

### Features

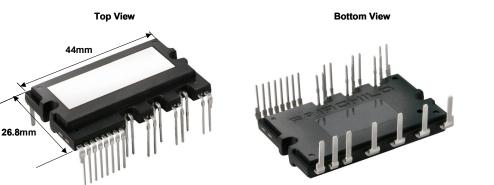
- UL Certified No.E209204(SPM27-BA package)
- 600V-5A 3-phase IGBT inverter bridge including control ICs for gate driving and protection
- Divided negative dc-link terminals for inverter current sensing applications
- · Single-grounded power supply due to built-in HVIC
- Isolation rating of 2500Vrms/min.
- Very low leakage current due to using ceramic substrate

### Applications

- AC 100V ~ 253V three-phase inverter drive for small power ac motor drives
- Home appliances applications like air conditioner and washing machine.

### **General Description**

It is an advanced smart power module (SPM<sup>TM</sup>) that Fairchild has newly developed and designed to provide very compact and high performance ac motor drives mainly targeting lowpower inverter-driven application like air conditioner and washing machine. It combines optimized circuit protection and drive matched to low-loss IGBTs. System reliability is further enhanced by the integrated under-voltage lock-out and shortcircuit protection. The high speed built-in HVIC provides optocoupler-less single-supply IGBT gate driving capability that further reduce the overall size of the inverter system design. Each phase current of inverter can be monitored separately due to the divided negative dc terminals.





April 2005

**SPM**<sup>™</sup>

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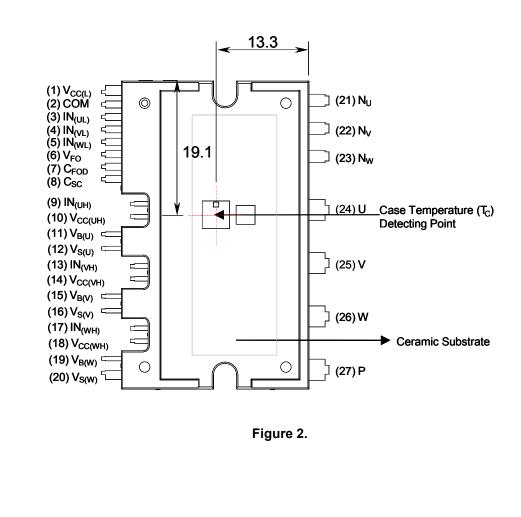
# **Integrated Power Functions**

• 600V-5A IGBT inverter for three-phase DC/AC power conversion (Please refer to Figure 3)

# Integrated Drive, Protection and System Control Functions

- For inverter high-side IGBTs: Gate drive circuit, High voltage isolated high-speed level shifting
  Control circuit under-voltage (UV) protection
  Note) Available bootstrap circuit example is given in Figures 10 and 11.
- For inverter low-side IGBTs: Gate drive circuit, Short circuit protection (SC) Control supply circuit under-voltage (UV) protection
- Fault signaling: Corresponding to a UV fault (Low-side supply)
- · Input interface: 3.3/5V CMOS/LSTTL compatible, Schmitt trigger inpu

