

## High Speed Dual MOSFET Driver

### Features

- ▶ 6.0ns rise and fall time with 1000pF load
- ▶ 2.0A peak output source/sink current
- ▶ 1.8 to 5.0V input CMOS compatible
- ▶ 4.5 to 13V total supply voltage
- ▶ Smart logic threshold
- ▶ Low jitter design
- ▶ Two matched channels
- ▶ Outputs can swing below ground
- ▶ Low inductance package
- ▶ Thermally-enhanced package

### Applications

- ▶ Medical ultrasound imaging
- ▶ Piezoelectric transducer drivers
- ▶ Nondestructive evaluation
- ▶ PIN diode driver
- ▶ CCD clock driver/buffer
- ▶ High speed level translator

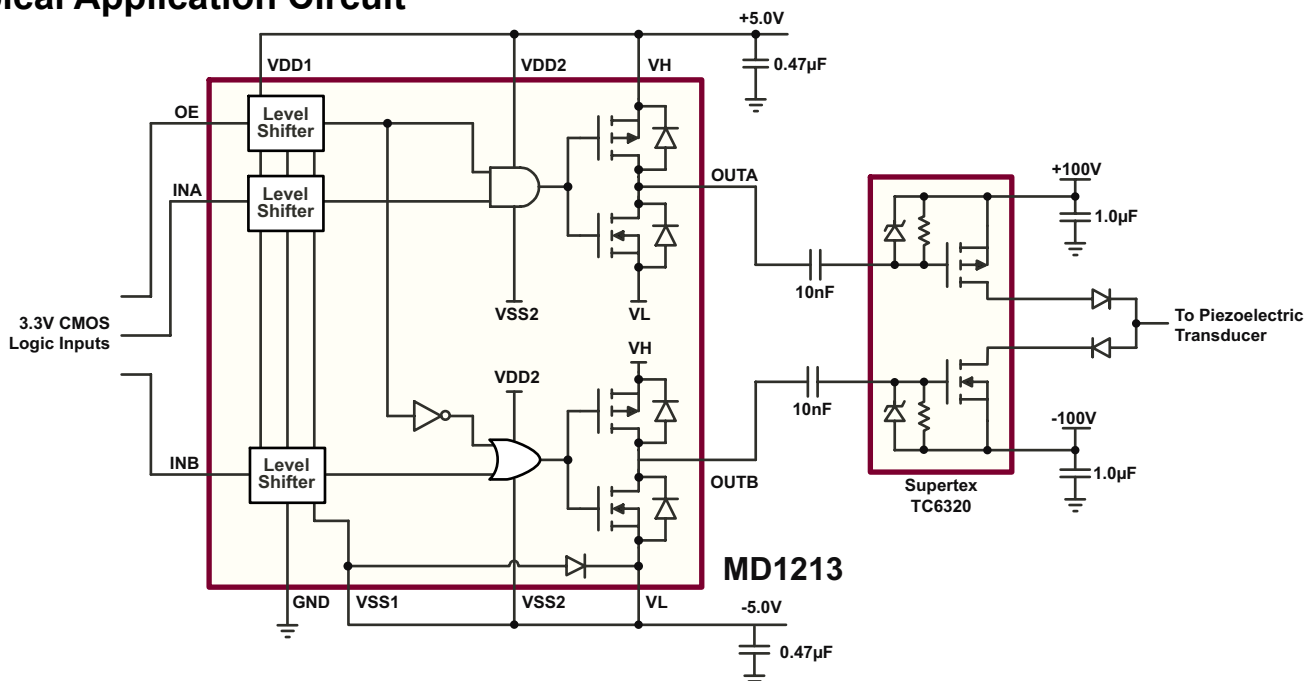
### General Description

The Supertex MD1213 is a high speed, dual MOSFET driver. It is designed to drive high voltage P and N-channel MOSFET transistors for medical ultrasound and other applications requiring a high output current for a capacitive load. The high-speed input stage of the MD1213 can operate from 1.8 to 5.0V logic interface with an optimum operating input signal range of 1.8 to 3.3V. An adaptive threshold circuit is used to set the level translator switch threshold to the average of the input logic 0 and logic 1 levels. The input logic levels may be ground referenced, even though the driver is putting out bipolar signals. The level translator uses a proprietary circuit, which provides DC coupling together with high-speed operation.

