

## LM3492/LM3492Q

# Two-Channel Individual Dimmable LED Driver with Boost Converter and Fast Current Regulator

### General Description

The LM3492 integrates a boost converter and a two-channel current regulator to implement a high efficient and cost effective LED driver for driving two individually dimmable LED strings with a maximum power of 15W and an output voltage of up to 65V. The boost converter employs a proprietary Projected-On-Time control method to give a fast transient response with no compensation required, and a nearly constant switching frequency programmable from 200 kHz to 1 MHz. The application circuit is stable with ceramic capacitors and produces no audible noise on dimming. The programmable peak current limit and soft-start features reduce current surges at startup, and an integrated 190 mΩ, 3.9A N-Channel MOSFET switch minimizes the solution size. The fast slew rate current regulator allows high frequency and narrow pulse width dimming signals to achieve a very high contrast ratio of 1000:1 at a dimming frequency of more than 3 kHz. The LED current is programmable from 50 mA to 200 mA by a single resistor.

To maximize the efficiency, Dynamic Headroom Control (DHC) automatically adjusts the output voltage to a minimum. DHC also facilitates a single BOM for different number of LED in a string, which is required for backlight panels of different size, thereby reducing overall development time and cost. The LM3492 comes with a versatile COMM pin which serves as a bi-directional I/O pin interfacing with an external MCU for the following functions: power-good, over-temperature, IOUT over- and under-voltage indications, switching frequency tuning, and channel 1 disabling. Other supervisory functions of the LM3492 include precise enable, VCC under-voltage lock-out, current regulator over-power protection, and thermal shutdown protection. The LM3492 is available in the thermally enhanced eTSSOP-20 package.

### Features

#### Boost Converter:

- LM3492Q is an Automotive Grade Product that is AEC Q100 grade 1 qualified
- Very wide input voltage ranged from 4.5V-65V
- Programmable soft-start
- No loop compensation required
- Stable with ceramic and other low ESR capacitors with no audible noise
- Nearly constant switching frequency programmable from 200 kHz to 1 MHz

#### Current Regulators:

- Programmable LED current from 50 mA to 200 mA
- 1000:1 contrast ratio at a dimming frequency of more than 3 kHz, minimum LED current pulse width is 300 ns
- Two individual dimmable LED strings up to 65V, total 15W (typically 28 LEDs @ 150 mA)
- Dynamic Headroom Control maximizes efficiency
- Over-Power protection
- ±3% current accuracy

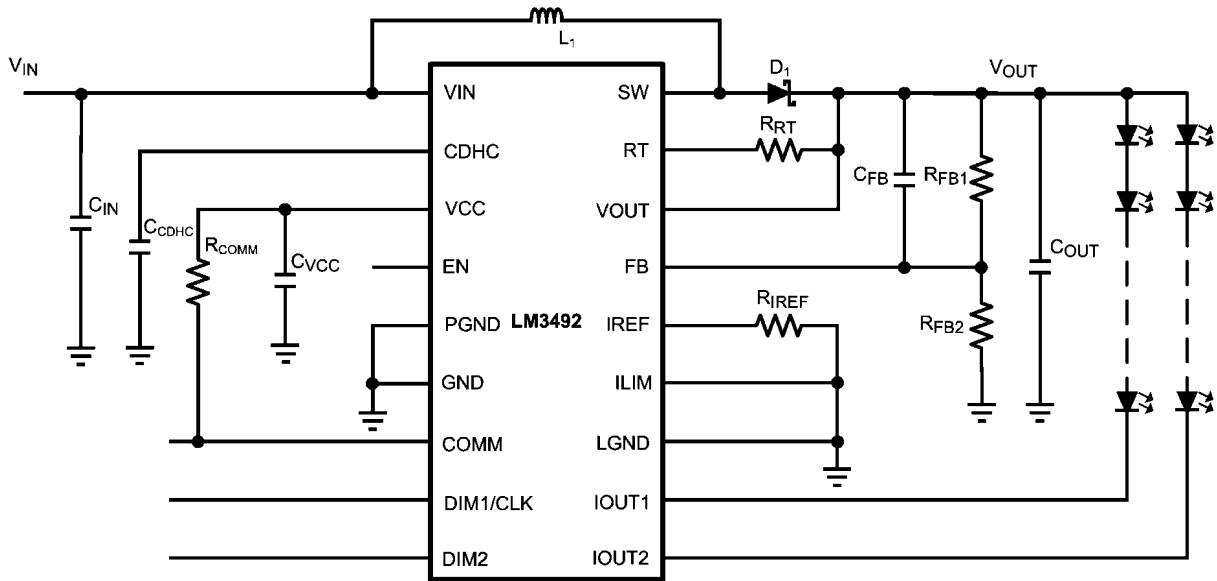
#### Supervisory Functions:

- Precision enable
- COMM I/O pin for diagnostic and commands
- Thermal shutdown protection
- Thermally enhanced eTSSOP-20 package

### Applications

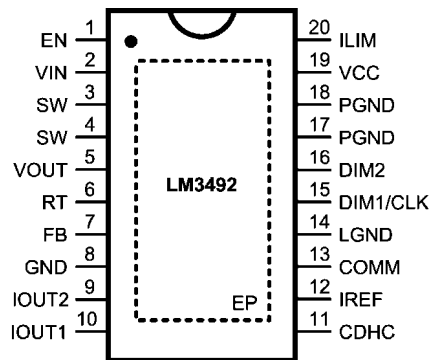
- Ultra-high contrast ratio 6.5"-10" LCD display backlight up to 28 LEDs
- Automotive or marine GPS display

## Typical Application



30116101

## Connection Diagram



30116102

Top View  
20-Lead Plastic eTSSOP (MXA20A)

## Ordering Information

Order Number	Package Type	NSC Package Drawing	Supplied As	Feature
LM3492MH	Exposed Pad TSSOP-20	MXA20A	73 Units per Anti-Static Tube	AEC-Q100 Grade 1 qualified. Automotive Grade Production Flow*
LM3492MHX			2500 Units on Tape and Reel	
LM3492QMH			73 Units per Anti-Static Tube	
LM3492QMHX			2500 Units on Tape and Reel	

\*Automotive Grade (Q) product incorporates enhanced manufacturing and support processes for the automotive market, including defect detection methodologies. Reliability qualification is compliant with the requirements and temperature grades defined in the AEC-Q100 standard. Automotive grade products are identified with the letter Q. For more information go to <http://www.national.com/automotive>.

## Pin Descriptions

Pin	Name	Description	Application Information
1	EN	Enable	Internally pull-up. Connect to a voltage higher than 1.63V to provide precision enable for the device.
2	VIN	Input Supply Voltage	Supply pin to the device. Input range is 4.5V to 65V.
3, 4	SW	Switch Node	Internally connected to the drain of the integrated MOSFET.
5	VOUT	Output Voltage Sense	Sense the output voltage for nearly constant switching frequency control.
6	RT	Frequency Control	An external resistor from the VOUT pin to this pin sets the switching frequency.
7	FB	Output Voltage Feedback	The output voltage is connected to this pin through a feedback resistor divider for output voltage regulation. The voltage of this pin is from 1.05V to 2.5V.
8	GND	Analog Ground	Signal Ground
9	IOUT2	Current Regulator Input of Channel 2	Input of the current regulator of channel 2. The regulated current is programmable (refer to the IREF pin).
10	IOUT1	Current Regulator Input of Channel 1	Input of the current regulator of channel 1. The regulated current is programmable (refer to the IREF pin).
11	CDHC	Dynamic Headroom Control	An external capacitor connected to this pin sets the DHC sensitivity. At startup, a 120 $\mu$ A internal current source charges an external capacitor to provide a soft-start function.
12	IREF	Current Setting of the Current Regulator	An external resistor connected from this pin to ground programs the regulated current of the current regulator of channels 1 and 2.
13	COMM	Bi-directional Logic Communication	This pin is open drain for various indications (power-good, over-temperature, IOUT over- and under-voltage) and command sending (switching frequency tuning and channel 1 disabling).
14	LGND	Ground of the Current Regulator	Current regulator ground. Must be connected to the GND pin for normal operation. The LGND and GND pins are not internally connected.
15	DIM1/CLK	Dimming Control of Channel 1	Control the on/off of the current regulator of channel 1. This pin is internally pulled low by a 5 $\mu$ A current. This pin also serves as a clock signal for latching input/output data of the COMM pin.
16	DIM2	Dimming Control of Channel 2	Control the on/off of the current regulator of channel 2. This pin is internally pulled low by a 5 $\mu$ A current.
17, 18	PGND	Power Ground	Integrated MOSFET ground. Must be connected to the GND pin for normal operation. The PGND and GND pins are not internally connected.
19	VCC	LDO Regulator Output	Nominally regulated to 5.5V. Connect a capacitor of larger than 0.47 $\mu$ F between the VCC and GND pins.
20	ILIM	Peak Current Limit Adjust	Connect an external resistor from the ILIM pin to the VCC pin reduces peak current limit. Connect the ILIM pin to the ground to obtain the maximum current limit.
DAP	DAP	Exposed Pad	Thermal connection pad. Connect to a ground plane.