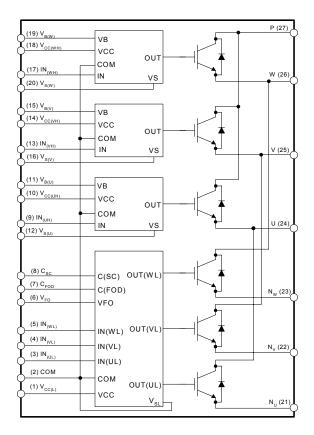
# **Pin Configuration**



### **Top View**

Pin Number	Pin Name	Pin Description	Pin Description		
1	V <sub>CC(L)</sub>	Low-side Common Bias Voltage for IC and IGBTs Driving			
2	СОМ	Common Supply Ground			
3	IN <sub>(UL)</sub>	Signal Input for Low-side U Phase			
4	IN <sub>(VL)</sub>	Signal Input for Low-side V Phase			
5	IN <sub>(WL)</sub>	Signal Input for Low-side W Phase			
6	V <sub>FO</sub>	Fault Output			
7	C <sub>FOD</sub>	Capacitor for Fault Output Duration Time Selection			
8	C <sub>SC</sub>	Capacitor (Low-pass Filter) for Short-Current Detection Input			
9	IN <sub>(UH)</sub>	Signal Input for High-side U Phase			
10	V <sub>CC(UH)</sub>	High-side Bias Voltage for U Phase IC			
11	V <sub>B(U)</sub>	High-side Bias Voltage for U Phase IGBT Driving			
12	V <sub>S(U)</sub>	High-side Bias Voltage Ground for U Phase IGBT Driving			
13	IN <sub>(VH)</sub>	Signal Input for High-side V Phase			
14	V <sub>CC(VH)</sub>	High-side Bias Voltage for V Phase IC			
15	V <sub>B(V)</sub>	High-side Bias Voltage for V Phase IGBT Driving			
16	V <sub>S(V)</sub>	High-side Bias Voltage Ground for V Phase IGBT Driving			
17	IN <sub>(WH)</sub>	Signal Input for High-side W Phase			
18	V <sub>CC(WH)</sub>	High-side Bias Voltage for W Phase IC			
19	V <sub>B(W)</sub>	High-side Bias Voltage for W Phase IGBT Driving			
20	V <sub>S(W)</sub>	High-side Bias Voltage Ground for W Phase IGBT Driving			
21	NU	Negative DC-Link Input for U Phase			
22	N <sub>V</sub>	Negative DC-Link Input for V Phase			
23	N <sub>W</sub>	Negative DC-Link Input for W Phase			
24	U	Output for U Phase			
25	V	Output for V Phase			
26	W	Output for W Phase			
27	Р	Positive DC–Link Input			

# Internal Equivalent Circuit and Input/Output Pins



#### Note:

1. Inverter low-side is composed of three IGBTs, freewheeling diodes for each IGBT and one control IC. It has gate drive and protection functions.

2. Inverter power side is composed of four inverter dc-link input terminals and three inverter output terminals.

3. Inverter high-side is composed of three IGBTs, freewheeling diodes and three drive ICs for each IGBT.

### Figure 3.

# Absolute Maximum Ratings (T<sub>J</sub> = 25°C, Unless Otherwise Specified)

### **Inverter Part**

Symbol	Parameter	Parameter Conditions		Unit
V <sub>PN</sub>	Supply Voltage	Applied between P- $N_U$ , $N_V$ , $N_W$	450	V
V <sub>PN(Surge)</sub>	Supply Voltage (Surge)	Applied between P- $N_U$ , $N_V$ , $N_W$	500	V
V <sub>CES</sub>	Collector-emitter Voltage		600	V
± I <sub>C</sub>	Each IGBT Collector Current	$T_{\rm C} = 25^{\circ}{\rm C}$	5	A
± I <sub>CP</sub>	Each IGBT Collector Current (Peak)	$T_{C}$ = 25°C, Under 1ms Pulse Width	10	A
P <sub>C</sub>	Collector Dissipation	T <sub>C</sub> = 25°C per One Chip	25	W
TJ	Operating Junction Temperature	(Note 1)	-20 ~ 125	°C

Note:

The maximum junction temperature rating of the power chips integrated within the SPM is 150 °C(@T<sub>C</sub> ≤ 100°C). However, to insure safe operation of the SPM, the average junction temperature should be limited to T<sub>J(ave)</sub> ≤ 125°C (@T<sub>C</sub> ≤ 100°C)

### **Control Part**

Symbol	Parameter Conditions		Rating	Units
V <sub>CC</sub>	Control Supply Voltage	Applied between V <sub>CC(UH)</sub> , V <sub>CC(VH)</sub> , V <sub>CC(WH)</sub> , V <sub>CC(L)</sub> - COM	20	V
V <sub>BS</sub>	High-side Control Bias Voltage	Applied between V <sub>B(U)</sub> - V <sub>S(U)</sub> , V <sub>B(V)</sub> - V <sub>S(V)</sub> , V <sub>B(W)</sub> - V <sub>S(W)</sub>	20	V
V <sub>IN</sub>	Input Signal Voltage	Applied between IN <sub>(UH)</sub> , IN <sub>(VH)</sub> , IN <sub>(WH)</sub> , IN <sub>(UL)</sub> , IN <sub>(VL)</sub> , IN <sub>(VL)</sub> , IN <sub>(WL)</sub> ,	-0.3~17	V
V <sub>FO</sub>	Fault Output Supply Voltage	Applied between V <sub>FO</sub> - COM	-0.3~V <sub>CC</sub> +0.3	V
I <sub>FO</sub>	Fault Output Current	Sink Current at V <sub>FO</sub> Pin	5	mA
V <sub>SC</sub>	Current Sensing Input Voltage	Applied between C <sub>SC</sub> - COM	-0.3~V <sub>CC</sub> +0.3	V