The output stage of the MD1213 has separate power connections enabling the output signal L and H levels to be chosen independently from the supply voltages used for the majority of the circuit. As an example, the input logic levels may be 0 and 1.8V, the control logic may be powered by +5.0 to -5.0V, and the output L and H levels may be varied anywhere over the range of -5.0 to +5.0V. The output stage is capable of peak currents of up to  $\pm 2.0A$ , depending on the supply voltages used and load capacitance present.

The OE pin serves a dual purpose. First, its logic H level is used to compute the threshold voltage level for the channel input level translators. Secondly, when OE is low, the outputs are disabled, with the A output high and the B output low. This assists in properly pre-charging the AC coupling capacitors that may be used in series in the gate drive circuit of an external PMOS and NMOS transistor pair.

### Typical Application Circuit



## Ordering Information

Device	12-Lead QFN 4.00x4.00mm body 1.00mm height (max) 0.80mm pitch
MD1213	MD1213K6-G

-G indicates package is RoHS compliant ("Green")



## Absolute Maximum Ratings

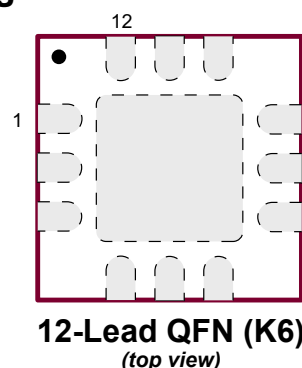
Parameter	Value
$V_{DD} - V_{SS}$ , logic supply voltage	-0.5V to +13.5V
$V_H$ , output high supply voltage	$V_L - 0.5V$ to $V_{DD} + 0.5V$
$V_L$ , output low supply voltage	$V_{SS} - 0.5V$ to $V_H + 0.5V$
$V_{SS}$ , low side supply voltage	-7.0V to +0.5V
Logic input levels	$V_{SS} - 0.5V$ to GND +7.0V
Maximum junction temperature	+125°C
Storage temperature	-65°C to 150°C
Operating temperature	-20°C to 85°C
Thermal resistance to air, $\theta_{JA}$	47°C/W
Thermal resistance to case, $\theta_{JC}$	7.0°C/W

Absolute Maximum Ratings are those values beyond which damage to the device may occur. Functional operation under these conditions is not implied. Continuous operation of the device at the absolute rating level may affect device reliability. All voltages are referenced to device ground.

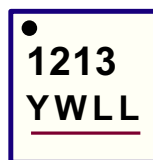
### Note:

- 1oz. 4-layer 3x4" PCB with thermal pad and thermal via array.

## Pin Configuration



## Package Marking



Y = Last Digit of Year Sealed  
W = Code for Week Sealed  
L = Lot Number  
— = "Green" Packaging

Package may or may not include the following marks: Si or

## 12-Lead QFN (K6)

## DC Electrical Characteristics

(Over operating conditions unless otherwise specified,  $V_H = V_{DD1} = V_{DD2} = 12V$ ,  $V_L = V_{SS1} = V_{SS2} = 0V$ ,  $V_{OE} = 3.3V$ ,  $T_A = 25^\circ C$ )

Sym	Parameter	Min	Typ	Max	Units	Conditions
$V_{DD} - V_{SS}$	Logic supply voltage	4.5	-	13	V	$2.5V \leq V_{DD} \leq 13V$
$V_{SS}$	Logic side supply voltage	-5.5	-	0	V	---
$V_H$	Output high supply voltage	$V_{SS} + 2.0$	-	$V_{DD}$	V	---
$V_L$	Output low supply voltage	$V_{SS}$	-	$V_{DD} - 2.0$	V	---
$I_{DD1Q}$	$V_{DD1}$ quiescent current	-	0.55	-	mA	No input transitions
$I_{DD2Q}$	$V_{DD2}$ quiescent current	-	-	10	$\mu A$	
$I_{HQ}$	$V_H$ quiescent current	-	-	10	$\mu A$	
$I_{DD1}$	$V_{DD1}$ average current	-	0.88	-	mA	One channel on at 5.0Mhz, No load
$I_{DD2}$	$V_{DD2}$ average current	-	6.6	-	mA	
$I_H$	$V_H$ average current	-	23	-	mA	
$V_{IH}$	Input logic voltage high	$V_{OE} - 0.3$	-	5.0	V	For logic inputs INA and INB
$V_{IL}$	Input logic voltage low	0	-	0.3	V	
$I_{IH}$	Input logic current high	-	-	1.0	$\mu A$	
$I_{IL}$	Input logic current low	-	-	1.0	$\mu A$	

