

## **RF3861** WIDE BANDWIDTH, HIGH LINEARITY

## LOW NOISE AMPLIFIER

Package Style: QFN, 16-Pin, 3mmx3mm



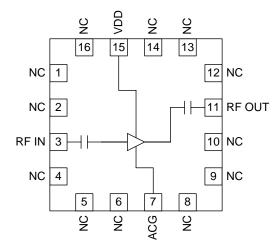


#### Features

- Low Noise and High Intercept Point
- Adjustable Bias for Enhanced IP3
- Single 2.5V to 6.0V Power Supply
- 400 MHz to 3800 MHz Operation
- QFN16, 3mmx3mm Package

#### Applications

- GSM/EDGE. CDMA, PCS, UMTSLNA/Linear Driver
- WLAN LNA/Linear Driver
- WiMAX LNA/Linear Driver
- 900 MHz LNA/Linear Driver
- General Purpose Amplification



**Functional Block Diagram** 

#### **Product Description**

The RF3861 is a low noise amplifier with a high output IP3. The amplifier is self-biased from a single voltage supply with 50 $\Omega$  input and output ports. The useful frequency range is from 400MHz to 3800MHz. A 1dB noise figure and 36dBm OIP3 performance is achieved with a 5V V<sub>DD</sub>, 90 mA. Current can be increased to raise OIP3 while having minimal effect on noise figure. The IC is featured in a standard QFN, 16-pin, 3mmx3mm package.

#### **Ordering Information**

RF3861 RF3861PCK-410

Wide Bandwidth, High Linearity Low Noise Amplifier Fully Assembled Evaluation Board with 5 Sample Parts 1.5GHz to 2.7GHz

#### **Optimum Technology Matching® Applied**

🗌 GaAs HBT	□ SiGe BiCMOS	🗹 GaAs pHEMT	🗌 GaN HEMT
GaAs MESFET	🗌 Si BiCMOS	🗌 Si CMOS	RF MEMS
🗌 InGaP HBT	SiGe HBT	🗌 Si BJT	

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#### **Absolute Maximum Ratings**

Parameter	Rating	Unit		
Supply Voltage	6V	V <sub>DC</sub>		
Input RF Level	+10	dBm		
Current Drain, I <sub>DD</sub>	150	mA		
Operating Ambient Temperature	-40 to +85	°C		
Storage Temperature	-40 to +150	°C		

Note 1: Max continuous RF IN is +10dBm. The max transient RF IN is +20dBm.



**Caution!** ESD sensitive device.

Exceeding any one or a combination of the Absolute Maximum Rating conditions may cause permanent damage to the device. Extended application of Absolute Maximum Rating conditions to the device may reduce device reliability. Specified typical performance or functional operation of the device under Absolute Maximum Rating conditions is not implied.

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Devementer		Specification		Unit	Condition	
Parameter	Min.	Тур.	Max.	Unit	Condition	
High Band						
Frequency	3.3	3.5	3.8	GHz	V <sub>DD</sub> =5V	
Current		90	110	mA		
Gain		10		dB	At 3.5GHz	
Noise Figure		1.1		dB	+25°C, V <sub>DD</sub> =5V, I <sub>DD</sub> =90mA, 3500MHz unless specified	
OIP3		37.0		dBm		
OP1dB		22.0		dBm		
S11		-12		dB	f <sub>1</sub> =3500MHz, f <sub>2</sub> =3501MHz	
S22		-18		dB		
Mid Band						
Frequency	1500		2700	MHz		
Current		90	110	mA	V <sub>DD</sub> =5V	
Gain	13.5	14.5	16.5	dB	+25°C, V <sub>DD</sub> =5V, I <sub>DD</sub> =90mA, 2000MHz unless specified	
Noise Figure		1.0	1.2	dB		
OIP3	33.0	35.5		dBm	f <sub>1</sub> =2000MHz, f <sub>2</sub> =2001MHz	
OP1dB	21.0	22.5	25.0	dBm		
S11		-10		dB		
S22		-20		dB		
Low Band						
Frequency	700		1100	MHz		
Current		90	110	mA	V <sub>DD</sub> =5V	
Gain		16		dB	+25°C, V <sub>DD</sub> =5V, I <sub>DD</sub> =90mA, 850MHz unless specified	
Noise Figure		1.2		dB		
OIP3		36.0		dBm	f <sub>1</sub> =850MHz, f <sub>2</sub> =851MHz	
OP1dB		22.5		dBm		
S11		-14		dB		
S22		-18		dB		
Thermal						
Theta <sub>JC</sub>		51		°C/W		
Power Supply						
Device Operating Voltage	2.5	5.0	6.0	V		
Operating Current	65	90	110	mA	V <sub>DD</sub> =5.0V, R2=open	

