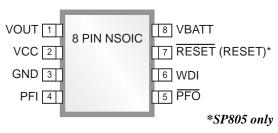


## SP690A/692A/802L/ 802M/805L/805M

### Low Power Microprocessor Supervisory with Battery Switch-Over

#### FEATURES

- Precision Voltage Monitor: SP690A/SP802L/SP805L at 4.65V SP692A/SP802M/SP805M at 4.40V
- Reset Time Delay 200ms
- Watchdog Timer 1.6 sec timeout
- Minimum component count
- 60µA Maximum Operating Supply Current
- 0.6µA Maximum Battery Backup Current
- 0.1µA Maximum Battery Standby Current
- Power Switching
  250mA Output in V<sub>cc</sub> Mode (0.6Ω)
  25mA Output in Battery Mode (5Ω)
- Voltage Monitor for Power Fail or Low Battery Warning
- Available in 8 pin SO and DIP packages
- **RESET** asserted down to  $V_{cc} = 1V$



Now Available in Lead Free Packaging

Pin Compatible Upgrades to MAX690A/692A/802L/802M/805L

#### APPLICATIONS

- Critical µP Power Monitoring
- Intellegent Instruments
- Computers
- Controllers

#### DESCRIPTION

The **SP690A/692A/802L/802M/805L/805M** are a family of microprocessor ( $\mu$ P) supervisory circuits that integrate a myriad of components involved in discrete solutions to monitor power-supply and battery-control functions in  $\mu$ P and digital systems. The series will significantly improve system reliability and operational efficiency when compared to discrete solutions. The features of the **SP690A/692A/802L/802M/805L/805M** include a watchdog timer, a  $\mu$ P reset and backup-battery switchover, and power-failure warning, a complete  $\mu$ P monitoring and watchdog solution. The series is ideal for applications in automotive systems, computers, controllers, and intelligent instruments. All designs where it is critical to monitor the power supply to the  $\mu$ P and it's related digital components will find the series to be an ideal solution.

PART NUMBER	RESET Threshold	RESET Accuracy	RESET Active	PFI Accuracy
SP690A	4.65 V	125mV	LOW	4%
SP692A	4.40 V	125mV	LOW	4%
SP802L	4.65 V	75mV	LOW	2%
SP802M	4.40 V	75mV	LOW	2%
SP805L	4.65 V	125mV	HIGH	4%
SP805M	4.40 V	125mV	HIGH	4%

Date: 11/29/04

#### **ABSOLUTE MAXIMUM RATINGS**

These are stress ratings only and functional operation of the device at these ratings or any other above those indicated in the operation sections of the specifications below is not implied. Exposure to absolute maximum rating conditions for extended periods of time may affect reliability and cause permanent damage to the device.

the device.
V <sub>cc</sub> 0.3V to 6.0V V <sub>BATT</sub> 0.3V to 6.0V
1/1000
V <sub>BATT</sub> 0.3V 10 0.0V
All Other Inputs (NOTE 1)0.3V to $(V_{cc} \text{ to } 0.3V)$
Input Current:
V
V <sub>CC</sub> 250mA V <sub>BATT</sub>
V BATT
GÑD20mA
Output Current:
V <sub>our</sub> Short-Circuit Protected for up to 10sec All Other Inputs20mA
V <sub>our</sub> Short-Circuit Protected for up to 10sec All Other Inputs20mA
V <sub>our</sub> Short-Circuit Protected for up to 10sec All Other Inputs20mA
V <sub>out</sub> Short-Circuit Protected for up to 10sec All Other Inputs20mA Rate of Rise, V <sub>cc</sub> ,V <sub>BATT</sub> 100V/µs Continuous Power Dissipation500mW
V <sub>out</sub> Short-Circuit Protected for up to 10sec All Other Inputs20mA Rate of Rise, V <sub>cc</sub> ,V <sub>BATT</sub> 100V/µs Continuous Power Dissipation500mW
V <sub>out</sub> Short-Circuit Protected for up to 10sec All Other Inputs20mA Rate of Rise, V <sub>cc</sub> ,V <sub>BATT</sub> 100V/μs Continuous Power Dissipation500mW Storage Temperature65°C to +160°C
V <sub>out</sub> Short-Circuit Protected for up to 10sec All Other Inputs20mA Rate of Rise, V <sub>CC</sub> ,V <sub>BATT</sub> 100V/μs Continuous Power Dissipation500mW Storage Temperature65°C to +160°C Lead Temperature(soldering,10sec)+300°C
V <sub>out</sub> Short-Circuit Protected for up to 10sec All Other Inputs20mA Rate of Rise, V <sub>cc</sub> ,V <sub>BATT</sub> 100V/μs Continuous Power Dissipation500mW Storage Temperature65°C to +160°C



**CAUTION:** ESD (ElectroStatic Discharge) sensitive device. Permanent damage may occur on unconnected devices subject to high energy electrostatic fields. Unused devices must be stored in conductive foam or shunts. Personnel should be properly grounded prior to handling this device. The protective foam should be discharged to the destination socket before devices are removed.

