AMMC-5033

17.7 - 32 GHz Power Amplifier

Data Sheet





Chip Size: 2730 x 1300 μ m (108 x 51.6 mils) Chip Size Tolerance: \pm 10 μ m (\pm 0.4 mils) Chip Thickness: 100 \pm 10 μ m (4 \pm 0.4 mils) Pad Dimensions: 80 x 80 μ m (2.95 \pm 0.4 mils)

Description

Avago's AMMC-5033 is a MMIC power amplifier designed for use in wireless transmitters that operate within 17.7 GHz to 32 GHz range. At 25 GHz, it provides 27 dBm of output power (P-1dB) and 20 dB of small- signal gain from a small easy-to-use device. The device has input and output matching circuitry for use in 50 Ω environments. The AMMC- 5033 also integrates a temperature compensated RF power detection circuit that enables power detection of 0.1V/W at 22 GHz. For improved reliability and moisture protection, the die is passivated at the active areas.

AMMC-5033 Absolute Maximum Ratings^[1]

Symbol	Parameters/Conditions	Units	Min.	Max.
V _{d1,2}	Positive Drain Voltage	V		7
V_{g1}, V_{gg}	Gate Supply Voltage	V	-3	0.5
Det Bias	Applied Detector Bias (Optional)	V		7
l _{d1}	First Stage Drain Current	mA		320
l _{d2}	Second Stage Drain Current	mA		640
P _{in}	CW Input Power	dBm		23
T _{ch}	Operating Channel Temp.	°C		+150
T _{stg}	Storage Case Temp.	°C	-65	+150
T _{max}	Maximum Assembly Temp. (60 sec max)	°C		+300

Note:

1. Operation in excess of any one of these conditions may result in permanent damage to this device.

Features

- Wide frequency range: 17.7 32 GHz
- High power: P-1dB @ 25 GHz = 27 dBm
- High gain: 20 dB
- Return loss: Input: -13 dB, Output: -20 dB
- Integrated RF power detector

Applications

- Designed for use in transmitters that operate in various frequency bands between 17.7 GHz and 32 GHz.
- Can be driven by the AMMC-5040 (20-40 GHz) or the AMMC-5618 (6-20 GHz) MMIC amplifiers, increasing the power handling capability of transmitters requiring linear operation.

Note: These devices are ESD sensitive. The following precautions are strongly recommended: Ensure that an ESD approved carrier is used when dice are transported from one destination to another. Personal grounding is to be worn at all times when handling these devices.

Symbol	Parameters and Test Conditions	Units	Min.	Тур.	Max.
l _{d1}	First Stage Drain Supply Current ($V_{d1} = 3.5 V, V_{g1} = Open, V_{gg}$ set for I _{d2} Typical)	mA		280	320
l _{d2}	Second Stage Drain Supply Current ($V_{d2} = 5 V$, $V_{g1} = Open$, V_{gg} set for I_{d2} Typical)	V		500	
Vgg	Gate Supply Operating Voltage $(I_{d1(Q)} + I_{d2(Q)} = 780 \text{ (mA)})$	V	-0.95	-0.6	-0.4
DETBias	Detector Bias Voltage (Optional)	V		Vd2	
$\theta_{c1(ch-bs)}$	First Stage Thermal Resistance ^[2] (Backside Temperature, Tb = 25°C)	°C/W		31	
$\theta_{c2(ch-bs)}$	Second Stage Thermal Resistance ^[2, 3] (Backside Temperature, Tb = 25°C)	°C/W		19	

AMMC-5033	DC	Specifications	/Physical	Properties ^[1]
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Notes:

1. Backside temperature $T_{b} = 25^{\circ}C$ unless otherwise noted.

2. Channel-to-backside Thermal Resistance (θch-b) = 42°C/W at Tchannel (Tc) = 150°C as measured using infrared microscopy. Thermal Resistance at backside temperature (Tb) = 25°C calculated from measured data.

AMMC-5033 RF Specifications^[4, 5]

 $\underline{\text{Tb}=25^{\circ}\text{C},\text{V}_{_{d1}}=3.5\text{ V},\text{V}_{_{d2}}=5\text{ V},\text{ I}_{_{d1(Q)}}=280\text{ mA},\text{I}_{_{d2(Q)}}=500\text{ mA},\text{Zo}=50\ \Omega}$