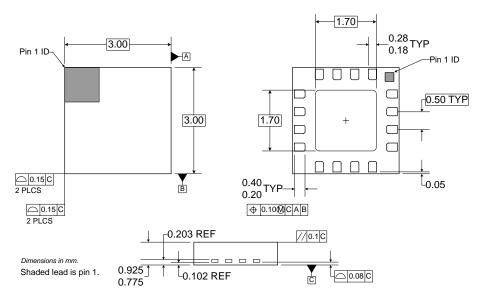
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Pin	Function	Description	Interface Schematic
1	NC	Not connected.	
2	NC	Not connected.	
3	RF IN	RF input pin. 50 $\Omega$ matched. This pin is DC-blocked.	
4	NC	Not connected.	
5	NC	Not connected.	
6	NC	Not connected.	
7	ACG	AC ground. Shunt cap may be added for tuning. Shunt resistor may be added to increase $\rm I_{\rm DD}/\rm IP3.$	
8	NC	Not connected.	
9	NC	Not connected.	
10	NC	Not connected.	
11	<b>RF OUT</b>	RF output pin. 50 $\Omega$ matched. This pin is DC-blocked.	
12	NC	Not connected.	
13	NC	Not connected.	
14	NC	Not connected.	
15	VD	Bias voltage. 2.5V to 6.0V applied through bias inductor.	
16	NC	Not connected.	
Pkg Base	GND	Ground connection.	

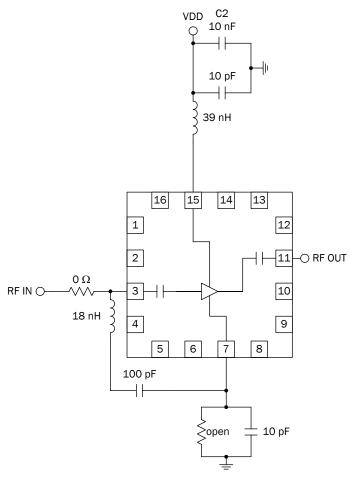
### **Package Drawing**







#### Application Schematic 400 MHz to 1300 MHz

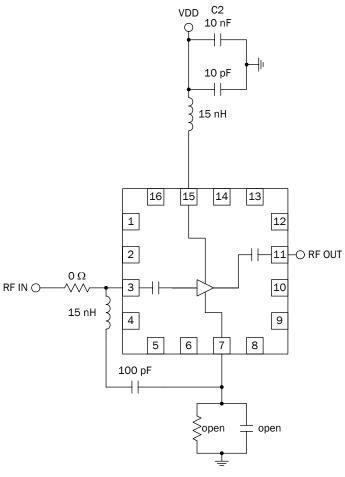


FREQ	VCC	NF	GAIN	OIP3	OP1dB
MHz	Volts	dB	dB	dBm	dBm
400	5	1.7	15.23	35.45	21.27
600	5	1.4	15.81	37.33	21.84
800	5	1.2	16.49	36.56	22.05
1000	5	1.1	16.7	36.22	22.02
1300	5	1.1	15.28	36.88	22.19

Note: This schematic is equivalent to standard 700 MHz to 1100 MHz evaluation board. Specification in above table shows tested performance over extended frequency range. See Theory of Operation section for details.



### Application Schematic 800MHz to 2200MHz

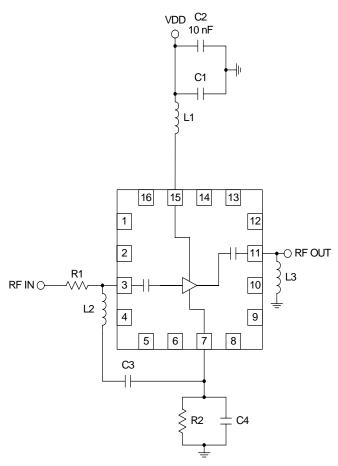


FREQ	VCC	NF	GAIN	OIP3	OP1dB
MHz	Volts	dB	dB	dBm	dBm
800	5	1.3	16	36	22
2200	5	0.9	13.9	36	22





