High-Voltage, Current-Mode PWM Controller

Features

- ▶ 10 to 450V input voltage range
- <1.3mA supply current</p>
- >1.0MHz clock
- >20:1 dynamic range @ 500KHz
- 49% Maximum duty cycle version
- Low internal noise

Applications

- ▶ Off-line high frequency power supplies
- Universal input power supplies
- High density power supplies
- Very high efficiency power supplies
- Extra wide load range power supplies

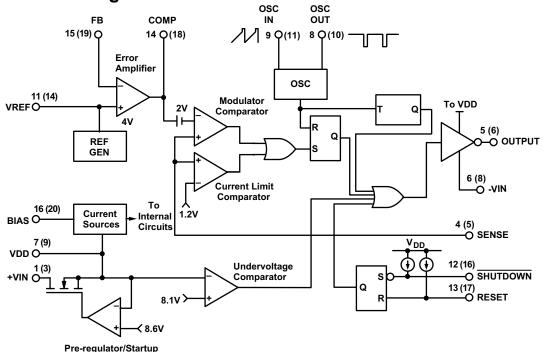
General Description

The Supertex HV9120 is a Switch Mode Power Supply (SMPS) controller subsystem that can start and run directly from almost any DC input, from a 12V battery to a rectified and filtered 240V AC line. It contains all the elements required to build a single-switch converter except for the switch, magnetic assembly, output rectifier(s) and filter(s).

A unique input circuit allows the 9120 to self-start directly from a high voltage input, and subsequently take the power to operate from one of the outputs of the converter it is controlling, allowing very efficient operation while maintaining input-to-output galvanic isolation limited in voltage only by the insulation system of the associated magnetic assembly. A ±2% internal bandgap reference, internal operational amplifier, very high speed comparator, and output buffer allow production of rugged, high performance, high efficiency power supplies of 50W or more, which can still be over 80% efficient at outputs of 1.0W or less. The wide dynamic range of the controller system allows designs with extremely wide line and load variations with much less difficulty and much higher efficiency than usual. The exceptionally wide input voltage range also allows better usage of energy stored in input dropout capacitors than with other PWM ICs. Remote on/off controls allow either latching or nonlatching remote shutdown. During shutdown, the power required is under 6.0mW.

For detailed circuit and application information, please refer to application notes AN-H13 and AN-H21 to AN-H24.

Functional Block Diagram



Note.

Pin numbers in parentheses are for PLCC package.

Ordering Information

	Package Options									
Device	16-Lead SOIC 9.90x3.90mm body 1.75mm height (max) 1.27mm pitch	16-Lead PDIP .790x.250in body .210in height (max) .100in pitch	20-Lead PLCC .353x.353in body .180in height (max) .050in pitch							
HV9120	HV9120NG-G	HV9120P-G	HV9120PJ-G							

⁻G indicates package is RoHS compliant ('Green')



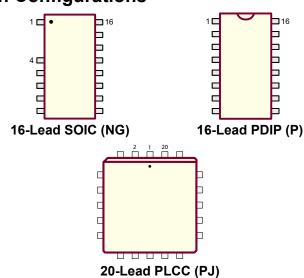


Absolute Maximum Ratings

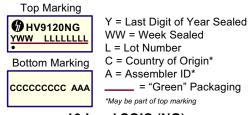
Parameter	Value
Input voltage, +V _{IN}	450V
Device supply voltage, V _{DD}	15.5V
Logic input voltage	-0.3V to V _{DD} +0.3V
Linear input voltage	-0.3V to V _{DD} +0.3V
Pre regulator input current (continuous), I _{IN}	2.5mA
Operating junction temperature, T _J	150°C
Storage temperature	-65 to +150°C
Power dissipation:	
16-Lead SOIC	900mW
16-Lead PDIP	1000mW
20-Lead PLCC	1400mW

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability. **Voltages are referenced to -V**_{Inc}.