**Absolute Maximum Ratings** (Note 1)

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/Distributors for availability and specifications.

VIN, RT, VOUT to GND	-0.3V to 67V
SW to GND	-0.3V to 67V
SW to GND (Transient)	-2V (<100 ns)
ILIM to GND	-0.3V to 0.3V
FB to GND	-0.3V to 5V
COMM, DIM1, DIM2, to GND	-0.3V to 6V

## ESD Rating (Note 2)

Human Body Model	±2kV
Storage Temperature Range	-65°C to +150°C
Junction Temperature (T <sub>J</sub> )	150°C

**Operating Ratings** (Note 1)

Supply Voltage (VIN)	4.5V to 65V
Operation Temperature Range (T <sub>A</sub> )	-40°C to +125°C
Thermal Resistance (θ <sub>JA</sub> ) (Note 3)	32.7°C/W

**Electrical Characteristics**

Specification with standard type are for T<sub>A</sub> = T<sub>J</sub> = +25°C only; limits in boldface type apply over the full Operating Junction Temperature (T<sub>J</sub>) range. Minimum and Maximum are guaranteed through test, design or statistical correlation. Typical values represent the most likely parametric norm at T<sub>J</sub> = +25°C, and are provided for reference purposes only. Unless otherwise stated the following conditions apply: V<sub>IN</sub> = 12V.

Symbol	Parameter	Conditions	Min	Typ	Max	Units
<b>Start-Up Regulator, VCC</b>						
V <sub>CC</sub>	VCC pin output voltage	C <sub>VCC</sub> = 0.47 μF, no load	<b>4.7</b>	5.5	<b>6.3</b>	V
		I <sub>CC</sub> = 2 mA	<b>4.7</b>	5.5	<b>6.3</b>	V
V <sub>CC-UVLO</sub>	VCC pin under-voltage lockout threshold (UVLO)	V <sub>CC</sub> increasing	3.56	3.78	4.00	V
V <sub>CC-UVLO-HYS</sub>	VCC pin UVLO hysteresis	V <sub>CC</sub> decreasing		310		mV
I <sub>IN</sub>	I <sub>IN</sub> operating current	No switching, V <sub>FB</sub> = 0V		3.6	<b>5.2</b>	mA
I <sub>IN-SD</sub>	I <sub>IN</sub> operating current, Device shutdown	V <sub>EN</sub> = 0V		30	<b>95</b>	μA
I <sub>VCC</sub>	VCC pin current limit (Note 4)	V <sub>CC</sub> = 0V	<b>18</b>	30		mA
V <sub>CC-VOUT</sub>	VCC pin output voltage when supplied by VOUT	V <sub>IN</sub> = Open, I <sub>CC</sub> = 1 mA, V <sub>OUT</sub> = 18V	3.5	4.1	4.7	V
<b>Enable Input</b>						
V <sub>EN</sub>	EN pin input threshold	V <sub>EN</sub> rising	<b>1.55</b>	1.63	<b>1.71</b>	V
V <sub>EN-HYS</sub>	EN pin threshold hysteresis	V <sub>EN</sub> falling		194		mV
I <sub>EN-SHUT</sub>	Enable Pull-up Current at shutdown	V <sub>EN</sub> = 0V		2		μA
I <sub>EN-OPER</sub>	Enable Pull-up Current during operation	V <sub>EN</sub> = 2V		40		μA
<b>Current Regulator</b>						
V <sub>IREF</sub>	IREF pin voltage	V <sub>IN</sub> = 4.5V to 65V	<b>1.231</b>	1.256	<b>1.281</b>	V
V <sub>DHC50</sub>	V <sub>IOUT</sub> under DHC at I <sub>OUT</sub> = 50 mA	R <sub>IREF</sub> = 25 kΩ	<b>0.160</b>	0.225	<b>0.290</b>	V
V <sub>DHC100</sub>	V <sub>IOUT</sub> under DHC at I <sub>OUT</sub> = 100 mA	R <sub>IREF</sub> = 12.5 kΩ	<b>0.38</b>	0.48	<b>0.58</b>	V
V <sub>DHC200</sub>	V <sub>IOUT</sub> under DHC at I <sub>OUT</sub> = 200 mA	R <sub>IREF</sub> = 6.25 kΩ	<b>0.81</b>	0.99	<b>1.17</b>	V
I <sub>OUT50</sub>	Current Output under DHC at V <sub>IOUT</sub> = V <sub>DHC50</sub>	R <sub>IREF</sub> = 25 kΩ, V <sub>IOUT</sub> = V <sub>DHC50</sub>	47.5	50	52.5	mA
		R <sub>IREF</sub> = 25 kΩ, V <sub>IOUT</sub> = V <sub>DHC50</sub>	<b>46.5</b>	50	<b>53.5</b>	mA
I <sub>OUT100</sub>	Current Output under DHC at V <sub>IOUT</sub> = V <sub>DHC100</sub>	R <sub>IREF</sub> = 12.5 kΩ, V <sub>IOUT</sub> = V <sub>DHC100</sub>	97	100	103	mA
		R <sub>IREF</sub> = 12.5 kΩ, V <sub>IOUT</sub> = V <sub>DHC100</sub>	<b>96</b>	100	<b>104</b>	mA
I <sub>OUT200</sub>	Current Output under DHC at V <sub>IOUT</sub> = V <sub>DHC200</sub>	R <sub>IREF</sub> = 6.25 kΩ, V <sub>IOUT</sub> = V <sub>DHC200</sub>	194	200	206	mA
		R <sub>IREF</sub> = 6.25 kΩ, V <sub>IOUT</sub> = V <sub>DHC200</sub>	<b>192</b>	200	<b>208</b>	mA
I <sub>OUTOFF</sub>	Leakage at Maximum Work Voltage	V <sub>DIM</sub> = 0, V <sub>IOUT</sub> = 65V			5	μA
V <sub>IOUT50-MIN</sub>	Minimum Work Voltage, 50 mA	R <sub>IREF</sub> = 25 kΩ, I <sub>OUT</sub> = 0.98 × I <sub>OUT50</sub>		0.1	0.15	V
V <sub>IOUT100-MIN</sub>	Minimum Work Voltage, 100 mA	R <sub>IREF</sub> = 12.5 kΩ, I <sub>OUT</sub> = 0.98 × I <sub>OUT100</sub>		0.2	0.35	V
V <sub>IOUT200-MIN</sub>	Minimum Work Voltage, 200 mA	R <sub>IREF</sub> = 6.25 kΩ, I <sub>OUT</sub> = 0.98 × I <sub>OUT200</sub>		0.4	0.65	V
V <sub>DIM-HIGH</sub>	DIM Voltage HIGH		<b>1.17</b>			V
V <sub>DIM-LOW</sub>	DIM Voltage LOW				<b>0.7</b>	V

