependent on proper output matching and PCB layout. 5. No Envelope Shaping.

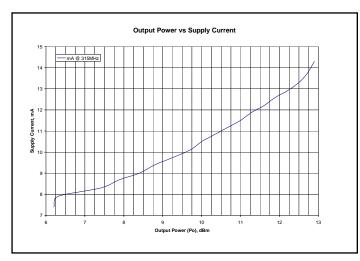
Typical Operating Characteristics

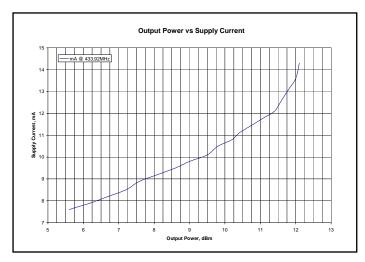


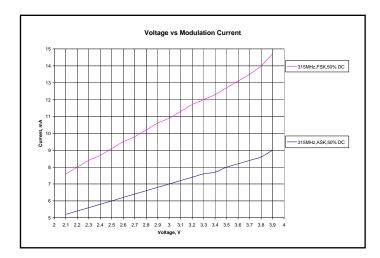


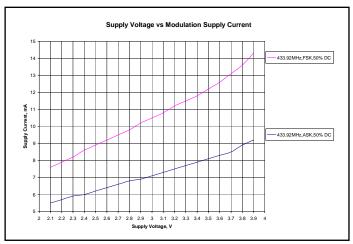


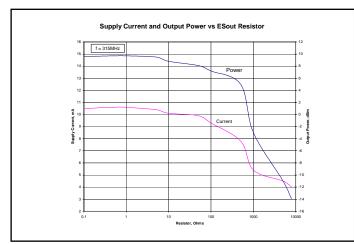


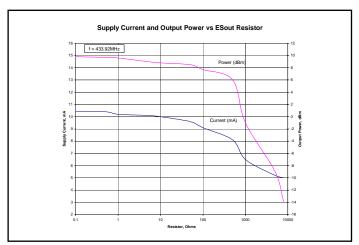












Theory of Operation

Introduction

The TXC100 is a crystal-referenced transmitter designed to operate in the 315/433 MHz frequency spectrum. The carrier and crystal reference relation is given by:

$$f_C = f_{XTAL} * 32$$

It is capable of supporting ASK and FSK data transmissions at 100 and 20 kbps, respectively. The output power is adjustable from -10 dBm to +10 dBm thru a resistor at the ES_{OUT} (pin 7). The FSK frequency deviation is programmable with up to eight different deviation values. The IC also provides a scaled and buffered clock output of the reference crystal for use by an external processor. The clock output frequency scaling programmable.

Frequency Synthesizer

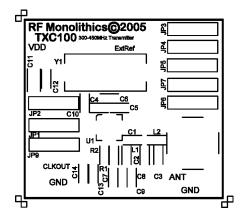
The frequency synthesizer is a phase locked loop (PLL) circuit with a loop bandwidth of 300 kHz. The PLL contains a phase detector, charge pump, VCO, integrated loop filter, ÷32 clock divider, and crystal oscillator drive circuit. The internal PLL is self contained and requires no external components for filtering or dividing. Only a reference crystal is needed.

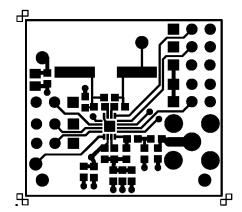
50Ω Output Matching

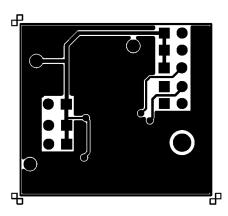
When properly matched, the TXC100 can output up to \pm 12 dBm into a 50 Ω load. The output is an open-drain configuration which requires a pull-up inductor for biasing. The pull-up inductance serves to provide biasing for the power amplifier and is a high frequency choke to reduce unwanted coupling back into the power supply. Maximum power transfer occurs when the output is closely matched to 250 Ω . For best performance use wirewound inductors instead of chip inductors. Wirewound inductors provide lower insertion loss as opposed to chip inductors. See *Typical Application Circuit* (section IV) for topology and matching component values.

PCB Layout Considerations

PCB layout is critical to proper and consistent operation. Always use controlled impedance lines from the PA_{OUT} (pin 8). For a .062" thick FR4 board, a 50Ω impedance line is approximately 0.110" wide. Component spacing is critical as well. Keep all output matching components as close together as possible to minimize stray inductance and capacitance that can detune the matching network. Keep top side ground planes at least a board thickness away from the signal output leading to the antenna or RF connector.





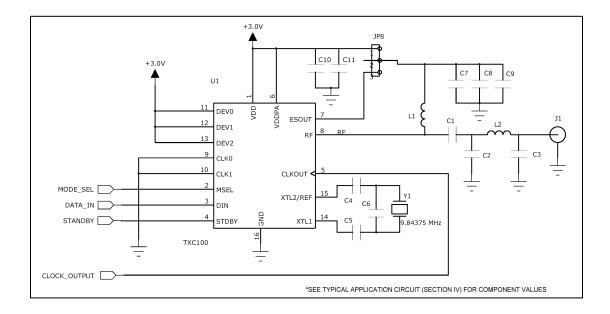


Antenna Layout Considerations

Most compact wireless designs require a small antenna. Loop antennas are often used as they can be designed into small spaces. The design of loop antennas is a fairly lengthy and detailed discussion which is beyond the scope of this datasheet. The object here is to provide a "rule of thumb" approach to achieve an appropriate starting point. Empirical data will provide the best path to take.

The circumference of the antenna should be less than $\lambda/4$ so that the antenna appears inductive. For this, a series matching capacitor is used to tune out the inductance of the antenna, since the antenna appears inductive. The capacitor may be located at the feed point of the antenna or at the "grounded" end. The capacitor may be a variable type or several fixed values may be attempted until an optimal match is reached. The use of a good network analyzer is essential for proper matching and maximum power transfer. For additional information on antenna design see the Application Notes section of our website: http://www.rfm.com/corp/apnotes.htm.

Typical Test Circuit



Package Dimensions - 3x3 mm 16-pin TQFN Package

(all values in mm)