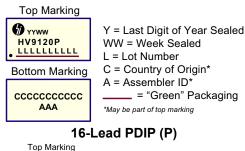
Pin Configurations

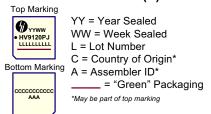


Product Marking



16-Lead SOIC (NG)





20-Lead PLCC (PJ)

Electrical Characteristics

(Unless otherwise specified, V_{DD} = 10V, + V_{IN} = 48V, R_{BIAS} = 390K Ω , R_{OSC} = 330K Ω , T_A = 25°C.)

Parameter	#	Min Typ Max			Units	Conditions			
e									
Output voltage		3.92	4.00	4.08	V	$R_L = 10M\Omega$			
Output voltage		3.84	4.00	4.16	V	$R_{L} = 10M\Omega, T_{A} = -55 \text{ to } 125^{\circ}C$			
Output impedance	#	15	30	45	ΚΩ				
Short circuit current	-	-	125	250	μA	$V_{REF} = -V_{IN}$			
Change in V_{REF} with temperature	#	-	0.25	-	mV/°C	T _A = -55 to 125°C			
•									
Oscillator frequency	-	1.0	3.0	-	MHz	$R_{OSC} = 0\Omega$			
Initial accuracy	-	80	100	120	VU→	$R_{OSC} = 330K\Omega$			
illilial accuracy	-	160	200	240	KHZ	$R_{OSC} = 150K\Omega$			
Voltage stability	-	-	-	15	%	9.5V < V _{DD} < 13.5V			
Temperature coefficient	#	-	170	-	ppm/°C	T _A = -55 to 125°C			
Maximum duty cycle	#	49.0	49.4	49.6	%				
Minimum duty cycle	-	-	-	0	%				
Maximum pulse width before pulse drops out	#	-	80	125	ns				
imit			,						
Maximum input signal	-	1.0	1.2	1.4	V	V _{FB} = 0V			
Delay to output	#	-	80 120		ns	V _{SENSE} = 1.5V, V _{COMP} ≤ 2.0V			
plifier									
Feedback voltage	-	3.92	4.00	4.08	V	V _{FB} shorted to COMP			
Input bias current	-	-	25	500	nA	V _{FB} = 4.0V			
Input offset voltage	-	nulled during		rim	-				
Open loop voltage gain	#	60	80	-	dB				
Unity gain bandwidth	#	1.0	1.3	-	MHz				
Out impedance	#	see	e Fig. 1		Ω				
Output source current	-	-1.4	-2.0	-	mA	V _{FB} = 3.4V			
Output sink current	- 1	0.12	0.15	-	mA	V _{FB} = 4.5V			
Power supply rejection	#	see	e Fig. 2		dB				
	Output voltage Output impedance Short circuit current Change in V _{REF} with temperature Oscillator frequency Initial accuracy¹ Voltage stability Temperature coefficient Maximum duty cycle Minimum duty cycle Minimum pulse width before pulse drops out imit Maximum input signal Delay to output Diffier Feedback voltage Input bias current Input offset voltage Open loop voltage gain Unity gain bandwidth Out impedance Output source current Output sink current	Output voltage - Output impedance # Short circuit current - Change in V _{REF} with temperature # Oscillator frequency - Initial accuracy¹ - Voltage stability - Temperature coefficient # Maximum duty cycle # Minimum duty cycle - Maximum pulse width before pulse drops out # imit Maximum input signal - Delay to output # olifier Feedback voltage - Input bias current - Input offset voltage - Open loop voltage gain # Unity gain bandwidth # Out impedance # Output source current - Output sink current Output sink current	Output voltage	Output voltage Output impedance # 15 30 Short circuit current - 125 Change in V _{REF} with temperature # - 0.25 Oscillator frequency - 1.0 3.0 Initial accuracy¹ - 160 200 Voltage stability 170 Maximum duty cycle # 49.0 49.4 Minimum duty cycle 80 Maximum pulse width before pulse drops out # - 80 imit Maximum input signal - 1.0 1.2 Delay to output # - 80 Input bias current - 25 Input offset voltage - nulled during to Open loop voltage gain # 60 80 Unity gain bandwidth # 1.0 1.3 Out impedance # see Fig. 1 Output source current1.4 -2.0 Output source current - 0.12 0.15	Output voltage Output impedance # 15 30 45 Short circuit current - 125 250 Change in V _{REF} with temperature # - 0.25 - Oscillator frequency - 1.0 3.0 - Initial accuracy¹ - 160 200 240 Voltage stability 15 Temperature coefficient # - 170 - Maximum duty cycle # 49.0 49.4 49.6 Minimum duty cycle 0 Maximum pulse width before pulse drops out # - 80 125 imit Maximum input signal - 1.0 1.2 1.4 Delay to output # - 80 120 Diffier Feedback voltage - 3.92 4.00 4.08 Input bias current - 25 500 Input offset voltage - nulled during trim Open loop voltage gain # 60 80 - Unity gain bandwidth # 1.0 1.3 - Out impedance # see Fig. 1 Output source current1.4 -2.0 - Output source current1.4 -2.0 - Output source current1.4 -2.0 - Output source current1.4 -2.0 -	Output voltage - 3.92 4.00 4.08 V Output impedance # 15 30 45 KΩ Short circuit current - - 125 250 μA Change in V _{REF} with temperature # - 0.25 - mV/°C Oscillator frequency - 1.0 3.0 - MHz Initial accuracy¹ - 1.0 100 120 KHz Voltage stability - - 15 % Temperature coefficient # - 170 - ppm/°C Maximum duty cycle # 49.0 49.4 49.6 % Minimum duty cycle # 49.0 49.4 49.6 % Maximum pulse width before pulse drops out # - 80 125 ns Imit Maximum input signal - 1.0 1.2 1.4 V Delay to output # - 80 125 ns Si			