### **Total System**

Symbol	Parameter	Condition	Rating	Unit
V <sub>PN(PROT)</sub>	Self Protection Supply Voltage Limit (Short Circuit Protection Capability)	$V_{CC} = V_{BS} = 13.5 \sim 16.5 V$ T <sub>J</sub> = 125°C, Non-repetitive, less than 2µs	400	V
T <sub>C</sub>	Module Case Operation Temperature	-20°C $\leq$ T_J $\leq$ 125°C, See Figure 2	-20 ~ 100	°C
T <sub>STG</sub>	Storage Temperature		-40 ~ 125	°C
V <sub>ISO</sub>	Isolation Voltage	60Hz, Sinusoidal, AC 1 minute, Connection Pins to ceramic substrate	2500	V <sub>rms</sub>

# **Thermal Resistance**

Symbol	Parameter	Conditions	Min.	Тур.	Max.	Units
R <sub>th(j-c)Q</sub>		Inverter IGBT part (per 1/6 module)	-	-	3.9	°C/W
R <sub>th(j-c)F</sub>	Resistance	Inverter FWD part (per 1/6 module)	-	-	5.3	°C/W

Note:

2. For the measurement point of case temperature(T $_{C}$ ), please refer to Figure 2.

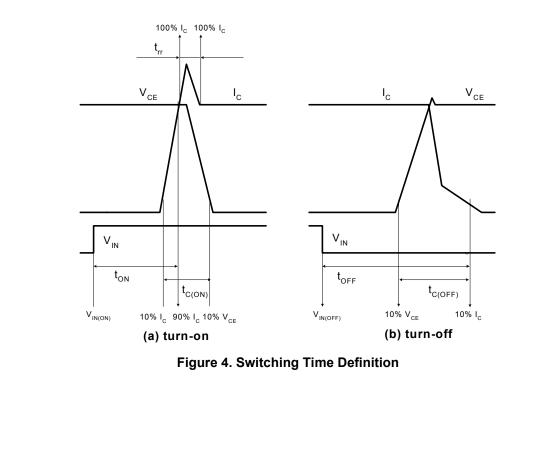
# Electrical Characteristics (T<sub>J</sub> = 25°C, Unless Otherwise Specified)

### Inverter Part

S	ymbol	Parameter	Cond	itions	Min.	Тур.	Max.	Units
V	CE(SAT)	Collector-Emitter Saturation Voltage	$V_{CC} = V_{BS} = 15V$ $V_{IN} = 5V$			-	2.3	V
	V <sub>F</sub>	FWD Forward Voltage	V <sub>IN</sub> = 0V	$V_{IN} = 0V$ $I_C = 5A, T_J = 25^{\circ}C$		-	2.1	V
HS	t <sub>ON</sub>	Switching Times		V <sub>PN</sub> = 300V, V <sub>CC</sub> = V <sub>BS</sub> = 15V		0.40	-	μs
	t <sub>C(ON)</sub>		$I_{\rm C} = 5A$	ivo Lood	-	0.21	-	μs
	t <sub>OFF</sub>		$V_{IN} = 0V \leftrightarrow 5V$ , Inductive Load (Note 3)		-	0.68	-	μs
	t <sub>C(OFF)</sub>				-	0.25	-	μs
	t <sub>rr</sub>				-	0.10	-	μs
LS	t <sub>ON</sub>		V <sub>PN</sub> = 300V, V <sub>CC</sub> = V <sub>BS</sub> = 15V		-	0.56	-	μs
	t <sub>C(ON)</sub>		$I_{C} = 5A$ $V_{IN} = 0V \leftrightarrow 5V$ , Induct	ive Load	-	0.28	-	μs
	t <sub>OFF</sub>		(Note 3)		-	0.72	-	μs
	t <sub>C(OFF)</sub>				-	0.25	-	μs
	t <sub>rr</sub>				-	0.10	-	μs
	I <sub>CES</sub>	Collector-Emitter Leakage Current	$V_{CE} = V_{CES}$		-	-	250	μΑ

#### Note:

3.  $t_{ON}$  and  $t_{OFF}$  include the propagation delay time of the internal drive IC.  $t_{C(ON)}$  and  $t_{C(OFF)}$  are the switching time of IGBT itself under the given gate driving condition internally. For the detailed information, please see Figure 4.