Symbol	Parameter	Conditions	Min	Typ	Max	Units
<b>Boost Converter</b>						
$I_{CDHC-SRC}$	CDHC pin source current	$V_{CDHC} = 1.6V, V_{FB} = 3V, V_{IOUT} = 0V, DIM = High$		60		$\mu A$
$I_{CDHC-SINK}$	CDHC pin sink current	$V_{CDHC} = 1.6V, V_{FB} = 3V, V_{IOUT} = 3V, DIM = High$		56		$\mu A$
$I_{CDHC-LEAKAGE}$	CDHC pin leakage current	$DIM = Low, V_{CDHC} = 2.6V$		5	46	nA
$I_{CL-MAX}$	Integrated MOSFET peak current limit threshold		<b>3.3</b>	3.9	<b>4.5</b>	A
$I_{CL-HALF}$	Half integrated MOSFET peak current limit threshold	$R_{ILIM} = 11 k\Omega$		2.0		A
$R_{DS(on)}$	Integrated MOSFET $R_{DS(on)}$	$I_{SW} = 500 mA$		0.19	<b>0.43</b>	$\Omega$
$V_{FBTH-PWRGD}$	Power-Good FB pin threshold			2.25		V
$I_{FB}$	Feedback pin input current	$V_{FB} = 3V$			1	$\mu A$
$t_{on}$	ON timer pulse width	$V_{IN} = 12V, V_{OUT} = 65V, R_{RT} = 300 k\Omega$		1460		ns
		$V_{IN} = 24V, V_{OUT} = 32.5V, R_{RT} = 300 k\Omega$		800		ns
		$V_{IN} = 12V, V_{OUT} = 65V, R_{RT} = 100 k\Omega$		550		ns
		$V_{IN} = 24V, V_{OUT} = 32.5V, R_{RT} = 100 k\Omega$		350		ns
$t_{on-min-ILIM}$	ON timer minimum pulse width at current limit			145		ns
$t_{off}$	OFF timer pulse width			145	<b>350</b>	ns
<b>COMM PIN</b>						
$V_{IOUT-OV}$	IOUT pin over-voltage threshold	COMM goes LOW during $V_{IOUT}$ rising, other $V_{IOUT} = 1.2V$	<b>5.6</b>	6.7	<b>7.8</b>	V
$V_{COMM-LOW}$	COMM pin at LOW	5 mA into COMM			0.7	V
$I_{LEAK-FAULT}$	COMM pin Open Leakage	$V_{COMM} = 5V$			<b>5</b>	$\mu A$
<b>Thermal Protection</b>						
$T_{OTM}$	Over-temperature indication	$T_J$ rising		135		$^{\circ}C$
$T_{OTM-HYS}$	Over-temperature indication hysteresis	$T_J$ falling		15		$^{\circ}C$
$T_{SD}$	Thermal shutdown temperature	$T_J$ rising		165		$^{\circ}C$
$T_{SD-HYS}$	Thermal shutdown temperature hysteresis	$T_J$ falling		20		$^{\circ}C$

**Note 1:** Absolute Maximum Ratings are limits which damage to the device may occur. Operating ratings are conditions under which operation of the device is intended to be functional. For guaranteed specifications and test conditions, see the electrical characteristics. Thermal shutdown might occur within ambient operating temperature range as junction temperature rises above TSD level, customer should refer to efficiency data and thermal resistance data to estimate the junction temperature to ambient temperature delta.

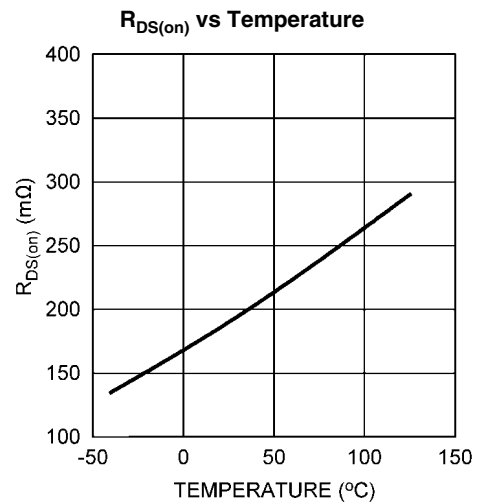
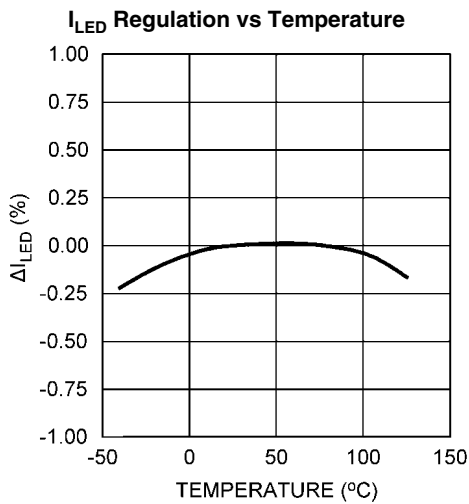
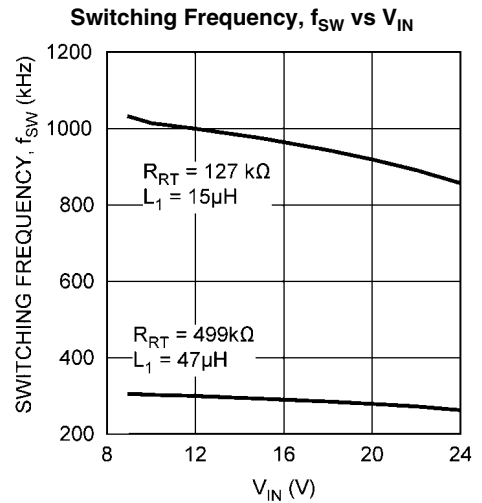
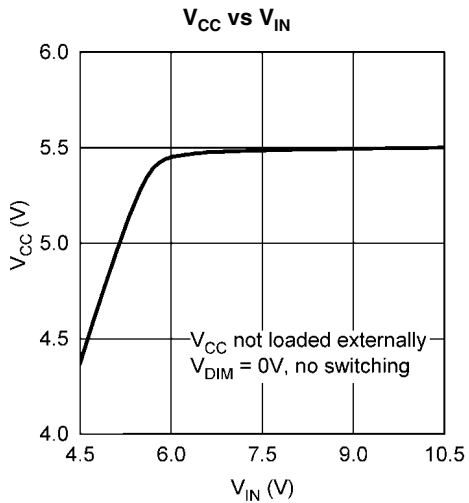
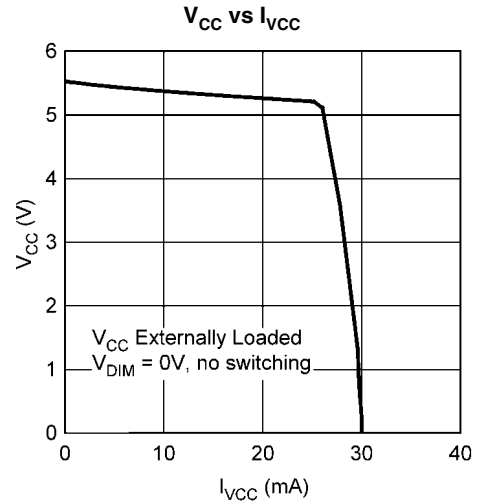
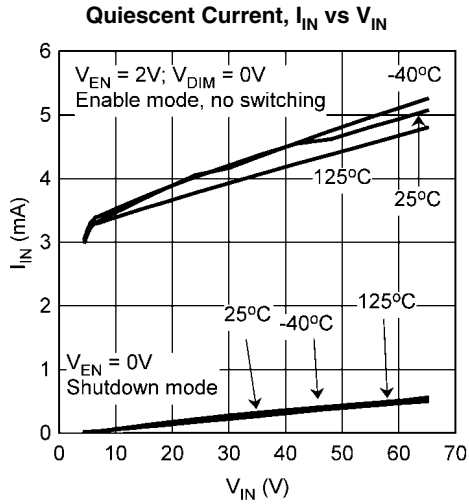
**Note 2:** The human body model is a 100 pF capacitor discharged through a 1.5 k $\Omega$  resistor into each pin.

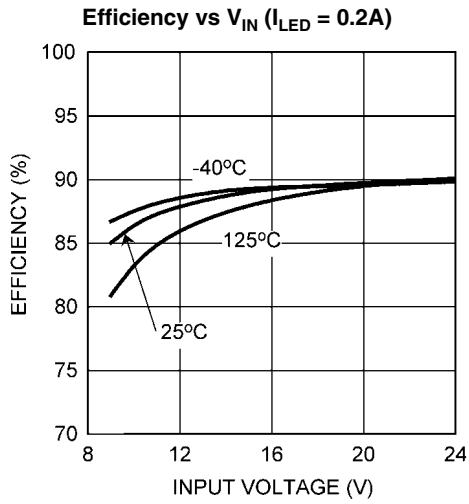
**Note 3:** The  $\theta_{JA}$  is measured on a 4-layer standard JEDEC thermal test board with 12 vias, no air flow and 1W power dissipation. Thermal shutdown will occur if the junction temperature exceeds 165 $^{\circ}C$ . The maximum power dissipation is a function of  $T_{J(MAX)}$ ,  $\theta_{JA}$  and  $T_A$ . The maximum allowable power dissipation at any ambient temperature is  $PD = (T_{J(MAX)} - T_A) / \theta_{JA}$ .

**Note 4:**  $V_{CC}$  provides self bias for the internal gate drive and control circuits. Device thermal limitations limit external loading.