Notes:

- # Guaranteed by design.
- 1. Stray capacitance on OSC In pin must be ≤5pF.

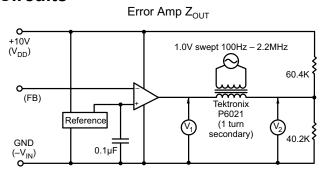
Electrical Characteristics (cont.) (Unless otherwise specified, V_{DD} = 10V, $+V_{IN}$ = 48V, R_{BIAS} = 390K Ω , R_{OSC} = 330K Ω , T_A = 25°C.)

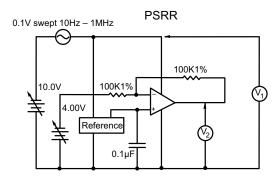
Sym	Parameter			Min	Тур	Max	Units	Conditions			
Pre-Regu	Pre-Regulator/Startup										
+V _{IN}	Input voltage			10	-	450	V	$I_{IN} < 10 \mu A; V_{CC} > 9.4 V$			
+1,,	Input leakage current			-	-	10	μA	V _{DD} > 9.4V			
V _{TH}	V _{DD} pre-regulator turn threshold voltage	n-off	-	8.0	8.7	9.4	V	I _{PREREG} = 10μA			
V _{LOCK}	Undervoltage lockout	t	-	7.0	8.1	8.9	V				
Supply											
I _{DD}	Supply current		-	-	0.75	1.3	mA	C _L < 75pF			
I _Q	Quiescent supply cur	rent	-	-	0.55	-	mA	SHUTDOWN = -V _{IN}			
I _{BIAS}	Nominal bias current		-	-	20	-	μA				
V _{DD}	Operating range		-	9.0	-	13.5	V				
Shutdow	n Logic										
t _{sd}	SHUTDOWN delay	#	-	50	100	ns	$C_L = 500pF, V_{SENSE} = -V_{IN}$				
t _{sw}	SHUTDOWN pulse width			50	-	-	ns				
t _{RW}	RESET pulse width			50	-	-	ns				
t _{LW}	Latching pulse width			25	-	-	ns	SHUTDOWN and RESET low			
V _{IL}	Input low voltage		-	-	-	2.0	V				
V _{IH}	Input high voltage		-	7.0	-	-	V				
I _{IH}	Input current, input h	igh voltage	-	-	1.0	5.0	μA	$V_{IN} = V_{DD}$			
I _{IL}	Input current, input lo	w voltage	-	-	-25	-35	μA	V _{IN} = 0V			
Output											
			_	V _{DD} -0.25	-	-	V	I _{OUT} = 10mA			
V _{OH}	Output high voltage		-	V _{DD} -0.3	-	-	V	I _{OUT} = 10mA, T _A = -55 to 125°C			
			_	-	-	0.2	V	I _{OUT} = -10mA			
V _{oL}	Output low voltage		-	-	-	0.3	V	I _{OUT} = -10mA, T _A = -55 to 125°C			
		Pull up	-	-	15	25	Ω				
P	Output resistance	Pull down	-	-	8.0	20	Ω	OUT - TIOIIIA			
OUT	R _{OUT} Output resistance	Pull up	-	-	20	30	Ω	I _{OUT} = ±10mA,			
			-	-	10	30	Ω	T _A = -55 to 125°C			
t _R	Rise time		#	-	30	75	ns	C _L = 500pF			
t _F	Fall time		#	-	20	75	ns	C _L = 500pF			
Note:											