# Electrical Characteristics (T<sub>J</sub> = 25°C, Unless Otherwise Specified)

### Control Part

Symbol	Parameter	Co	nditions	Min.	Тур.	Max.	Units
IQCCL	Quiescent V <sub>CC</sub> Supply Current	V <sub>CC</sub> = 15V IN <sub>(UL, VL, WL)</sub> = 0V	V <sub>CC(L)</sub> - COM	-	-	23	mA
I <sub>QCCH</sub>		V <sub>CC</sub> = 15V IN <sub>(UH, VH, WH)</sub> = 0V	$\begin{array}{l} V_{CC(UH)},  V_{CC(VH)}, \\ V_{CC(WH)} - COM \end{array}$	-	-	100	μΑ
I <sub>QBS</sub>	Quiescent V <sub>BS</sub> Supply Current	V <sub>BS</sub> = 15V IN <sub>(UH, VH, WH)</sub> = 0V	$ \begin{array}{l} V_{B(U)} \mbox{-} V_{S(U)},  V_{B(V)} \mbox{-} V_{S(V)}, \\ V_{B(W)} \mbox{-} V_{S(W)} \end{array} $	-	-	500	μA
V <sub>FOH</sub>	Fault Output Voltage	$V_{SC}$ = 0V, $V_{FO}$ Circu	$V_{SC}$ = 0V, $V_{FO}$ Circuit: 4.7k $\Omega$ to 5V Pull-up		-	-	V
V <sub>FOL</sub>		$V_{SC}$ = 1V, $V_{FO}$ Circuit: 4.7k $\Omega$ to 5V Pull-up		-	-	0.8	V
V <sub>SC(ref)</sub>	Short Circuit Trip Level	V <sub>CC</sub> = 15V (Note 4)		0.45	0.5	0.55	V
UV <sub>CCD</sub>	Supply Circuit Under-	Detection Level		10.7	11.9	13.0	V
UV <sub>CCR</sub>	Voltage Protection	Reset Level	Reset Level		12.4	13.2	V
UV <sub>BSD</sub>		Detection Level		10.1	11.3	12.5	V
UV <sub>BSR</sub>		Reset Level		10.5	11.7	12.9	V
t <sub>FOD</sub>	Fault-out Pulse Width	C <sub>FOD</sub> = 33nF (Note 5)		1.0	1.8	-	ms
V <sub>IN(ON)</sub>	ON Threshold Voltage	Applied between $IN_{(UH)}$ , $IN_{(VH)}$ , $IN_{(WH)}$ , $IN_{(UL)}$ ,		3.0	-	-	V
V <sub>IN(OFF)</sub>	OFF Threshold Voltage	IN <sub>(VL)</sub> , IN <sub>(WL)</sub> - COM		-	-	0.8	V

Note:

4. Short-circuit current protection is functioning only at the low-sides.

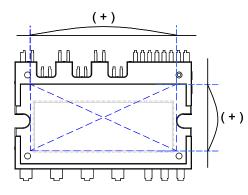
5. The fault-out pulse width  $t_{FOD}$  depends on the capacitance value of  $C_{FOD}$  according to the following approximate equation :  $C_{FOD}$  = 18.3 x 10<sup>-6</sup> x  $t_{FOD}[F]$ 