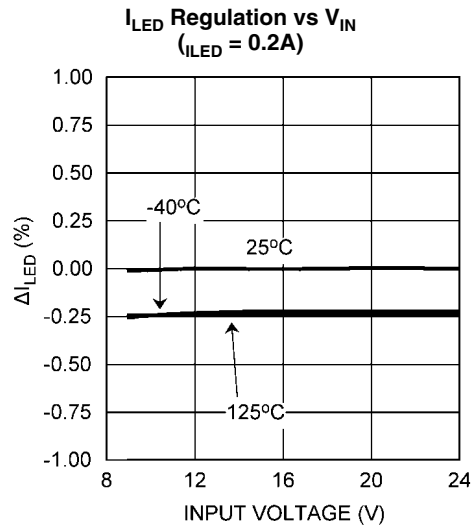
## Typical Performance Characteristics

Unless otherwise specified the following conditions apply:  $T_J = 25^\circ\text{C}$ ,  $V_{IN} = 12\text{V}$  with configuration in typical application circuit for  $I_{LED} = 200\text{ mA}$  shown in this datasheet.

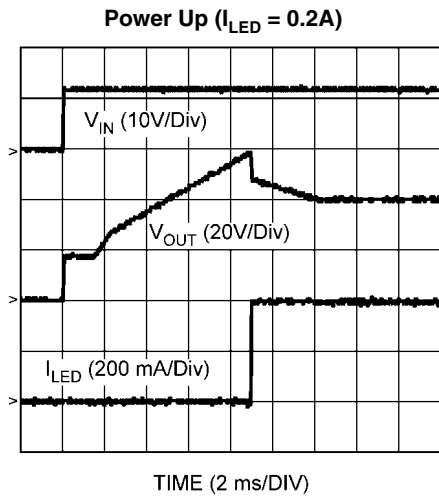




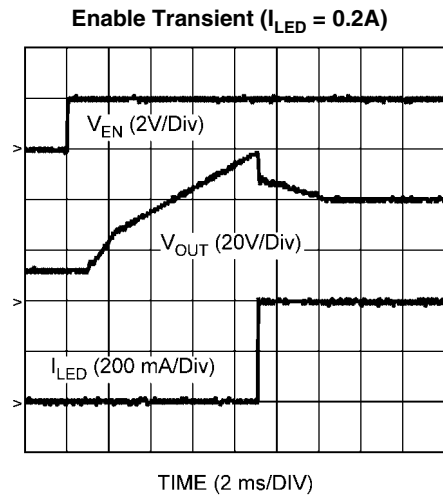
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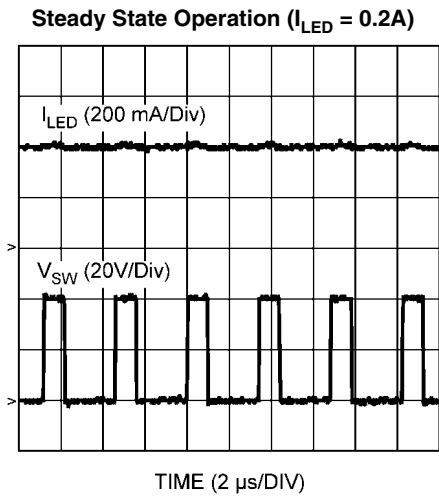
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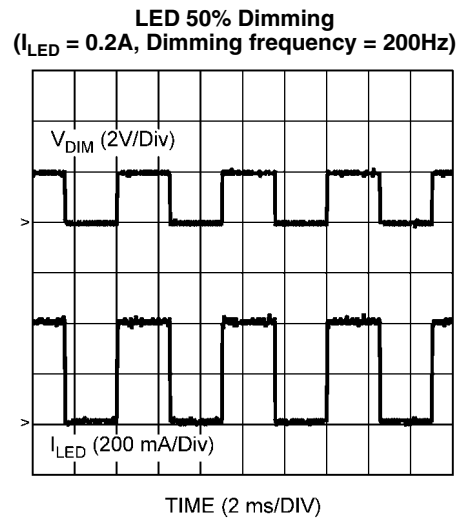
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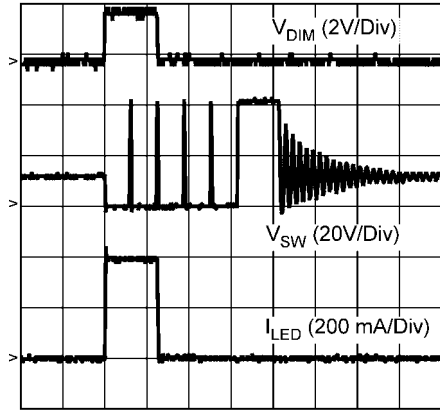


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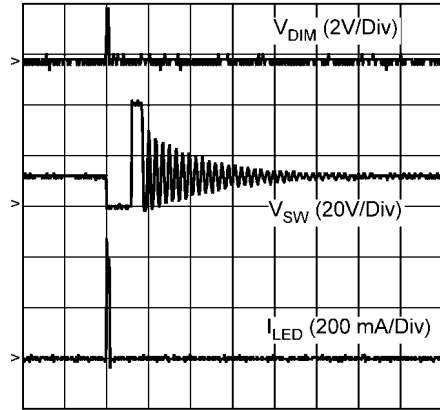
1000:1 LED Dimming  
( $I_{LED} = 0.2A$ , Dimming frequency = 200Hz)



TIME (4 μs/DIV)

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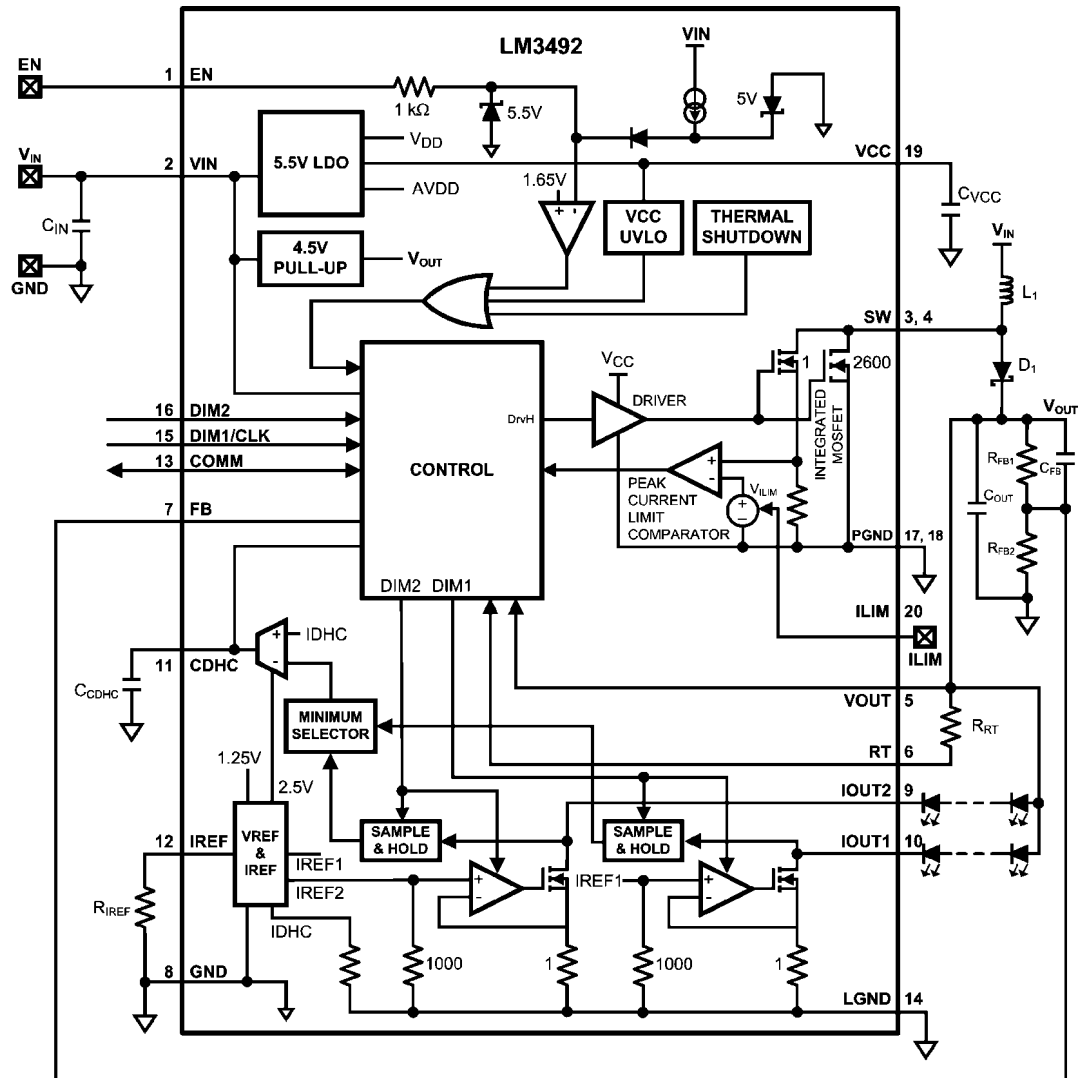
300 ns LED Dimming Pulse Width  
( $I_{LED} = 0.2A$ , Dimming frequency = 3.33 kHz)



TIME (4 μs/DIV)

30116124

### Simplified Functional Block Diagram



30116103



## Overview

The LM3492 integrates a boost converter and a two-channel current regulator to implement a high efficient and cost effective LED driver for driving two individually dimmable LED strings with a maximum power of 15W and an output voltage of up to 65V. The boost converter provides power for the LED strings, and the current regulator controls the dimming of the LED strings individually. The LM3492 integrates an N-channel MOSFET switch and a two-channel current regulator in order to minimize the component count and solution size.