**DC Electrical Characteristics (cont.)**(Over operating conditions unless otherwise specified,  $V_H = V_{DD1} = V_{DD2} = 12V$ ,  $V_L = V_{SS1} = V_{SS2} = 0V$ ,  $V_{OE} = 3.3V$ ,  $T_A = 25^\circ C$ )

Sym	Parameter	Min	Typ	Max	Units	Conditions
$V_{IH}$	OE input logic voltage high	1.8	-	5.0	V	For logic input OE
$V_{IL}$	OE input logic voltage low	0	-	0.3	V	
$R_{IN}$	OE input logic impedance to GND	12	20	30	K $\Omega$	
$C_{IN}$	Logic input capacitance	-	5.0	10	pF	All inputs

**Outputs** ( $V_H = V_{DD1} = V_{DD2} = 12V$ ,  $V_L = V_{SS1} = V_{SS2} = 0V$ ,  $V_{OE} = 3.3V$ ,  $T_A = 25^\circ C$ )

$R_{SINK}$	Output sink resistance	-	-	12.5	$\Omega$	$I_{SINK} = 50mA$
$R_{SOURCE}$	Output source resistance	-	-	12.5	$\Omega$	$I_{SOURCE} = 50mA$
$I_{SINK}$	Peak output sink current	-	2.0	-	A	---
$I_{SOURCE}$	Peak output source current	-	2.0	-	A	---

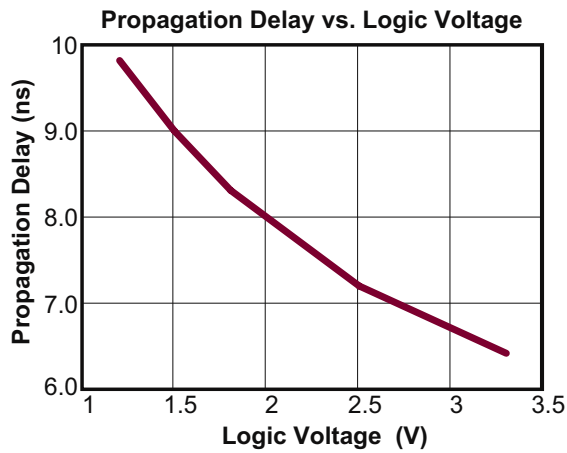
**AC Electrical Characteristics** ( $V_H = V_{DD1} = V_{DD2} = 12V$ ,  $V_L = V_{SS1} = V_{SS2} = 0V$ ,  $V_{OE} = 3.3V$ ,  $T_A = 25^\circ C$ )

Sym	Parameter	Min	Typ	Max	Units	Conditions
$t_{iff}$	Inputs or OE rise & fall time	-	-	10	ns	Logic input edge speed requirement
$t_{PLH}$	Propagation delay when output is from low to high	-	7.0	-	ns	$C_{LOAD} = 1000pF$ , see timing diagram Input signal rise/fall time of 2ns
$t_{PHL}$	Propagation delay when output is from high to low	-	7.0	-	ns	
$t_{POE}$	Propagation delay OE to outputs	-	9.0	-	ns	
$t_r$	Output rise time	-	6.0	-	ns	
$t_f$	Output fall time	-	6.0	-	ns	
$ t_r - t_f $	Rise and fall time matching	-	1.0	-	ns	
$ t_{PLH} - t_{PHL} $	Propagation low to high and high to low matching	-	1.0	-	ns	
$\Delta t_{dm}$	Propagation delay match	-	$\pm 2.0$	-	ns	Device to device delay match