			Lower Band Specifications (17.7 - 21 GHz)		Mid Band			Upper Band			
	Parameters and				Specifications (21 - 26.5 GHz)		Speci (26.5 - 32 GHz)		fications		
Symbol	Test Condition	Unit	Min.	Тур.	Max.	Min.	Тур.	Max.	Min.	Тур.	Max.
Gain	Small-Signal Gain ^[5]	dB	20	22		17.5	20		16.5	18.5	
P _{-1dB}	Output Power at 1dB Gain Compression ^[6]	dB	23.5	25		25.5	27		25	26.5	
P _{-3dB}	Output Power at 3dB Gain Compression ^[6]	dB		27			28			27	
OIP ₃	Output Third Order Intercept Point; ⁽⁶⁾ ; $\Delta f = 2 \text{ MHz}$; Pin = +2 dBm	dBm	27	29		29	32		29	32	
RL _{in}	Input Return Loss ^[5]	dB	11.5	13.5		11	13		11	13	
RL _{out}	Output Return Loss ^[5]	dB	14	20		14	19		15	22	
Isolation	Min. Reverse Isolation	dB		47			48			46	

Notes:

4. Data measured in wafer form $T_b = 25^{\circ}$ C. 5. 100% on-wafer RF test is done at frequency = 17.7, 21, 26.5 and 32 GHz.

6. 100% on-wafer test frequency = 17.7, 26.5 and 32 GHz.

AMMC-5033 Typical Performances

 $(T_{_{\rm D}} = 25^{\circ}\text{C}, V_{_{d1}} = 3.5 \text{ V}, \ I_{_{D1}} = 280 \text{ mA}, V_{_{d2}} = 5 \text{ V}, \ I_{_{d2}} = 500 \text{ mA}, Z_{_{\rm in}} = Z_{_{\rm out}} = 50 \Omega)$



Figure 1. Gain and reverse isolation



Figure 3. Output power at 1 dB and 3 dB gain compression



Figure 5. Output 3rd order intercept point



Figure 2. Return loss (input and output)



Figure 4. Noise figure

AMMC-5033 Typical Performance Curves (Over Temperature and Voltage)



Figure 6. Linear and log detector voltage and output power, freq. = 22 GHz, Det_B = 5 V



Figure 7. Gain and V_{d2} voltage, $V_{d1} = 3.5$ V (constant)



Figure 8. Output power at 1 dB gain compression and $V_{\rm d2}$ voltage, $V_{\rm d1}$ = 3.5 V (constant)



Figure 9. Return-loss with temperature



Figure 10. Output power at 1 dB gain compression and temperature



Figure 11. Output power, PAE, and total drain. Current vs. input power at 25 GHz



Figure 12. Gain with temperature

Typical Scattering Parameters^[1]

$(T_{1} = 25^{\circ}C, V_{2})$. = 3.5 V, I.,	$= 280 \text{ mA}, \text{V}_{12}$	= 5 V, I =	= 500 mA,	$Z_{.} = Z_{.}$	= 50 Ω)
(.b = 0,.d	1 0.0 ., . _{D1}				Lin Lout	