The boost converter of the LM3492 employs a Projected On-Time (POT) control method to determine the on-time of the MOSFET with respect to the input and output voltages and an external resistor  $R_{RT}$ . During the on-period, the boost inductor is charged up, and the output capacitor is discharged to provide power to the output. A cycle-by-cycle current limit (which is 3.9A typically and programmable by an external resistor) is imposed on the MOSFET for protection. After the on-period, the MOSFET is turned off such that the boost inductor is discharged. The next on-period is started when the voltage of the FB pin is dropped below a threshold which is determined by Dynamic Headroom Control (DHC) and is ranged from 1.05V to 2.5V (DHC affects the threshold only when the DIM1 and/or DIM2 pins are high). The boost converter under POT control can maintain the switching frequency nearly constant so that the switching frequency depends on only  $R_{RT}$  (Figure 1). Also, POT control requires no compensation circuit and gives a fast transient response of the output voltage.

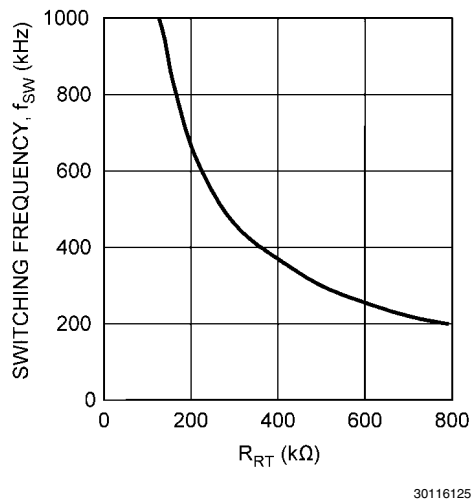


FIGURE 1. Switching Frequency

The two-channel current regulator of the LM3492 is fast response so that it can allow very high contrast ratio (1000:1 at 3 kHz LED dimming frequency, minimum pulse width of the dimming signal is 300 ns). The two channels are dimmable individually. Channel 1 of the current regulator can be disabled by a digital command send through the COMM pin. In this case, the DIM1 pin can serve only as a clock signal for the data flow of the COMM pin. The power dissipated by the current regulator is adaptively minimized by Dynamic Headroom Control in order to maximize efficiency.

The LM3492 can be applied in numerous applications like automotive LCD backlight panels. It can operate efficiently for inputs as high as 65V. Diagnostic functions including power good indication, over-temperature indication, IOUT over- and under-voltage indications facilitate the interface of the

LM3492 application circuit with external micro-processors (MCUs). The LM3492 will not latch off and continue to operate in the presence of the indications. Other useful features include thermal shutdown, VCC under-voltage lock-out, and precision enable. The LM3492 is available in the thermally enhanced eTSSOP-20 package.

## LDO Regulator

A 5.5V LDO regulator is integrated in the LM3492. For stability, an external capacitor  $C_{VCC}$  of more than 0.47  $\mu$ F should be connected between the VCC and GND pins. The current limit of the LDO is typically 30 mA. It can be used to pull-up the open-drain COMM pin with an external resistor, and inject current to the ILIM pin to adjust the current limit of the integrated MOSFET. When the voltage on the VCC pin ( $V_{CC}$ ) is higher than the under-voltage lock-out (UVLO) threshold of 3.78V, the LM3492 is enabled and the CDHC pin sources a current to charge up an external capacitor  $C_{CDHC}$  to provide a soft-start function.

## Enable and Disable

To enable the LM3492, the voltage on the EN pin ( $V_{EN}$ ) must be higher than an enable threshold of typically 1.63V. If  $V_{EN}$  is lower than 1.43V, the LM3492 is shutdown. In this case, the LDO regulator is turned off and the CDHC pin is internally grounded. The EN pin is internally pulled up. After enable, the EN pin is pulled up by a 40  $\mu$ A current source. If the EN pin is connected to low such that the LM3492 is shutdown, the pull-up current is reduced to 2  $\mu$ A. These take the advantages that the LM3492 can effectively avoid false disabling by noise during operation, and minimize power consumption during shutdown. The enable threshold is precise such that it can be used to implement an UVLO function for the input voltage as shown in Figure 2. The input voltage can be connected to the EN pin through a resistor divider consists of  $R_{EN1}$  and  $R_{EN2}$ . This can ensure that the LM3492 is operated after the input voltage reaches a minimum require value  $V_{IN(EN)}$ , which can be calculated by the following equation:

$$V_{IN(EN)} = 1.63V(1 + R_{EN1}/R_{EN2}) \quad (1)$$

A zener diode  $D_{EN}$  should be placed between the EN and GND pins to keep  $V_{EN}$  below its absolute maximum caused by the increase of the input voltage.

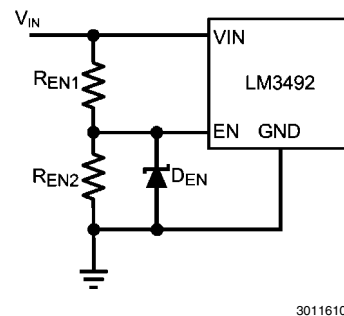


FIGURE 2. Input Voltage UVLO Implemented by Precision Enable

Once the EN pin is pulled low, the LM3492 will perform the following functions: reset IOUT over- and under-voltage indications and the corresponding COMM bit pattern, resume the switching frequency tuning to the normal frequency, and resume channel 1 of the current regulator if it is disabled. Pulling the EN pin low for a short period of about 200 ns can achieve

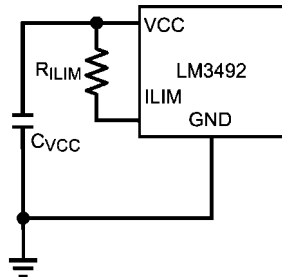
the above functions with nearly no effect on the operation of the boost converter and the current regulator.

## Current Limit

The current limit  $I_{CL}$  of the integrated MOSFET of the LM3492 provides a cycle-by-cycle current limit for protection. It can be decreased by injecting a small signal current  $I_{ILIM}$  into the ILIM pin, and the relationship between  $I_{CL}$  and  $I_{ILIM}$  is

$$I_{CL} = I_{CL(MAX)} - 4290 I_{ILIM} \quad (2)$$

where  $I_{CL(MAX)}$  is the maximum current limit. Its typical value is 3.9A. As shown in Figure 3,  $I_{ILIM}$  can be provided by connecting a resistor  $R_{ILIM}$  from the VCC pin to the ILIM pin. The typical voltage on the ILIM pin is 0.7V. To obtain the maximum current limit, connect the ILIM pin to the ground.



30116105

FIGURE 3. Programmable Current Limit

## Thermal Protection

Thermal protection is implemented by an internal thermal shutdown circuit, which activates at 165°C (typically) to disable the LM3492. In this case, the LDO regulator is turned off and the CDHC pin is internally grounded. Thermal protection helps prevent catastrophic failures from accidental device overheating. When the junction temperature of the LM3492 falls back below 145°C (typical hysteresis = 20°C), the LM3492 resumes normal operation.

## Dynamic Headroom Control and Soft-start

Dynamic headroom control (DHC) is implemented in the LM3492 to adjust the output voltage  $V_{OUT}$  of the boost converter in order to reduce the power loss of the current regulator to maximize efficiency. Let  $V_{LED,n}$  and  $V_{IOUT,n}$  be the forward voltage of an LED string connecting to the IOUTn pin and the voltage of the IOUTn pin, where n is 1, 2 for channels 1, 2 of the current regulator. Since  $V_{LED,n}$  is normally decreasing gradually (in terms of minutes) owing to the rise of LED die temperature during operation, DHC adjust  $V_{OUT}$  by adjusting a threshold which is reflected on the voltage of the FB pin with reference to  $V_{IOUT,n}$ , which is the difference between  $V_{OUT}$  and  $V_{LED,n}$ . The sensitivity of DHC, which is the response time on adjusting  $V_{OUT}$ , is set by  $C_{CDHC}$ . If  $C_{CDHC}$  is small,  $V_{OUT}$  is more sensitive to the variation of  $V_{LED,n}$ .