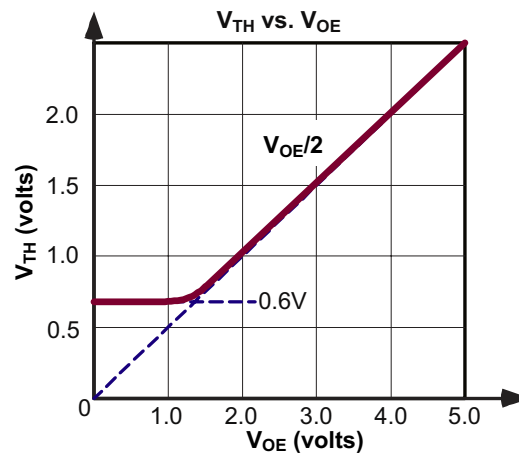
**Logic Truth Table**

Logic Inputs			Output	
OE	INA	INB	OUTA	OUTB
H	L	L	$V_H$	$V_H$
H	L	H	$V_H$	$V_L$
H	H	L	$V_L$	$V_H$
H	H	H	$V_L$	$V_L$
L	X	X	$V_H$	$V_L$

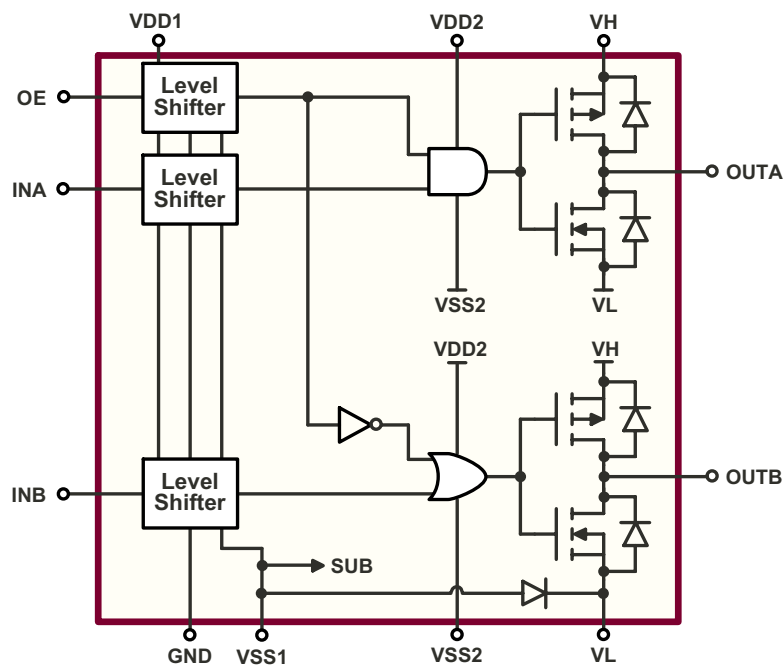
## Propagation Delay



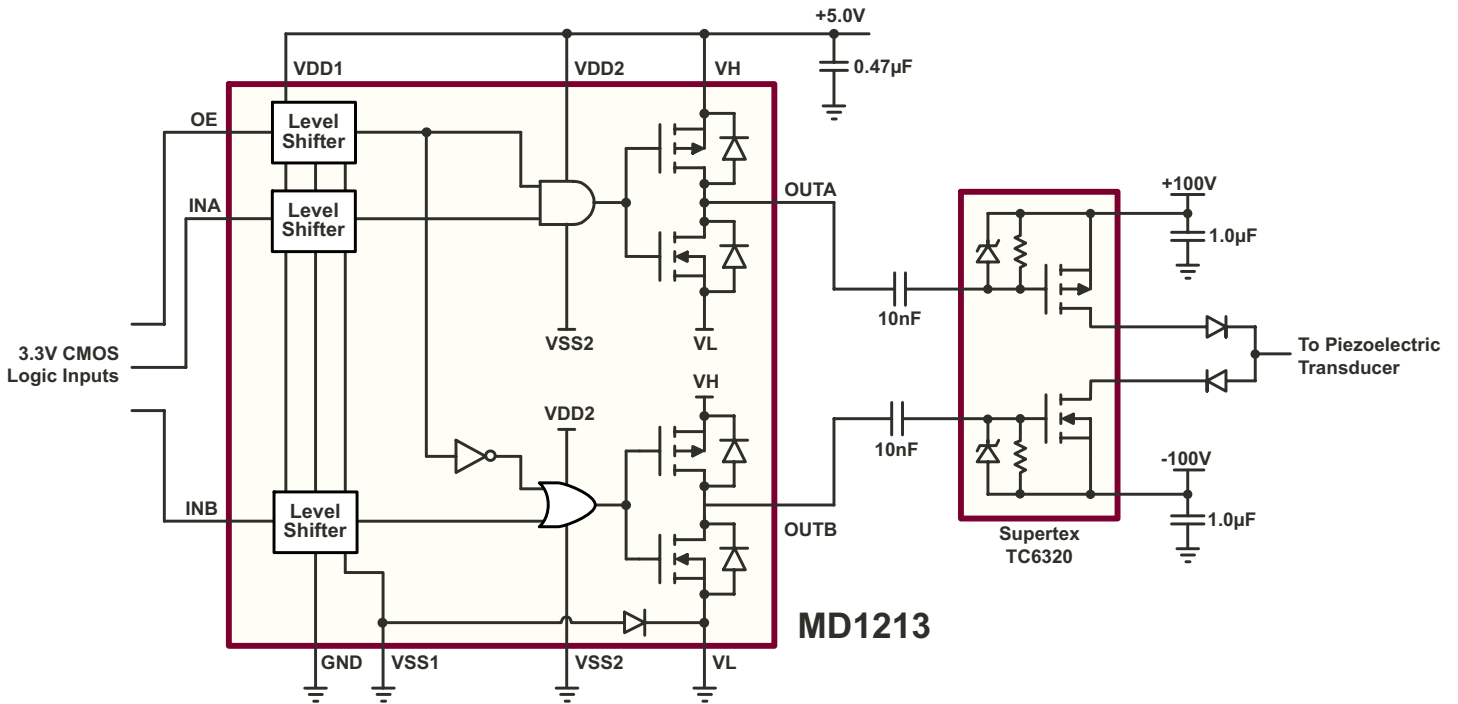
## Logic Input Threshold



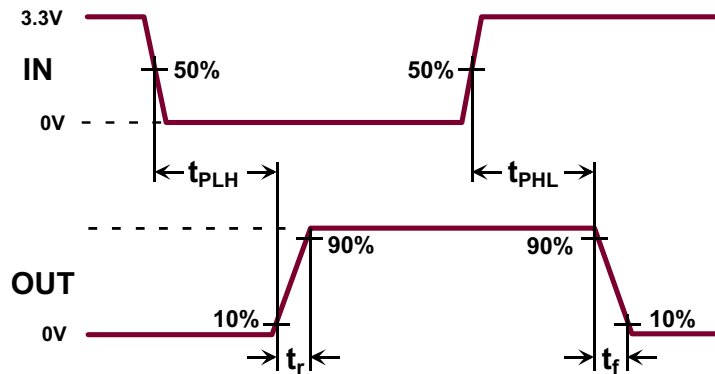
## Detailed Block Diagram



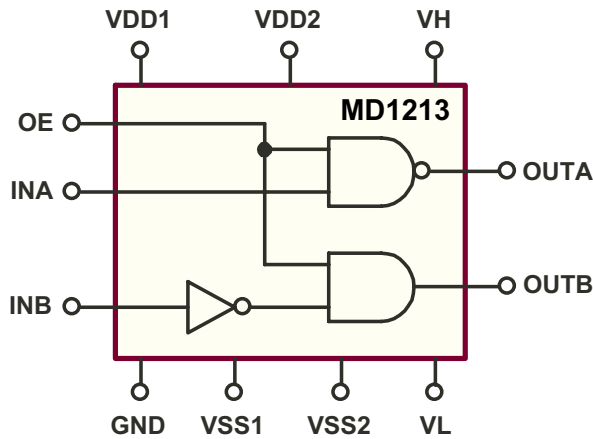
### Single Supply Application Circuit



### Timing Diagram



### Simplified Block Diagram



## Application Information

For proper operation of the MD1213, low inductance bypass capacitors should be used on the various supply pins. The GND input pin should be connected to the digital ground. The INA, INB, and OE pins should be connected to their logic source with a swing of GND to logic level high, which is 1.8 to 5.0V. Good trace practices should be followed corresponding to the desired operating speed. The internal circuitry of the MD1213 is capable of operating up to 100MHz, with the primary speed limitation being the loading effects of the load capacitance. Because of this speed and the high transient currents that result with capacitive loads, the bypass capacitors should be as close to the chip pins as possible. Unless the load specifically requires bipolar drive, the VSS1, VSS2, and VL pins should have low inductance feed-through connections directly to a ground plane. If these voltages are not zero, then they need bypass capacitors in a manner similar to the positive power supplies. The power connections VDD1 and VDD2 should have a ceramic bypass capacitor to the ground plane with short leads and decoupling components to prevent resonance in the power leads. A common capacitor and voltage source may be used for these two pins, which should always have the same DC voltage applied. For applications sensitive to jitter and noise, separate decoupling networks may be used for VDD1 and VDD2.