Note:

[#] Guaranteed by design.

Test Circuits





Note:

Set feedback voltage so that $V_{COMP} = V_{DIVIDE} \pm 1.0 \text{mV}$ before connecting transformer.

Detailed Description Pre regulator

The pre regulator/startup circuit for the HV9120 consists of a high-voltage n-channel depletion-mode DMOS transistor driven by an error amplifier to form a variable current path between the VIN terminal and the VDD terminal. Maximum current (about 20 mA) occurs when $V_{\rm DD}=0$, with current reducing as $V_{\rm DD}$ rises. This path shuts off altogether when $V_{\rm DD}$ rises to somewhere between 7.8 and 9.4V, so that if $V_{\rm DD}$ is held at 10 or 12V by an external source (generally the supply the chip is controlling), no current other than leakage is drawn through the high voltage transistor. This minimizes dissipation.

An external capacitor between VDD and VSS is generally required to store energy used by the chip in the time between shutoff of the high voltage path and the VDD supply's output rising enough to take over powering the chip. This capacitor should have a value of 100X or more the effective gate capacitance of the MOSFET being driven, i.e.,

 $C_{STORAGE} \ge 100 \text{ x (gate charge of FET at 10V)}$

as well as very good high frequency characteristics. Stacked polyester or ceramic caps work well. Electrolytic capacitors are generally not suitable. A common resistor divider string is used to monitor $V_{\rm DD}$ for both the undervoltage lockout circuit and the shutoff circuit of the high voltage FET. Setting the undervoltage sense point about 0.6V lower on the string than the FET shutoff point guarantees that the undervoltage lockout always releases before the FET shuts off.

Bias Circuit

An external bias resistor, connected between the bias pin and VSS is required by the HV9120 to set currents in a series of current mirrors used by the analog sections of the chip. Nominal external bias current requirement is 15 to $20\mu A$, which can be set by a 390 to $510K\Omega$ resistor if a 10V

 V_{DD} is used, or a 510 to 680K Ω resistor if V_{DD} will be 12V. A precision resistor is not required; ±5% is fine.

Clock Oscillator

The clock oscillator of the HV9120 consists of a ring of CMOS inverters, timing capacitors, a capacitor discharge FET, and a frequency dividing flip-flop. A single external resistor between the OSC IN and OSC OUT pins is required to set oscillator frequency (see graph).

One difference exists between the Supertex HV9120 and competitive 9120s: The oscillator is shut off when a shutoff command is received. This saves about 150µA of quiescent current, which aids in the construction of power supplies to meet CCITT specification I-430, and in other situations where an absolute minimum of quiescent power dissipation is required.