During startup, the voltage of the CDHC pin is risen from 0V to 2.25V at a speed depends on  $C_{CDHC}$ . This makes the voltage of the FB pin as well as the output voltage ramps up in a controlled manner, and effectively a soft-start function is implemented. The soft-start can be programmable by the  $C_{CDHC}$ .

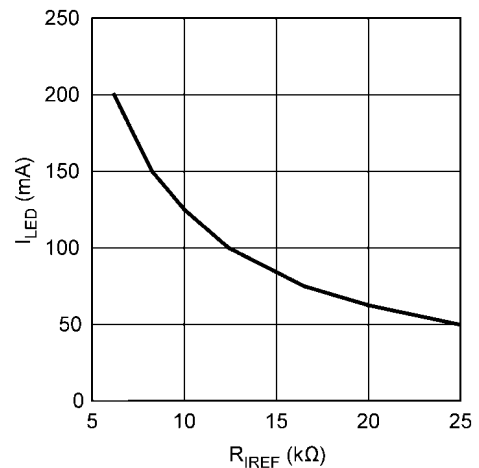
An internal switch grounds the CDHC pin if any of the following cases happens: (i)  $V_{CC}$  is below the VCC UVLO threshold;

(ii) a thermal shutdown occurs; or (iii) the EN pin is pulled low. The CDHC pin cannot be connected to the ground externally.

## Current Regulator

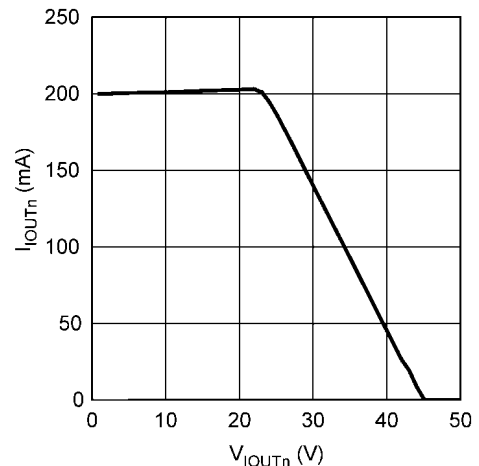
The LM3492 integrates a two-channel current regulator for controlling the current of two LED strings. The two LED strings are dimmable individually by dimming signals applied to the DIM1 and DIM2 pins for LED strings 1 and 2, which are connected from the VOUT pin to the IOUT1 and IOUT2 pins. The DIM1 and DIM2 pins are internally pulled low. The lowest contrast ratio is 1000:1. The finest pulse width of the dimming signal for the DIM1 and DIM2 pins is 300 ns, which implies that a contrast ratio of 1000:1 can be achieved at a dimming frequency of more than 3 kHz.

The current of an LED string ( $I_{LED}$ ) is programmable from 50 mA to 200 mA by an external resistor  $R_{IREF}$  connecting from the IREF pin to the ground. The relationship between  $I_{LED}$  and  $R_{IREF}$  is shown in Figure 4. The two channels of the current regulator can work in parallel for only one LED string by connecting the IOUT1 and IOUT2 pins together to provide an LED current of up to 400 mA. In this case, the DIM1 and DIM2 pins should also be connected together.



30116126

FIGURE 4.  $I_{LED}$  and  $R_{IREF}$



30116127

FIGURE 5. Over-power Protection

If the voltage on the IOUTn (n = 1, 2) pin is over 24V when channel n is on, the regulated current of channel n will be reduced linearly if the voltage further increases (as shown in Figure 5). The regulated current of another channel is not affected. This over-power protection feature avoids damaging the current regulator owing to the shorting of many LEDs in one string.

## Output Voltage Feedback

The output voltage is fed back to the FB pin through a feedback circuit consists of  $R_{FB1}$ ,  $R_{FB2}$ , and  $C_{FB}$  as shown in Figure 6. The value of  $C_{FB}$  is recommended to be 10 pF in order to help feed the AC component of the output voltage back. The DC component of the output voltage is fed back by  $R_{FB1}$  and  $R_{FB2}$ . The voltage of the FB pin  $V_{FB}$  can be adjusted from 1.05V to 2.5V by DHC. The maximum output voltage of the boost converter  $V_{OUT(MAX)}$  is

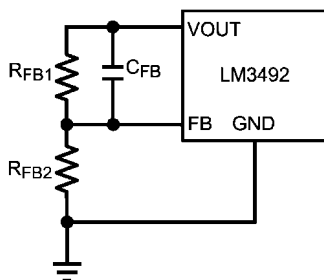
$$V_{OUT(MAX)} = 2.5V (1 + R_{FB1}/R_{FB2}) \quad (3)$$

Under DHC, the output voltage should be maintained at a nominal voltage but not the maximum. The nominal output voltage ( $V_{OUT(NOM)}$ ) is

$$V_{OUT(NOM)} = \max(V_{LED,n} + V_{IOUT,n}), n = 1, 2 \quad (4)$$

where  $V_{LED,n}$  is the forward voltage of LED string n and  $V_{IOUT,n}$  is the voltage of the IOUTn pin, where n is 1, 2 for channels 1, 2 of the current regulator). The minimum value of  $V_{IOUT,n}$  is about  $5\Omega \times I_{LED}$ . It is recommended that the nominal voltage of the FB pin ( $V_{FB(NOM)}$ ) is between 1.05V to 2V. Hence, the equation relating  $V_{OUT(MAX)}$ ,  $V_{OUT(NOM)}$ , and  $V_{FB(NOM)}$  is as follows:

$$V_{OUT(MAX)} = V_{OUT(NOM)} \times 2.5V / V_{FB(NOM)} \quad (5)$$



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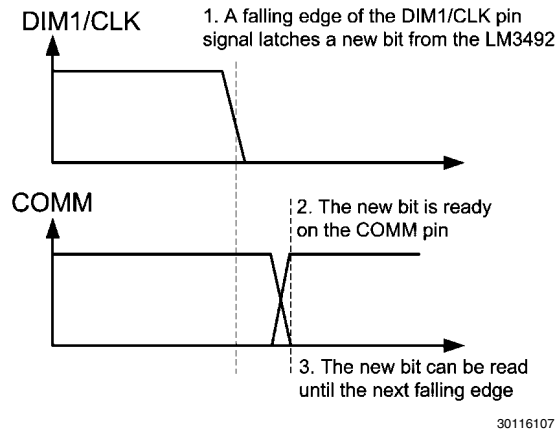
FIGURE 6. Output Voltage Feedback Circuit

## Bi-Directional Communication Pin

The COMM pin of the LM3492 is an open-drain bi-directional I/O pin for interfacing with an external MCU for the following functions: power-good indication, over-temperature indication, IOUT over- and under-voltage indications, switching frequency tuning, and channel 1 disabling. Except the power good indication and the over-temperature alert, all data flow through the COMM pin is serial and is latched by the falling edge of the signal applying to the DIM1 pin, even when channel 1 of the current regulator is disabled. If the DIM1 pin stays only low or only high, either by an external circuit or letting it open and pull low internally, data flow will not happen. Figures 7 and 8 show a timing diagram of reading and writing a bit from and to the LM3492 through the COMM pin.

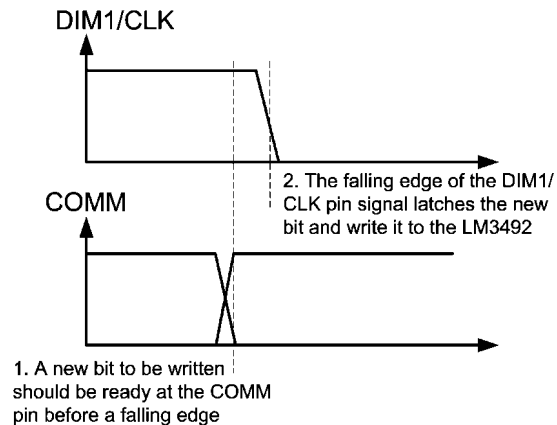
The COMM pin should be pull-up by an MCU I/O pin which has pull-up capability or an external resistor  $R_{COMM}$  to the VCC pin. Otherwise, the voltage of the COMM pin will remain at zero. The rise time of the output signal of the COMM pin

depends on the pull-up power. If the rise time is long ( $R_{COMM}$  is too large or pull-up power from the connecting MCU I/O pin is too weak), data may be ready after a longer duration after the falling edge. In this case, a longer delay between the falling edge latching and the (input or output) bit is required.