# **Recommended Operating Conditions**

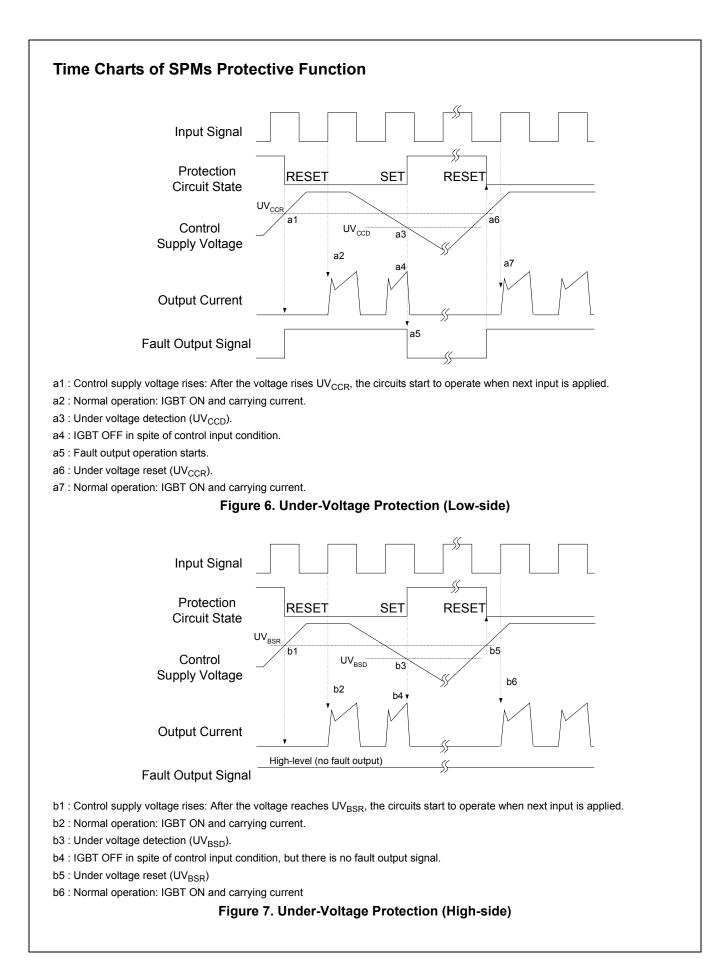
Symbol	Parameter	Conditions		Value		Units	
Symbol	Farameter	Conditions	Min.	Тур.	Max.	Units	
V <sub>PN</sub>	Supply Voltage	Applied between P - N <sub>U</sub> , N <sub>V</sub> , N <sub>W</sub>	-	300	400	V	
V <sub>CC</sub>	Control Supply Voltage	Applied between $V_{CC(UH)}$ , $V_{CC(VH)}$ , $V_{CC(WH)}$ , $V_{CC(WH)}$ , $V_{CC(WH)}$ ,		15	16.5	V	
V <sub>BS</sub>	High-side Bias Voltage	Applied between $V_{B(U)} - V_{S(U)}$ , $V_{B(V)} - V_{S(V)}$ , $V_{B(W)} - V_{S(W)}$		15	18.5	V	
DV <sub>CC</sub> /Dt, DV <sub>BS</sub> /Dt	Control supply variation			-	1	V/μs	
t <sub>dead</sub>	Blanking Time for Preventing Arm-short	For Each Input Signal		-	-	μs	
f <sub>PWM</sub>	PWM Input Signal	$-20^{\circ}C \leq T_C \leq 100^{\circ}C, \ -20^{\circ}C \leq T_J \leq 125^{\circ}C$	-	-	20	kHz	
$V_{SEN}$	Voltage for Current Sensing	Applied between $N_U$ , $N_V$ , $N_W$ - COM (Including surge voltage)	-4		4	V	

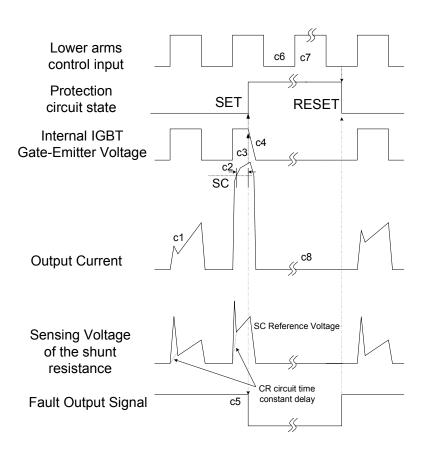
# **Mechanical Characteristics and Ratings**

Parameter	C	ondition		Limits		Units
Farameter		manion	Min.	Тур.	Max.	Units
Mounting Torque	Mounting Screw: - M3	Recommended 0.62N•m	0.51	0.62	0.72	N•m
Device Flatness		Note Figure 5	0	-	+120	μm
Weight			-	15.4	-	g









(with the external shunt resistance and CR connection)

c1 : Normal operation: IGBT ON and carrying current.

c2 : Short circuit current detection (SC trigger).

c3 : Hard IGBT gate interrupt.

c4 : IGBT turns OFF.

c5 : Fault output timer operation starts: The pulse width of the fault output signal is set by the external capacitor  $C_{FO}$ .