Reference

The Reference of the HV9120 consists of a stable bandgap reference followed by a buffer amplifier which scales the voltage up to approximately 4.0V. The scaling resistors of the reference buffer amplifier are trimmed during manufacture so that the output of the error amplifier, when connected in a gain of -1 configuration, is as close to 4.0V as possible. This nulls out any input offset of the error amplifier. As a consequence, even though the observed reference voltage of a specific part may not be exactly 4.0V, the feedback voltage required for proper regulation will be.

A ≈ 50KΩ resistor is placed internally between the output of the reference buffer amplifier and the circuitry it feeds (reference output pin and non-inverting input to the error amplifier). This allows overriding the internal reference with a low-impedance voltage source ≤6.0V. Using an external reference reinstates the input offset voltage of the error am-

plifier, and its effect of the exact value of feedback voltage required. In general, because the reference voltage of the Supertex HV9120 is not noisy, as some previous examples have been, overriding the reference should seldom be necessary.

Because the reference of the HV9120 is a high impedance node, and usually there will be significant electrical noise near it, a bypass capacitor between the reference pin and VSS is strongly recommended. The reference buffer amplifier is intentionally compensated to be stable with a capacitive load of 0.01 to $0.1\mu F$.

Error Amplifier

The error amplifier in the HV9120 is a true low-power differential input operational amplifier intended for around-the-amplifier compensation. It is of mixed CMOS-bipolar construction: A PMOS input stage is used so the common-mode range includes ground and the input impedance is very high. This is followed by bipolar gain stages which provide high gain without the electrical noise of all-MOS amplifiers. The amplifier is unity-gain stable.

Current Sense Comparators

The HV9120 uses a true dual-comparator system with independent comparators for modulation and current limiting. This allows the designer greater latitude in compensation design, as there are no clamps (except ESD protection) on the compensation pin. Like the error amplifier, the comparators are of low-noise BiCMOS construction.

Remote Shutdown

The shutdown and reset pins of the HV9120 can be used to perform either latching or non-latching shutdown of a converter as required. These pins have internal current source pull-ups so they can be driven from open-drain logic. When not used, they should be left open, or connected to VDD.

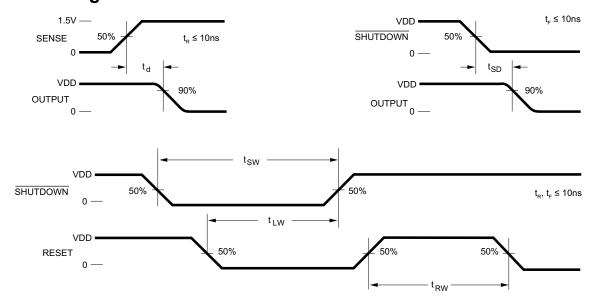
Output Buffer

The output buffer of the HV9120 is of standard CMOS construction (P-channel pull-up, N-channel pull-down). Thus the body-drain diodes of the output stage can be used for spike clipping if necessary, and external Schottky diode clamping of the output is not required.

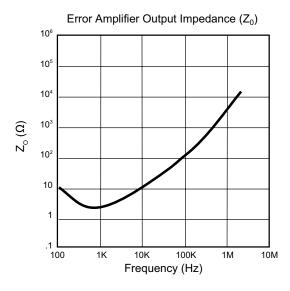
Truth Table

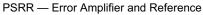
SHUTDOWN	RESET	Output
Н	Н	Normal operation
Н	H o L	Normal operation, no change
L	Н	Off, not latched
L	L	Off, latched
$L \rightarrow H$	L	Off, latched, no change