#### **ELECTRICAL CHARACTERISTICS**

 $V_{cc}$ =4.75v to 5.50V for SP690A/SP802L/SP805L,  $V_{cc}$ =4.50V to 5.50V for SP692A/SP802M/SP805M,  $V_{BATT}$ =2.80V,  $T_A = T_{MIN}$  to  $T_{MAX}$ , typical specified at 25°C, unless otherwise noted.

PARAMETERS	MIN.	TYP.	MAX.	UNITS	CONDITIONS
Operating Voltage Range,	0		5.5	Volts	
$V_{\rm CC}$ or $V_{\rm BATT}$ , Note 2					
Supply Current, I <sub>SUPPLY</sub> ,		35	60	μΑ	excluding I <sub>out</sub>
$I_{SUPPLY}$ in Battery Backup Mode, $V_{CC} = 0V, V_{BATT} = 2.8V$		0.001	0.6	μΑ	
V <sub>BATT</sub> Standby Current, NOTE 3	-0.1		0.02	μA	$V_{\rm CC} > V_{\rm BATT} + 0.2V$
V <sub>out</sub> Output	V <sub>cc</sub> - 0.1	V <sub>cc</sub> - 0.03 V <sub>cc</sub> - 0.15		Volts	Ι <sub>ουτ</sub> = 50mA Ι <sub>ουτ</sub> = 250mA
$V_{out}$ in Battery-Backup Mode $V_{cc} < V_{BATT}$ - 0.2V	V <sub>BATT</sub> -0.15	V <sub>BATT</sub> - 0.04 V <sub>BATT</sub> - 0.20		Volts	I <sub>out</sub> = 5mA I <sub>out</sub> = 25mA
Battery Switch Threshold, $V_{cc}$ to $V_{BATT}$		20 -20		mV	Power-up Power-down
Battery Switchover Hysteresis		40		mV	Peak to Peak
Reset Threshold	4.50 4.25 4.55 4.30	4.65 4.40	4.75 4.50 4.70 4.45	Volts	SP690A, SP802L, SP805L SP692A, SP802M, SP805M SP802L, $T_A = +25^{\circ}$ C, $V_{cc}$ falling SP802M, $T_A = +25^{\circ}$ C, $V_{cc}$ falling

#### **ELECTRICAL CHARACTERISTICS**

 $V_{\infty}$ =4.75v to 5.50V for SP690A/SP802L/SP805L,  $V_{CC}$ =4.50V to 5.50V for SP692A/SP802M/SP805M,  $V_{BATT}$ =2.80V,  $T_{A}$ = $T_{MIN}$  to  $T_{MAX}$ , typical specified at 25°C, unless otherwise noted.

PARAMETERS	MIN.	TYP.	MAX.	UNITS	CONDITIONS
Reset Threshold Hysteresis		40		mV	Peak to Peak
Reset Pulse Width, t <sub>RS</sub>	140	200	280	ms	
RESET Output Voltage,	V <sub>cc</sub> - 1.5				I <sub>SOURCE</sub> = 800μA
NOTE 5		0.1	0.4	Volts	I <sub>SINK</sub> = 3.2mA
		0.004	0.3		$I_{SINK} = 50\mu A, V_{CC} = 1.0$
RESET Output Voltage,	0.8				$I_{\text{SOURCE}} = 4\mu A, V_{\text{CC}} = 1.0V,$
NOTE 6	V <sub>cc</sub> - 1.5			Volts	I <sub>SOURCE</sub> = 800μA
		0.1	0.4		I <sub>SINK</sub> = 3.2mA
Watchdog Timeout, t <sub>wD</sub>	1.00	1.60	2.25	sec	
WDI Pulse Width, twp NOTE 7	50			ns	$V_{IL} = 0.4V, V_{IH} = (0.8)(V_{CC})$
WDI Input Threshold, V <sub>cc</sub> = 5V, NOTE 4	3.5		0.8	Volts	Logic low Logic high
WDI Input Current	-150	50 -50	150	μA	WDI =V <sub>cc</sub> WDI = 0V
PFI Input Threshold	1.200 1.225	1.250 1.250	1.300 1.275	Volts	SP690A/692A, SP805L/M SP802L/M
PFI Input Current	-25	0.01	25	nA	
PFO Output Voltage	V <sub>cc</sub> - 1.5	0.1	0.4	Volts	I <sub>SOURCE</sub> = 800μA I <sub>SINK</sub> = 3.2mA

**NOTE 1:** The input voltage limits on PFI (pin 4) and WDI (pin 6) may be exceeded if the current into these pins is limited to less than 10 mA.