Freq		S11			S21			S12			S22	
[GHz]	dB	Mag	Phase	dB	Mag	Phase	dB	Mag	Phase	dB	Mag	Phase
1	-10.7	0.29	173	-51.1	0.003	-163	-95.1	1.77E-05	128	-0.5	0.95	-26
2	-11.0	0.28	167	-70.1	0	79	-83.1	6.97E-05	76	-0.7	0.92	-51
3	-11.4	0.27	161	-46.6	0.005	-103	-74.5	1.89E-04	81	-1.2	0.87	-76
4	-12.1	0.25	153	-37.3	0.014	72	-74.5	1.88E-04	69	-2.2	0.78	-95
5	-15.3	0.17	140	-22.6	0.074	-31	-80.3	9.66E-05	-47	-2.4	0.76	-112
6	-12.2	0.25	149	-20.4	0.096	144	-80.1	9.90E-05	-126	-2.8	0.73	-130
7	-14.0	0.2	143	-20.5	0.095	79	-80.3	9.68E-05	94	-4.0	0.63	-146
8	-15.3	0.17	139	-25.4	0.053	-3	-74.1	1.97E-04	35	-4.2	0.62	-150
9	-17.6	0.13	138	-33.1	0.022	108	-81.4	8.52E-05	-62	-3.5	0.67	-166
10	-19.4	0.11	145	-18.9	0.113	54	-81.4	8.56E-05	-162	-3.8	0.64	177
11	-18.6	0.12	152	-18.2	0.123	-37	-81.3	8.59E-05	151	-4.4	0.6	161
12	-19.7	0.1	141	-29.0	0.035	-77	-74.6	1.86E-04	178	-5.2	0.55	146
13	-24.5	0.06	134	-15.4	0.169	103	-81.3	8.65E-05	-180	-6.3	0.48	131
14	-27.4	0.04	159	0.9	1.107	61	-81.2	8.70E-05	-20	-7.9	0.41	115
15	-30.6	0.03	-148	12.7	4.316	-8	-74.6	1.86E-04	152	-10.1	0.31	100
16	-24.1	0.06	-121	22.6	13.52	-87	-76.2	1.55E-04	144	-13.3	0.22	86
17	-21.2	0.09	-116	28.8	27.62	174	-74.7	1.84E-04	-164	-20.5	0.09	76
18	-18.0	0.13	-116	28.7	27.25	73	-64.8	5.75E-04	165	-20.0	0.1	133
19	-15.5	0.17	-123	26.4	20.92	3	-64.3	6.08F-04	123	-19.4	0.11	130
20	-14.0	0.2	-133	24.7	17 18	-53	-64.4	6.03E-04	90	-19.1	0.11	135
21	-13.3	0.22	-142	23.4	14.82	-103	-69.7	3 27F-04	76	-18.7	0.12	133
27	-13.0	0.22	-151	22.1	13.2	-151	-58.2	1 23E-03	80	-18.5	0.12	129
22	-12.9	0.22	-157	21.1	11.9	164	-63.3	6.80E-04	92	-19.0	0.12	122
23	-12.9	0.23	-163	20.8	10.97	121	-61.0	8 96E-04	44	-20.7	0.09	110
25	-13.0	0.23	-172	20.0	10.36	79	-66.1	4 97F-04	55	-21.8	0.09	122
26	-13.3	0.23	-178	19.9	9.895	37	-64 3	6.09E-04	53	-27.9	0.07	131
20	-13.9	0.22	174	19.7	9.691	-6	-63.1	7.00E-04	58	-22.9	0.07	135
27	-14.9	0.18	165	19.7	9.457	_49	-60.2	9 75E-04	68	-22.5	0.07	142
20	-15.8	0.16	155	19.5	9 384	_94	-61.9	8.00E-04	38	-22.0	0.08	136
30	-17.0	0.10	140	19.4	9.304	-141	-56.3	1.53E-03	25	-22.1	0.00	125
30	-19.1	0.11	113	19.5	8 972	171	-57.7	1.33E-03	15	-74.9	0.06	117
37	-21.0	0.09	75	18.6	8 5 1 9	121	-58.2	1.31E 03	15	-31.0	0.00	126
32	-20.5	0.05	30	18.1	7 989	69	-56.0	1.25E 03	-15	-31.2	0.03	-148
34	-17.0	0.1	_9	17.2	7.281	14	-57.7	1.37E-03	-12	-74.4	0.05	-141
35	-14.9	0.14	-31	16.2	6.44	-43	-59.0	1.51E 03	-36	-20.0	0.00	-145
36	-12.8	0.10	-45	14.6	5 378	-104	-60.8	9 14F-04	-40	-16.2	0.16	-152
37	-10.7	0.25	-58	12.1	4 014	-171	-62.9	7 13E-04	-31	-13.1	0.10	-167
38	_0.8	0.29	-30	77	2 / 2	122	-02.9	1.40E-03	-55	_11.1	0.22	17/
30	_9.0	0.35	-71	1.0	1 238	65	-61.0	8.94E-04	-61	-10.1	0.20	154
<u> </u>	-9.1	0.35	-85	-3.5	0.671	1/	-60.9	0.94E-04	-59	_0.8	0.31	13/
40 //1	-8.6	0.30	-05		0.071	_/1	-67.6	1 15E-04	-65	-9.0	0.33	116
42 17	-8.6	0.37	_02	-9.2	0.547	_00	-50.2	1.00E_03	_05	-9.7	0.33	08
42	-8.0	0.37	-92	-10.1	0.157	-90	-59.2	8.05E-04	-02	-10.1	0.31	<u>90</u>
45	-0.0	0.4	-92	22.2	0.009	172	62.0	7.06E.04	-75	11.4	0.29	62
44	-7.0	0.42	-90	-52.0	0.023	1/2	-02.0	7.90E-04	122	122	0.27	<u> </u>
+J 16	-0.0	0.5	-07	-51.7	0.020	164	-04.0	0.01E-04	-125	12.3	0.24	21
40	-4.4	0.0	-92	-40.7	0.009	-104	-102.4	0.04E-U4	-02	-13.2	0.22	21
+/	-5.5	0.07	-70	-40.Z	0.005	02	-102.4	0.02E.04	176	1/1.2	0.2	<u></u>
40 40	-2./	0.74	-102	-20.4	0.001	50	-00.1	9.93E-04	60	-14.8	0.10	-27
+7 50	-1.0	0.01	110	-40.4	0.005	112	-57.2	0.00E 0.4	-09	-14./	0.10	-49
50	-1./	0.83	-118	-44.2	0.006	113	-91.9	8.00E-04	20	-14.9	0.18	-//

Note:

1. Data obtained from on-wafer measurements.

Biasing and Operation

The recommended quiescent DC bias condition for optimum efficiency, performance, and reliability is $V_{d1} = 3.5$ volts and $V_{d2} = 5$ volts with V_{gg} set for $I_{d1} + I_{d2} = 780$ mA (no connection to V_{g1}). This bias arrangement results in default quiescent drain currents $I_{d1} = 280$ mA, $I_{d2} = 500$ mA. A single DC gate supply connected to V_{gq} will bias all gain stages.