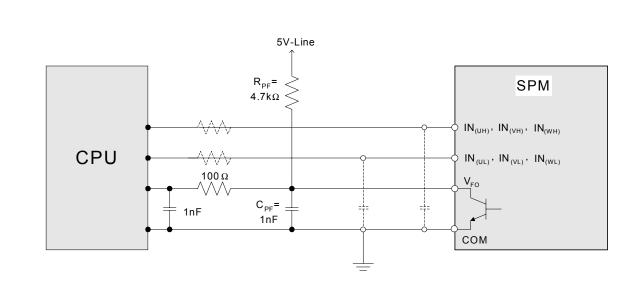
c6 : Input "L" : IGBT OFF state.

c7 : Input "H": IGBT ON state, but during the active period of fault output the IGBT doesn't turn ON.

c8 : IGBT OFF state

### Figure 8. Short-Circuit Current Protection (Low-side Operation only)

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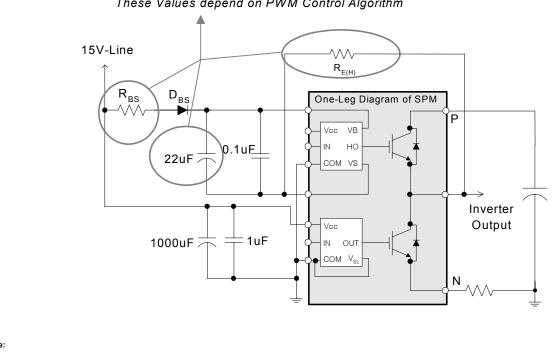


#### Note:

1. RC coupling at each input (parts shown dotted) might change depending on the PWM control scheme used in the application and the wiring impedance of the application's printed circuit board. The SPM input signal section integrates 3.3kΩ (typ.) pull-down resistor. Therefore, when using an external filtering resistor, please pay attention to the signal voltage drop at input terminal.

2. The logic input is compatible with standard CMOS or LSTTL outputs.

### Figure 9. Recommended CPU I/O Interface Circuit

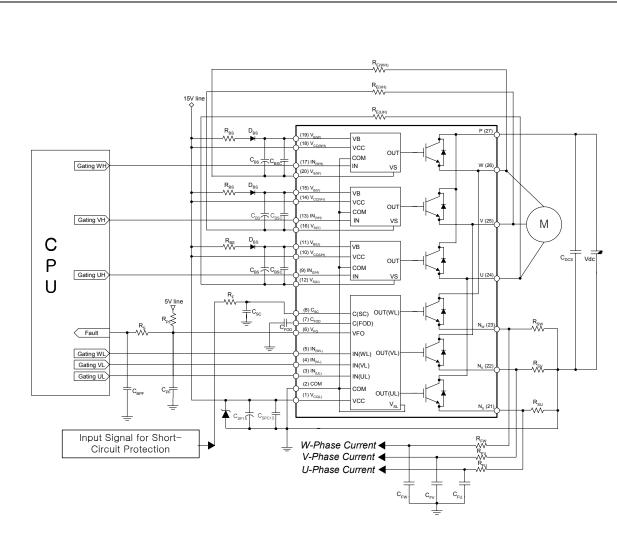


These Values depend on PWM Control Algorithm

Note:

- 1. It would be recommended that the bootstrap diode,  $\mathsf{D}_{\mathsf{BS}}$ , has soft and fast recovery characteristics.
- 2. The bootstrap resistor (R<sub>BS</sub>) should be 3 times greater than R<sub>E(H)</sub>. The recommended value of R<sub>E(H)</sub> is 5.6 $\Omega$ , but it can be increased up to 20 $\Omega$  (maximum) for a slower dv/dt of high-side
- 3. The ceramic capacitor placed between V<sub>CC</sub>-COM should be over 1uF and mounted as close to the pins of the SPM as possible.