30116107

FIGURE 7. Read from the COMM Pin



30116108

FIGURE 8. Write to the COMM Pin

### POWER-GOOD INDICATION

Upon startup, the COMM pin reads low. The output voltage of the boost converter of the LM3492 will rise until the voltage on the FB pin ( $V_{FB}$ ) reaches 2.25V, when the COMM pin reads high to indicate power-good. The power-good indication is independent of the signal applied on the DIM1 pin.

### OVER-TEMPERATURE INDICATION

If the junction temperature of the LM3492 reaches 135°C, the COMM pin reads low, showing an over-temperature indication. External MCU should consider to turn off or reduce the brightness of the LED strings in order to prevent over-temperature. The over-temperature indication is independent of the signal applied on the DIM1 pin. The COMM pin reads high if the junction temperature falls below 120°C. The LM3492 will not latch off and continue to operate in the presence of the over-temperature indication.

### IOUT UNDER-VOLTAGE INDICATION

The LM3492 gives an IOUTn (n = 1, 2) under-voltage indication if the voltage of the IOUTn pin when DIMn is high is lower than its minimum required voltage which can regulate  $I_{LED}$ .

and the voltage of the CDHC pin reaches its maximum, and these conditions last for 508 consecutively dimming signals applied on the DIMn pin. This means that the current of the LED string n does not reach its regulated value. In most case, the IOUT under-voltage indication can be regarded as an open fault of the LED string n. A bit pattern (refer to table 1) can be read from the COMM pin. The LM3492 will not latch off and continue to operate in the presence of the IOUT under-voltage indication.

### IOUT OVER-VOLTAGE INDICATION

The LM3492 gives an IOUTn (n = 1, 2) over-voltage indication if the voltage of the IOUTn pin when DIMn is high is higher than a threshold of typically 6.5V, and the condition lasts for 508 consecutively dimming signals applied on the DIMn pin. Except powering up the LM3492 at a very low dimming ratio such that  $V_{OUT}$  maintains at a maximum and DHC is not fast enough to reduce  $V_{OUT}$ , the IOUT over-voltage indication can be regarded as a short fault of the LED string n. A bit pattern (refer to table 1) can be read from the COMM pin. The LM3492 will not latch off and continue to operate in the presence of the IOUT over-voltage indication.

TABLE 1. COMM Bit Patterns

Indication	COMM Bit Pattern
IOUT1 over-voltage indication	0001
IOUT2 over-voltage indication	0011
IOUT1 under-voltage indication	0101
IOUT2 under-voltage indication	0111

### COMM BIT PATTERN

Table 1 summarized all COMM bit patterns of IOUT over- and under-voltage indications. A bit pattern can be read from the COMM pin continuously even the condition causing IOUT over- and under-voltage disappeared. An existing COMM bit pattern will be clear if one of the following condition occurs:

- the LM3492 is shutdown;
- the LM3492 is disabled by pulling the EN pin low;
- the over-temperature indication is appearing.

To clear the COMM bit pattern without affecting the operation of the boost converter and the current regulator, a reset signal which pulls the EN pin low for about 200 ns can be applied. In this case, the COMM pin will not give bit pattern any more unless a condition causing IOUT over- or under-voltage appears and lasts for 508 consecutively dimming signals.

If an external MCU is used to read the COMM bit pattern, it is recommended that a reset signal should be send to clear the COMM bit pattern after the bit pattern is read. This can avoid that the existing COMM bit pattern is overwritten by another pattern appeared in a later time.

In case of over-temperature, the COMM pin will be pulled low to give an over-temperature indication disregard of any existing COMM bit pattern. After that the over-temperature indication disappears, the COMM bit pattern appeared before the over-temperature indication will appear again.

### SWITCHING FREQUENCY TUNING

After power good, the switching frequency ( $f_{SW}$ ) of the LM3492 can be tuned down 20% or 40%, or resume normal by writing commands (refer to table 2) to the COMM pin. This

helps avoid interfering some sensitive devices, for example radios, working nearby the LM3492. Upon reset,  $f_{SW}$  of the LM3492 will resume normal by default. If the over-temperature indication or any COMM bit pattern has already presented, no command can be written to the LM3492.

### CHANNEL 1 DISABLE

After power good, channel 1 of the current regulator can be disabled by writing a command (refer to table 2) to the COMM pin. If LED string 1 is malfunctioning, channel 1 can be disabled and the signal applied on the DIM1 pin can serve as only a clock signal for the data flow of the COMM pin. Channel 1 is by default enabled after reset. If the over-temperature indication or any COMM bit pattern has already presented, no command can be written to the LM3492.

TABLE 2. Commands

Command	Command Bit Pattern
$f_{SW}$ resume normal	1111011101110111
$f_{SW}$ tune down by 20%	1111000100010001
$f_{SW}$ tune down by 40%	1111001100110011
Channel 1 disable	1111010101010101

## Application Information

### EXTERNAL COMPONENTS

The following procedures are to design an LED driver using the LM3492 with an input voltage ranged from 9V to 24V and two LED strings consists of 10 LEDs each with a forward voltage of 3.8V for each LED when running at 200 mA. The output power is 15.2W. The switching frequency  $f_{SW}$  is designed to be 300 kHz.

**$R_{FB1}$ ,  $R_{FB2}$ , and  $C_{FB}$ :** The nominal voltage of the LED string with 10 LEDs is 38V, and the minimum voltage of the IOUTn pin (n = 1, 2) is 1V when  $I_{LED}$  is 200 mA. Hence,  $V_{OUT(NOM)}$  is 39V. Since  $V_{FB(NOM)}$  is recommended to be ranged from 1.05V to 2V during operation, design  $V_{FB(NOM)}$  to be 1.5V when  $V_{OUT(NOM)}$  is 39V. From (5),  $V_{OUT(MAX)}$  is 65V. Also, design  $R_{FB2}$  to be 16.2 k $\Omega$ , from (3),  $R_{FB1}$  is calculated to be 405 k $\Omega$ , and a standard resistor value of 402 k $\Omega$  is selected.  $C_{FB}$  is selected to be 10 pF as recommended.

**$L_1$ :** The main parameter affected by the inductor is the peak to peak inductor current ripple ( $I_{LR}$ ). To maintain a continuous conduction mode (CCM) operation, the average inductor current  $I_{L1}$  should be larger than half of  $I_{LR}$ . For a boost converter,  $I_{L1}$  equals to the input current  $I_{IN}$ . Hence,

$$I_{IN} = (V_{OUT(NOM)} \times 2I_{LED}) / V_{IN} \quad (6)$$

Also,

$$t_{on} = (1 - V_{IN}/V_{OUT}) / f_{SW} \quad (7)$$

$$L_1 = (V_{IN} \times t_{on}) / 2I_{IN} \quad (8)$$

If  $V_{IN}$  is maximum, which is 24V in this example, and only one LED string is turned on (since the two channels of the LM3492 are individually dimmable),  $I_{IN}$  is minimum. From (6)-(8), it can be calculated that  $I_{IN(MIN)}$ ,  $t_{on}$ , and  $L_1$  are 0.325A, 1.28  $\mu$ s, and 47  $\mu$ H. On the other hand, from (6),  $I_{IN}$  is maximum when  $V_{IN}$  is minimum, which is 9V in this example, and the two LED strings are turned on together. Hence  $I_{IN(MAX)}$  is 1.73A. Then,  $I_{LR}$  is

$$I_{LR} = (V_{IN} \times t_{on}) / L_1 \quad (9)$$

From (7),  $t_{on}$  is 2.56  $\mu$ s. From (9),  $I_{LR}$  is 0.49A. The steady state peak inductor current  $I_{L1(PEAK)}$  is

$$I_{L1(PEAK)} = I_{L1} + I_{LR} / 2 \quad (10)$$

As a result,  $I_{L1(PEAK)}$  is 1.98A. A standard value of 47  $\mu\text{H}$  is selected for  $L_1$ , and its saturation current is larger than 1.98A.

**$D_1$ :** The selection of the boost diode  $D_1$  depends on two factors. The first factor is the reverse voltage, which equals to  $V_{OUT}$  for a boost converter. The second factor is the peak diode current at the steady state, which equals to the peak inductor current as shown in (10). In this example, a 100V 3A schottky diode is selected.