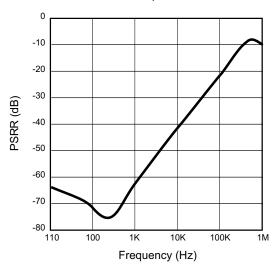
Shutdown Timing Waveforms

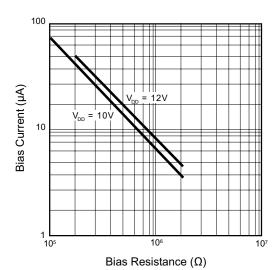


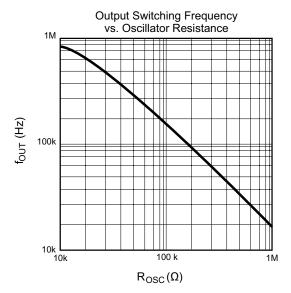
Typical Performance Curves

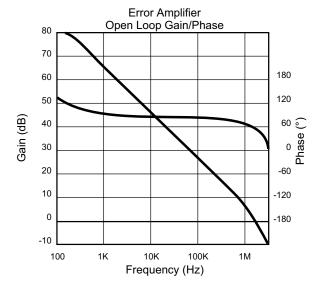












Pin Descriptions

16-Lead SOIC (NG)

Pin #	Description
1	+VIN
2	-
3	-
4	SENSE
5	OUTPUT
6	-VIN
7	VDD
8	OSC OUT

Pin#	Description
9	OSC IN
10	NC
11	VREF
12	SHUTDOWN
13	RESET
14	COMP
15	FB
16	BIAS

16-Lead PDIP (P)

Pin#	Description
1	+VIN
2	NC
3	NC
4	SENSE
5	OUTPUT
6	-VIN
7	VDD
8	OSC OUT

Pin#	Description
9	OSC IN
10	NC
11	VREF
12	SHUTDOWN
13	RESET
14	COMP
15	FB
16	BIAS

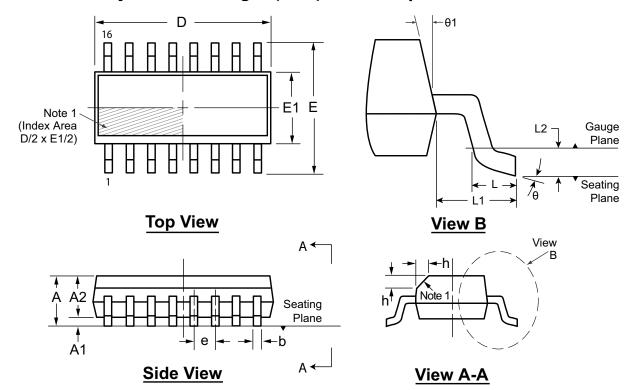
20-Lead PLCC (PJ)

Pin#	Description
1	NC
2	NC
3	+VIN
4	NC
5	SENSE
6	OUTPUT
7	NC
8	-VIN
9	VDD
10	OSC OUT

Pin#	Description
11	OSC IN
12	NC
13	NC
14	VREF
15	NC
16	SHUTDOWN
17	RESET
18	COMP
19	FB
20	BIAS

16-Lead SOIC (Narrow Body) Package Outline (NG)

9.90x3.90mm body, 1.75mm height (max), 1.27mm pitch



Note:

1. This chamfer feature is optional. If it is not present, then a Pin 1 identifier must be located in the index area indicated. The Pin 1 identifier can be: a molded mark/identifier; an embedded metal marker; or a printed indicator.

Symbo	ol	Α	A1	A2	b	D	E	E1	е	h	L	L1	L2	θ	θ1
	MIN	1.35*	0.10	1.25	0.31	9.80*	5.80*	3.80*		0.25	0.40			0 º	5 °
Dimension (mm)	NOM	-	-	-	-	9.90	6.00	3.90	1.27 BSC	-	-	1.04 REF	0.25 BSC	-	-
()	MAX	1.75	0.25	1.65*	0.51	10.00*	6.20*	4.00*		0.50	1.27			8 º	15°

JEDEC Registration MS-012, Variation AC, Issue E, Sept. 2005.