### Figure 10. Recommended Bootstrap Operation Circuit and Parameters



#### Note:

1. To avoid malfunction, the wiring of each input should be as short as possible. (less than 2-3cm)

2. By virtue of integrating an application specific type HVIC inside the SPM, direct coupling to CPU terminals without any opto-coupler or transformer isolation is possible.

V<sub>FO</sub> output is open collector type. This signal line should be pulled up to the positive side of the 5V power supply with approximately 4.7kΩ resistance. Please refer to Figure 9.
 C<sub>SP15</sub> of around 7 times larger than bootstrap capacitor C<sub>BS</sub> is recommended.

5. V<sub>FO</sub> output pulse width should be determined by connecting an external capacitor(C<sub>FOD</sub>) between C<sub>FOD</sub>(pin7) and COM(pin2). (Example : if C<sub>FOD</sub> = 33 nF, then t<sub>FO</sub> = 1.8ms (typ.)) Please refer to the note 5 for calculation method.

6. Input signal is High-Active type. There is a 3.3kΩ resistor inside the IC to pull down each input signal line to GND. When employing RC coupling circuits, set up such RC couple that input signal agree with turm-off/turm-on threshold voltage.

7. To prevent errors of the protection function, the wiring around  $\mathsf{R}_\mathsf{F}$  and  $\mathsf{C}_\mathsf{SC}$  should be as short as possible.

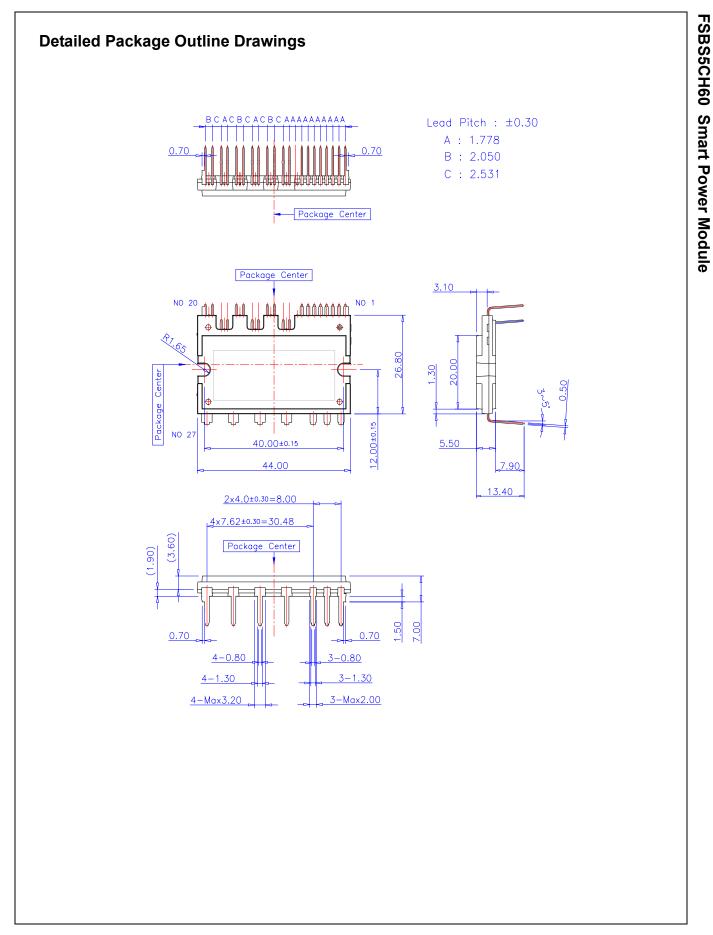
8. In the short-circuit protection circuit, please select the  $R_FC_{SC}$  time constant in the range 1.5~2  $\mu s.$ 