## Pin Description

Pin #	Name	Description
1	INA	Logic input. Controls OUTA when OE is high. Input logic high will cause the output to swing to VL. Input logic low will cause the output to swing to VH.
2	VL	Supply voltage for N-channel output stage.
3	INB	Logic input. Controls OUTB when OE is high. Input logic high will cause the output to swing to VL. Input logic low will cause the output to swing to VH.
4	GND	Logic input ground reference.
5	VSS1	Low side analog circuit and level shifter supply voltage. Should be at the same potential as VSS2.
6	VSS2	Low side gate drive supply voltage.
7	OUTB	Output driver. Swings from VH to VL. Intended to drive the gate of an external N-channel MOSFET via a series capacitor. When OE is low, the output is disabled. OUTB will swing to VL turning off the external N-channel MOSFET.
8	VH	Supply voltage for P-channel output stage.
9	OUTA	Output driver. Swings from VH to VL. Intended to drive the gate of an external P-channel MOSFET via a series capacitor. When OE is low, the output is disabled. OUTA will swing to VH turning off the external P-channel MOSFET.
10	VDD2	High side gate drive supply voltage.
11	VDD1	High side analog circuit and level shifter supply voltage. Should be at the same potential as VDD2.
12	OE	Output-enable logic input. When OE is high, $(V_{OE} + V_{GND})/2$ sets the threshold transition between logic level high and low for INA and INB. When OE is low, OUTA is at VH and OUTB is at VL regardless of INA and INB

### Notes:

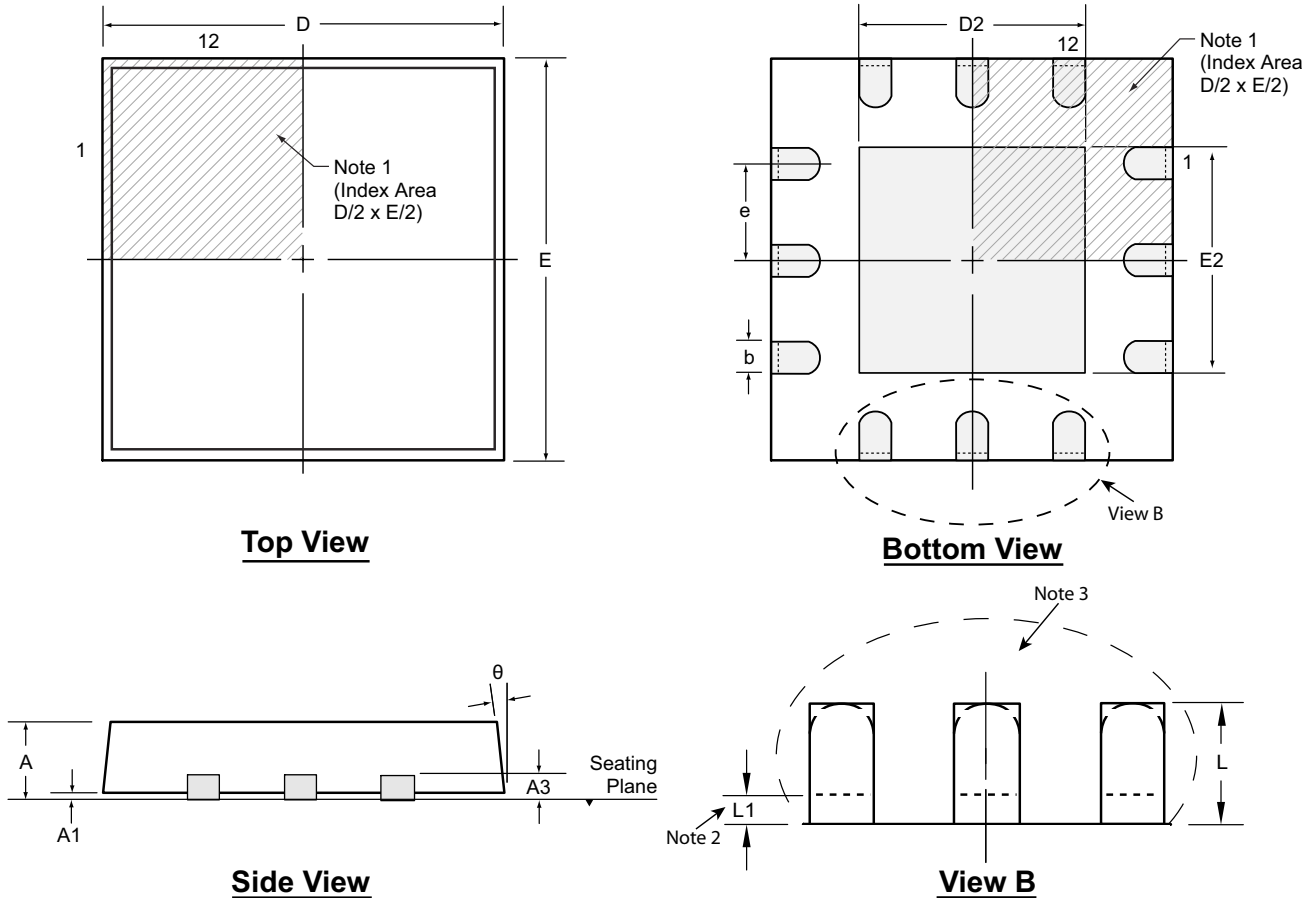
1. Thermal Pad and Pin #5 (VSS1) must be connected externally.
2. Index Pad and Thermal Pad are connected internally