**$C_{IN}$  and  $C_{OUT}$ :** The function of the input capacitor  $C_{IN}$  and the output capacitor  $C_{OUT}$  is to reduce the input and output voltage ripples. Experimentation is usually necessary to determine their value. The rated DC voltage of capacitors used should be higher than the maximum DC voltage applied. Owing to the concern of product lifetime, ceramic capacitors are recommended. But ceramic capacitors with high rated DC voltage and high capacitance are rare in general. Multiple capacitors connecting in parallel can be used for  $C_{IN}$  and  $C_{OUT}$ . In this example, two 10  $\mu\text{F}$  ceramic capacitor are used for  $C_{IN}$ , and two 2.2  $\mu\text{F}$  ceramic capacitor are used for  $C_{OUT}$ .

**$C_{VCC}$ :** The capacitor on the VCC pin provides noise filtering and stabilizes the LDO regulator. It also prevents false triggering of the VCC UVLO.  $C_{VCC}$  is recommended to be a 1  $\mu\text{F}$  good quality and low ESR ceramic capacitor.

**$C_{CDHC}$ :** The capacitor at the CDHC pin not only affects the sensitivity of the DHC but also determines the soft-start time  $t_{SS}$ , i.e. the time for the output voltage to rise until power good.  $t_{SS}$  is determined from the following equation:

$$t_{SS} = \frac{C_{CDHC} \times 2.25V}{120 \mu A}$$

In this example,  $C_{CDHC}$  is recommended to be a 0.47  $\mu\text{F}$  good quality and low ESR ceramic capacitor.

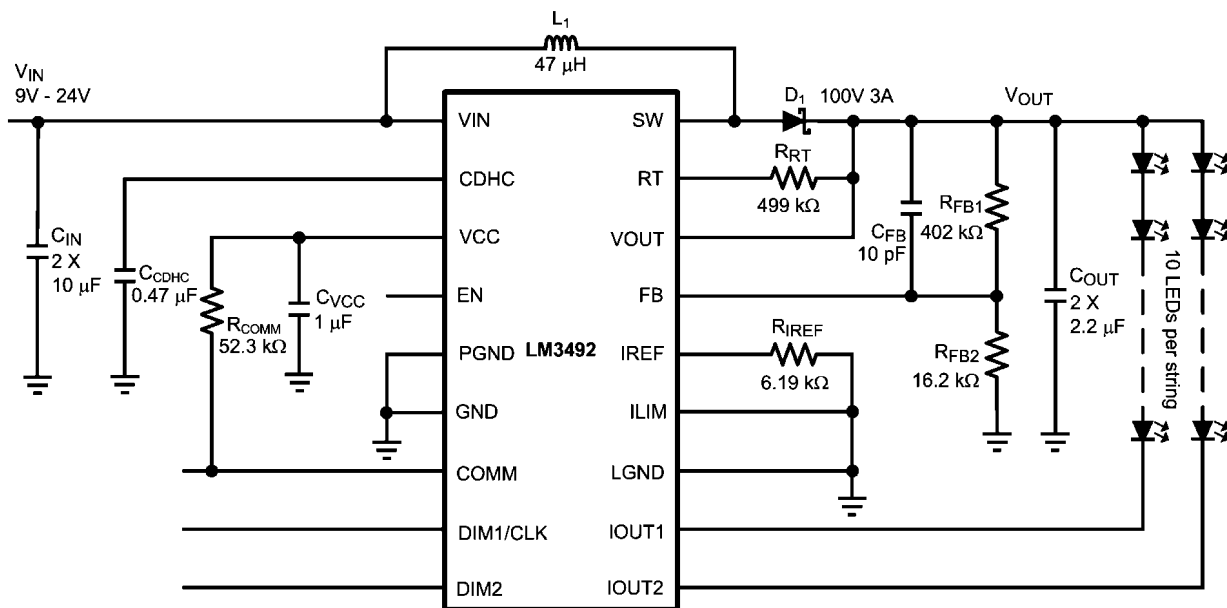
**$R_{RT}$  and  $R_{IREF}$ :** The resistors  $R_{RT}$  and  $R_{IREF}$  set the switching frequency  $f_{SW}$  of the boost converter and the LED current  $I_{LED}$  respectively. From Figure 1, if  $f_{SW}$  is 300 kHz,  $R_{RT}$  is selected to be 499 k $\Omega$ . From Figure 4, if  $I_{LED}$  is 200 mA,  $R_{IREF}$  is selected to be 6.19 k $\Omega$ .

**$R_{COMM}$ :** Since the COMM pin is open drain, a resistor  $R_{COMM}$  of 52.3 k $\Omega$  is used to connect the VCC and COMM pins to act as a pull-up function.

## PC Board Layout

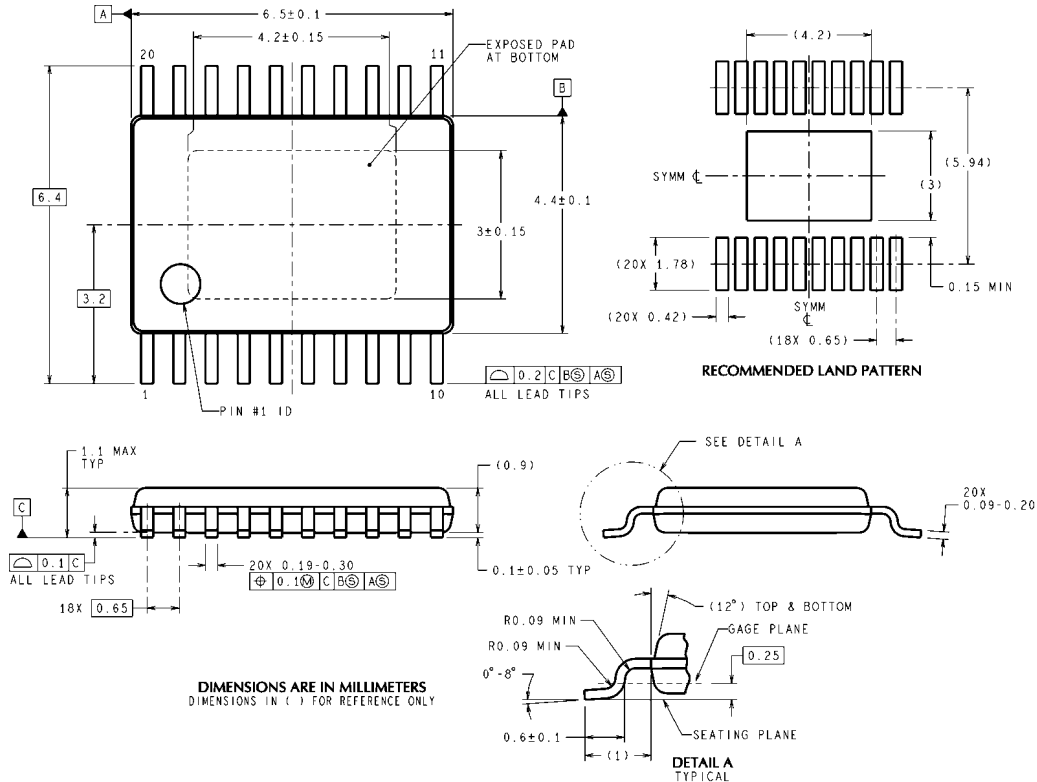
The layout of the printed circuit board is critical in order to optimize the performance of the LM3492 application circuit. In general, external components should be placed as close to the LM3492 and each other as possible in order to make copper traces short and direct. In particular, components of the boost converter  $C_{IN}$ ,  $L_1$ ,  $D_1$ ,  $C_{OUT}$ , and the LM3492 should be closed. Also, the output feedback capacitor  $C_{FB}$  should be closed to the output capacitor  $C_{OUT}$ . The ground plane connecting the GND, PGND, and LGND pins and the exposed pad of the LM3492 and the ground connection of the  $C_{IN}$  and  $C_{OUT}$  should be placed on the same copper layer.

Good heat dissipation helps optimize the performance of the LM3492. The ground plane should be used to connect the exposed pad of the LM3492, which is internally connected to the LM3492 die substrate. The area of the ground plane should be extended as much as possible on the same copper layer around the LM3492. Using numerous vias beneath the exposed pad to dissipate heat of the LM3492 to another copper layer is also a good practice.



30116110

**Physical Dimensions** inches (millimeters) unless otherwise noted



**20-Lead Plastic eTSSOP Package**  
**NS Package Number MXA20A**

MXA20A (Rev C)

## Notes

LM3492/LM3492Q

## Notes

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LVDS	<a href="http://www.national.com/lvds">www.national.com/lvds</a>	Packaging	<a href="http://www.national.com/packaging">www.national.com/packaging</a>
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