**NOTE 2:** Either  $V_{CC}$  or  $V_{BATT}$  can go to 0V if the other is greater than 2.0V.

**NOTE 3:** "-" equals the battery-charging current, "+" equals the battery-discharging current.

**NOTE 4:** WDI is guaranteed to be in an intermediate, non-logic level state if WDI is floating and  $V_{cc}$  is in the operating voltage range. WDI is internally biased to 35% of  $V_{cc}$  with an input impedance of 50K $\Omega$ .

NOTE 5: SP690A, SP692A, SP802L, and SP802M only.

**NOTE 6:** SP805L and SP805M only.

**NOTE 7:** WDI Minimum Rise/Fall time is 2µs.

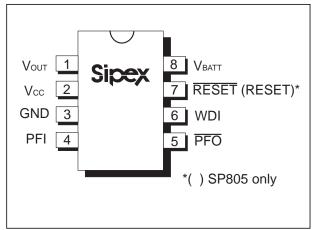


Figure 10. Pinout

#### **PIN ASSIGNMENTS**

- Pin 1  $V_{OUT}$  Output Supply Voltage.  $V_{OUT}$ connects to  $V_{CC}$  when  $V_{CC}$  is greater than  $V_{BATT}$  and  $V_{CC}$  is above the reset threshold. When  $V_{CC}$  falls below  $V_{BATT}$  and  $V_{CC}$  is below the reset threshold,  $V_{OUT}$ connects to  $V_{BATT}$ . Connect a 0.1µF capacitor from  $V_{OUT}$  to GND.
- Pin 2  $V_{CC}$  +5V Supply Input
- Pin3 GND Ground reference for all signals
- Pin 4 PFI Power-Fail Input. This is the noninverting input to the power-fail comparator. When PFI is less than 1.25V,  $\overline{PFO}$  goes low. Connect PFI to GND or  $V_{OUT}$  when not used.
- Pin 5  $-\overline{PFO}$  Power-Fail Output.
- Pin 6 WDI Watchdog Input. WDI is a three level input. If WDI remains high or low for 1.6sec, the internal watchdog timer triggers a reset. If WDI is left floating or connected to a high-impedance tri-state buffer, the watchdog feature is disabled. The internal watchdog timer clears whenever reset is asserted.
- Pin 7 for SP690A/692A/802 only RESET (Active Low)–Reset Output. RESET Output goes low whenever  $V_{CC}$  falls below the reset threshold or whenever WDI remains high or low longer than 1.6 seconds. RESET remains low for 200ms after  $V_{CC}$  crosses the reset threshold voltage on power-up or after being triggered by WDI.