If operation with both V_{d1} and V_{d2} at 5 volts is desired, an additional wire bond connection from the V_{g1} pad to V_{gg} external bypass chip capacitor (shorting V_{g1} to V_{g9}) will balance the current in each gain stage. V_{g9} (= V_{g1}) can be adjusted for I_{d1} + I_{d2} = 780 mA. Muting can be accomplished by setting V_{g1} and/or V_{g9} to the pinchoff voltage V_p.

An optional output power detector network is also provided. Detector sensitivity can be adjusted by biasing the diodes with typically 1 to 5 volts applied to the Det-bias terminal. Simply connecting Det-Bias to the V_{d2} supply is a convenient method of biasing this detector network. The differential voltage between the Det-Ref and Det-Out pads can be correlated with the RF power emerging from the RF output port. The detected voltage is given by:

$$V = (V_{ref} - V_{det}) - V_{ofs}$$

Where V_{ref} is the voltage at the DET_REF port, V_{det} is a voltage at the DET_OUT port, and V_{ofs} is the zero-input-power offset voltage. There are three methods to calculate V_{ofs} :

- V_{ofs} can be measured before each detector measurement (by removing or switching off the power source and measuring V_{ref} V_{det}). This method gives an error due to temperature drift of less than 0.0002 dB/°C.
- 2. V_{ofs} can be measured at a single reference temperature. The drift error will be less than 0.25 dB.
- 3. V_{ofs} can either be characterized over temperature and stored in a lookup table, or it can be measured at two temperatures and a linear fit used to calculate V_{ofs} at any temperature. This method gives an error close to method #1.

With reference to Figure 13, the RF input is DC coupled to a shunt 50 Ω resistor but it is DC blocked to the input of the first stage. The RF output is DC blocked to the output of the second stage, however, it is DC coupled to the detector bias circuit. If the output detector is biased using the on-chip optional Det-Bias network, an external DC blocking capacitor may be required at the RF Output port.

No ground wires are needed since ground connections are made with plated through-holes to the backside of the device.

Assembly Techniques

The backside of the AMMC- 5033 chip is RF ground. For microstripline applications, the chip should be attached directly to the ground plane (e.g., circuit carrier or heat-sink) using electrically conductive epoxy.^[1,2]

For best performance, the topside of the MMIC should be brought up to the same height as the circuit surrounding it. This can be accomplished by mounting a gold plated metal shim (same length and width as the MMIC) under the chip, which is of the correct thickness to make the chip and adjacent circuit coplanar.

The amount of epoxy used for chip and or shim attachment should be just enough to provide a thin fillet around the bottom perimeter of the chip or shim. The ground plane should be free of any residue that may jeopardize electrical or mechanical attachment.

The location of the RF bond pads is shown in Figure 14. Note that all the RF input and output ports are in a Ground-Signal-Ground configuration.

RF connections should be kept as short as reasonable to minimize performance degradation due to undesirable series inductance. A single bond wire is sufficient for signal connections, however double-bonding with 0.7 mil gold wire or the use of gold mesh is recommended for best performance, especially near the high end of the frequency range.

Thermosonic wedge bonding is the preferred method for wire attachment to the bond pads. Gold mesh can be attached using a 2 mil round tracking tool and a tool force of approximately 22 grams with an ultrasonic power of roughly 55 dB for a duration of 76 \pm 8 mS. A guided wedge at an ultrasonic power level of 64 dB can be used for the 0.7 mil wire. The recommended wire bond stage temperature is 150 \pm 2°C.

Caution should be taken to not exceed the Absolute Maximum Rating for assembly temperature and time.

The chip is 100 μ m thick and should be handled with care. This MMIC has exposed air bridges on the top surface and should be handled by the edges or with a custom collet (do not pick up die with vacuum on die center.)

This MMIC is also static sensitive and ESD handling precautions should be taken.

Notes:

^{1.} Ablebond 84-1 LM1 silver epoxy is recommended.

^{2.} Eutectic attach is not recommended and may jeopardize reliability of the device.







Figure 14. AMMC-5033 bonding pad locations, dimensions are in microns



Figure 15. AMMC-5033 assembly diagram

Ordering Information:

AMMC-5033-W10 = 10 devices per tray AMMC-5033-W50 = 50 devices per tray

For product information and a complete list of distributors, please go to our website: www.avagotech.com

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