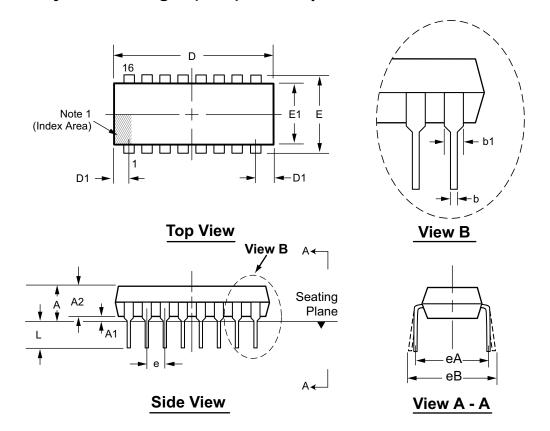
* This dimension is not specified in the original JEDEC drawing. The value listed is for reference only.

Drawings are not to scale.

Supertex Doc. #: DSPD-16SONG, Version F101708.

16-Lead PDIP (.300in Row Spacing) Package Outline (P)

.790x.250in body, .210in height (max), .100in pitch



Note:

 A Pin 1 identifier must be located in the index area indicated. The Pin 1 identifier can be: a molded mark/identifier; an embedded metal marker; or a printed indicator.

Symbol		Α	A1	A2	b	b1	D	D1	E	E1	е	eA	еВ	L
Dimension (inches)	MIN	.130*	.015	.115	.014	.045	.780	.005	.290 [†]	.240	.100 BSC	.300 BSC	.300*	.115
	NOM	-	-	.130	.018	.060	.790	-	.310	.250			-	.130
	MAX	.210	.035*	.195	.023 [†]	.070	.810 [†]	.050*	.325	.280			.430	.150

JEDEC Registration MS-001, Variation AB, Issue D, June, 1993.

Drawings not to scale.

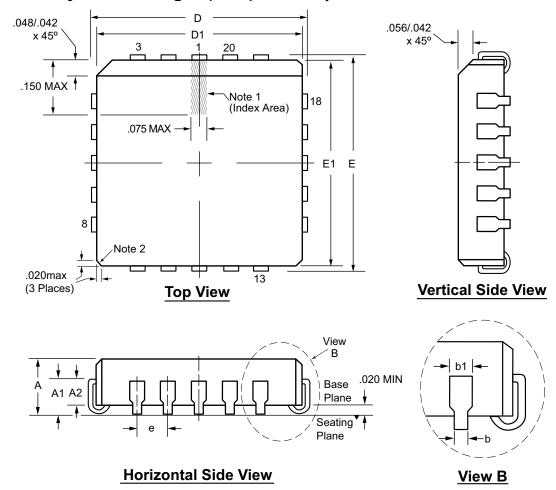
Supertex Doc. #: DSPD-16DIPP, Version A120108.

^{*} This dimension is not specified in the original JEDEC drawing. The value listed is for reference only.

[†] This dimension is a non-JEDEC dimension.

20-Lead PLCC Package Outline (PJ)

.353x.353in body, .180in height (max), .050in pitch



Notes:

- A Pin 1 identifier must be located in the index area indicated. The Pin 1 identifier can be: a molded mark/identifier; an embedded metal marker; or a printed indicator.
- 2. Actual shape of this feature may vary.

Symbol		Α	A1	A2	b	b1	D	D1	E	E1	е
Dimension (inches)	MIN	.165	.090	.062	.013	.026	.385	.350	.385	.350	.050 BSC
	NOM	.172	.105	-	-	-	.390	.353	.390	.353	
	MAX	.180	.120	.083	.021	.032	.395	.356	.395	.356	

JEDEC Registration MS-018, Variation AA, Issue A, June, 1993.

Drawings not to scale.

Supertex Doc. #: DSPD-20PLCCPJ, Version B092408

(The package drawing(s) in this data sheet may not reflect the most current specifications. For the latest package outline information go to http://www.supertex.com/packaging.html.)

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