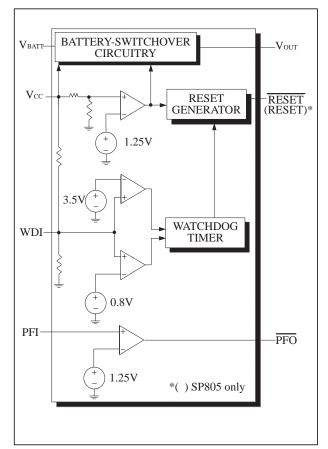


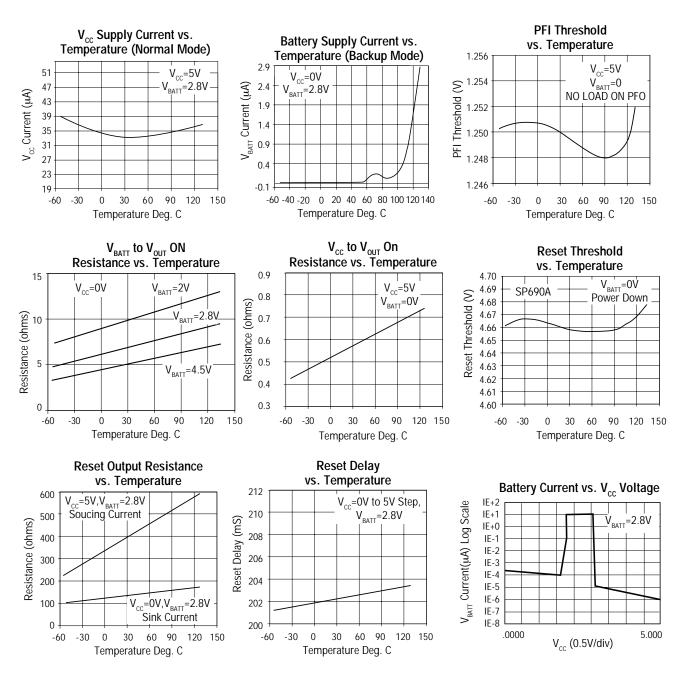
Figure 11. Internal Block Diagram

- Pin 7 for **SP805 only** RESET (Active High)– Reset Output is the inverse of RESET; when RESET is asserted, the RESET output voltage =  $V_{CC}$  or  $V_{BATT}$ , whichever is higher.
- $\begin{array}{l} \mbox{Pin 8} \mbox{V}_{BATT} \mbox{Backup-Battery Input. When} \\ \mbox{V}_{CC} \mbox{falls below the reset threshold, V}_{BATT} \\ \mbox{will be switched to V}_{OUT} \mbox{ if V}_{BATT} \mbox{ is } \\ \mbox{20mV greater than V}_{CC}. \mbox{ When V}_{CC} \mbox{ rises} \\ \mbox{20mV above V}_{BATT}, \mbox{ V}_{OUT} \mbox{ will be } \\ \mbox{ reconnected to V}_{CC}. \mbox{ The 40mV} \\ \mbox{ hysteresis prevents repeated switching if } \\ \mbox{V}_{CC} \mbox{ falls slowly.} \end{array}$

Date: 11/29/04

SP690A/692A Low Power Microprocessor Supervisory with Battery Switch-Over

#### **TYPICAL PERFORMANCE CHARACTERISTICS**



(25°C, unless otherwise noted)