The supplied voltages of VH and VL determine the output logic levels. These two pins can draw fast transient currents of up to 2.0A, so they should be provided with an appropriate bypass capacitor located next to the chip pins. A ceramic capacitor of up to 1.0 $\mu$ F may be appropriate, with a series ferrite bead to prevent resonance in the power supply lead coming to the capacitor. Pay particular attention to minimizing trace lengths and using sufficient trace width to reduce inductance. Surface mount components are highly recommended. Since the output impedance of this driver is very low, in some cases it may be desirable to add a small series resistor in series with the output signal to obtain better waveform integrity at the load terminals.

This will of course reduce the output voltage slew rate at the terminals of a capacitive load. Pay particular attention to the parasitic coupling from the driver output to the input signal terminals. This feedback may cause oscillations or spurious waveform shapes on the edges of signal transitions. Since the input operates with signals down to 1.8V, even small coupled voltages may cause problems. Use of a solid ground plane and good power and signal layout practices will prevent this problem. Be careful that the circulating ground return current from a capacitive load cannot react with common inductance to cause noise voltages in the input logic circuitry.

# 12-Lead QFN Package Outline (K6)

4.00x4.00mm body, 1.00mm height (max), 0.80mm pitch



- Notes:**
1. A Pin 1 identifier must be located in the index area indicated. The Pin 1 identifier can be: a molded mark/identifier; an embedded metal marker; or a printed indicator.
  2. Depending on the method of manufacturing, a maximum of 0.15mm pullback (L1) may be present.
  3. The inner tip of the lead may be either rounded or square.

Symbol	A	A1	A3	b	D	D2	E	E2	e	L	L1	θ	
Dimension (mm)	MIN	0.80	0.00	0.20 REF	0.25	3.85*	0.75	3.85*	0.75	0.80 BSC	0.35	0.00	0°
	NOM	0.90	0.02		0.30	4.00	1.70	4.00	1.70		0.55	-	-
	MAX	1.00	0.05		0.35	4.15*	2.25	4.15*	2.25		0.75	0.15	14°

JEDEC Registration MO-220, Variation VGGB, Issue K, June 2006.

\* This dimension is not specified in the JEDEC drawing.

**Drawings not to scale.**

**Supertex Doc. #:** DSPD-12QFNK64X4P080, Version C041309.

(The package drawing(s) in this data sheet may not reflect the most current specifications. For the latest package outline information go to <http://www.supertex.com/packaging.html>.)

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