#### Evaluation Board Schematic 700MHz to 3800MHz

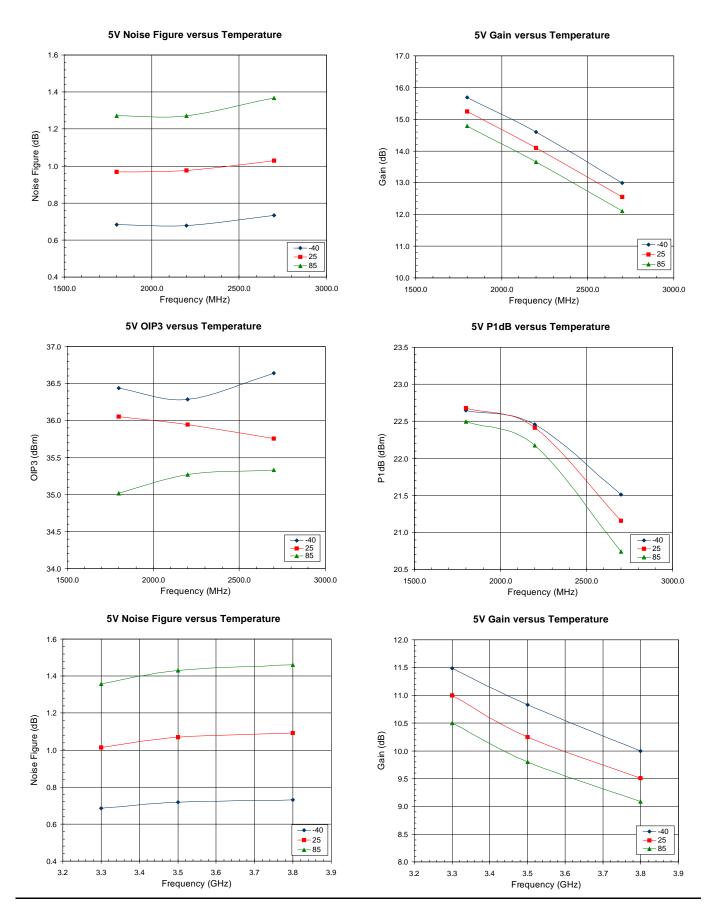


Components	700-1100 MHz	1.5-2.7 GHz	3.3-3.8 GHz
C1 (pF)	10	10	10
C3 (pF)	100	100	100
C4 (pF)	10	DNP	DNP
L1 (nH)	39	3.9	2.2
L2 (nH)	18	4.7	4.7
R1 (ohm)	0	0	0
R2 (ohm)	DNP	DNP	DNP
L3 (nH)	DNP	DNP	DNP

R2 is DNP for standard 90 mA current draw. If R2 is added, the I<sub>DD</sub> will increase. A 20  $\Omega$  R2 will raise the current to achieve higher linearity.







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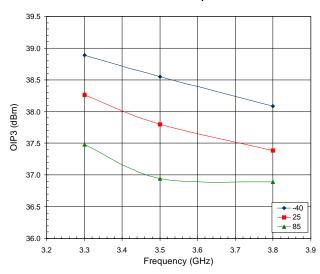
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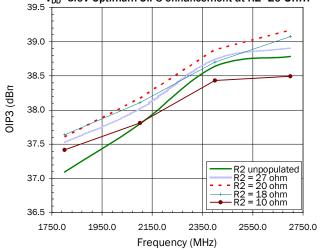
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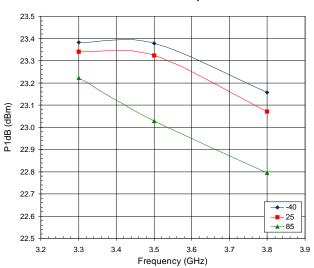
5V OIP3 versus Temperature

5V P1dB versus Temperature

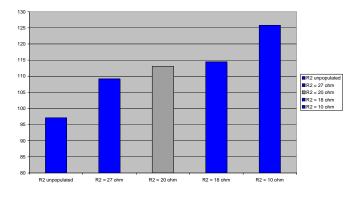


410 Evaluation Board, OIP3 versus R2 value V<sub>DD</sub>=5.0V Optimum OIP3 enhancement at R2=20 Ohm





RF3861 Icc vs R2 Vdd = 5.0 V R2 = 20 ohm for optimal OIP3





#### **Theory of Operation**

Low noise figure/high IP3 make RF3861 ideal for use as both receive LNA and transmit driver for cellular/DCS/PCS/UMTS and WiMax platforms, in addition to many other general purpose applications. Standard evaluation boards cover 700MHz to 1100MHz, 1500MHz to 2700MHz, and 3300MHz to 3800MHz. Viewing the data sheet evaluation board schematic, refer to below for purpose/function of external components:

- R1/L3 (0Ω/unpopulated on standard evaluation boards): These unused components were placed for convenience and flexibility when needed to optimize matching for an out of band application.
- L2/C3/C4: Placed to optimize input match, and enhance out of band low frequency stability.
- R2: Optionally placed to increase bias current and IP3. It has been found that 20Ω value is best case (see graph section of data sheet).
- L1/C1: Influence output return loss.

RF3861 has internal DC blocking capacitors at RFin/RFout. In addition, it has been shown impedance seen looking out at pins 7/15 influence response. As a result, two port s-parameters become non-applicable. In the event matching is desired for frequency bands outside of those provided with standard evaluation boards, the following approach can be used:

- Start with matching seen for standard evaluation board closest to desired band of operation.
- Optimize values at L2/C4/L1 to obtain response/performance.

The application schematic section of data sheet shows matching arrived at using above procedure, for 400MHz to 1300MHz and 800MHz to 2200MHz. These schematics, along with standard evaluation boards, cover 82% of the useable bandwidth from 400MHz to 3800MHz.

One interesting note concerning these application schematics, the 400MHz to 1300MHz example shows same component values as seen with 700MHz to 1100MHz evaluation board. So, it is, in actuality, the same board. The specification chart next to the schematic simply extends the frequency band by 500MHz. Input and output return losses are better than 10dB over the entire 900MHz bandwidth.