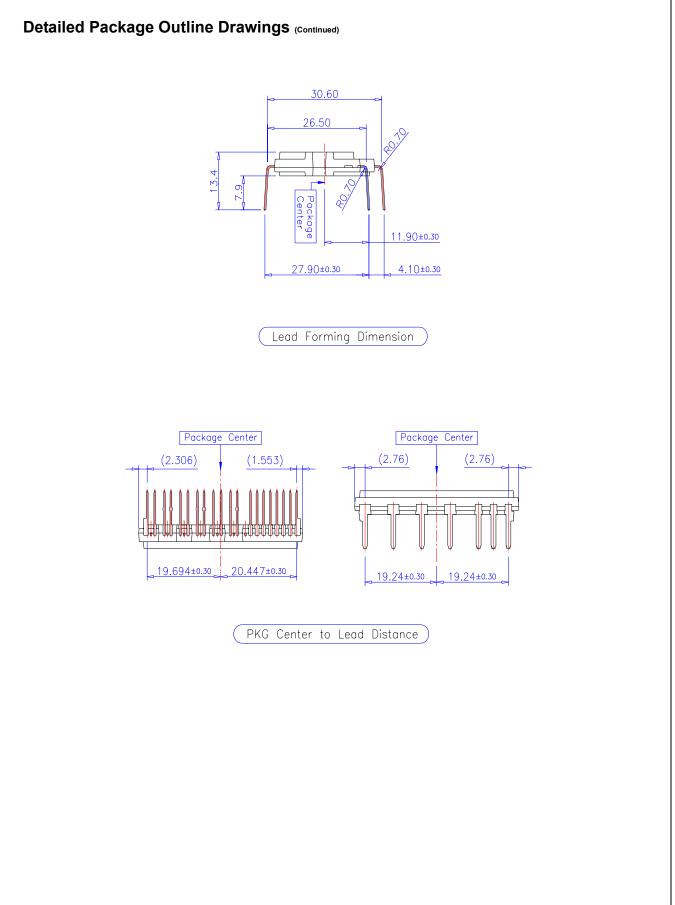
9. Each capacitor should be mounted as close to the pins of the SPM as possible.

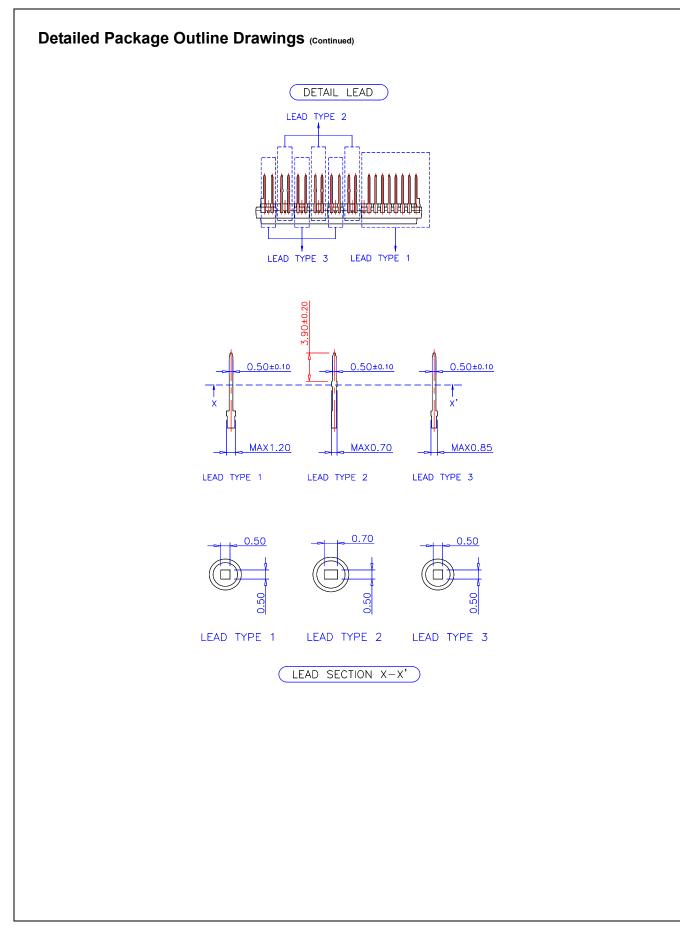
10. To prevent surge destruction, the wiring between the smoothing capacitor and the P&GND pins should be as short as possible. The use of a high frequency non-inductive capacitor of around 0.1~0.22 uF between the P&GND pins is recommended.

11. Relays are used at almost every systems of electrical equipments of home appliances. In these cases, there should be sufficient distance between the CPU and the relays. 12. C<sub>SPC15</sub> should be over 1uF and mounted as close to the pins of the SPM as possible.

### Figure 11. Typical Application Circuit







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