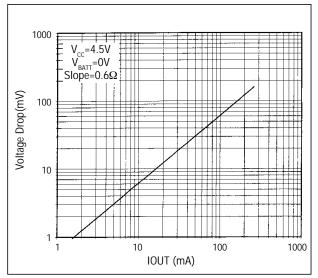


Figure 1.  $V_{\rm CC}$  to  $V_{\rm OUT}$  Vs. Output Current

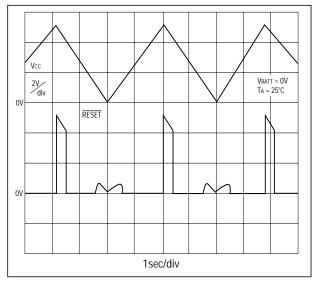


Figure 3A. SP690A RESET Output Voltage vs. Supply Voltage

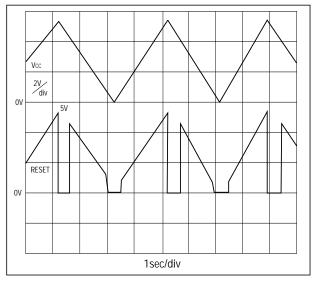


Figure 4A. SP805L RESET Output Voltage vs. Supply Voltage

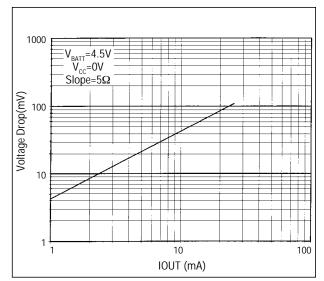


Figure 2. V<sub>BATT</sub> to V<sub>OUT</sub> Vs. Output Current

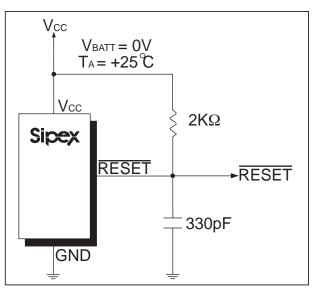


Figure 3B. Circuit for the SP690A/802L RESET Output Voltage vs. Supply Voltage

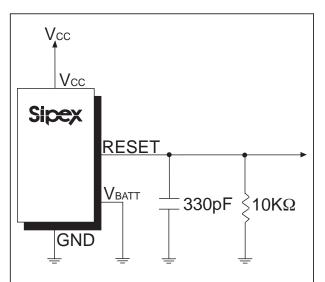


Figure 4B. Circuit for the SP805 RESET Output Voltage vs. Supply Voltage

Date: 11/29/04

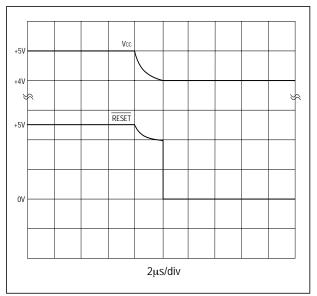


Figure 5A. SP690A RESET Response Time

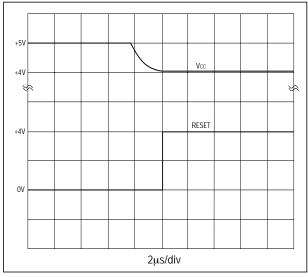


Figure 6A. SP805L RESET Response Time

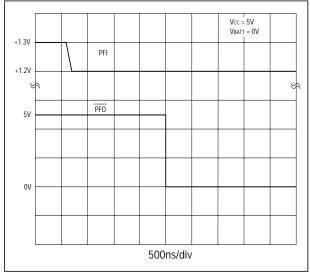


Figure 7A. Power-Fail Comparator Response Time (FALL)

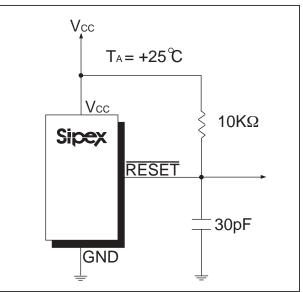


Figure 5B. Circuit for the SP690A/802L RESET Response Time

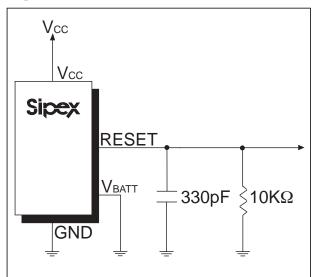


Figure 6B. Circuit for the SP805 RESET Response Time

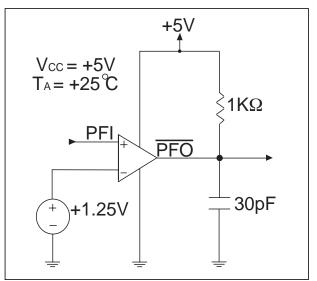
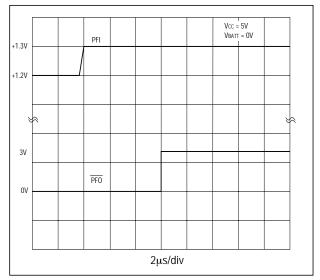


Figure 7B. Circuit for the Power-Fail Comparator Response Time (FALL)

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SP690A/692A Low Power Microprocessor Supervisory with Battery Switch-Over



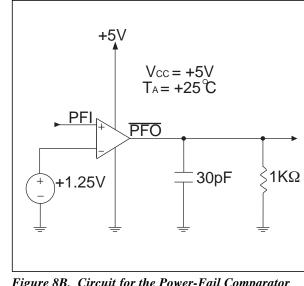


Figure 8A. Power-Fail Comparator Response Time (RISE)

Figure 8B. Circuit for the Power-Fail Comparator Response Time (RISE)

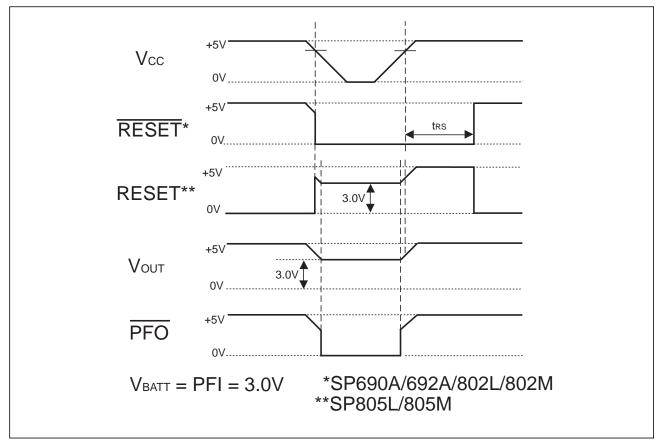


Figure 9. Timing Diagram

#### FEATURES

The **SP690A/692A/802L/802M/805L/805M** provide four key functions:

1. A battery backup switching for CMOS RAM, CMOS microprocessors, or other logic.

2. A reset output during power-up, power-down and brownout conditions.

3. A reset pulse if the optional watchdog timer has not been toggled within a specified time.

4. A 1.25V threshold detector for power-fail warning, low battery detection, or to monitor a power supply other than +5V.

The parts differ in their reset-voltage threshold levels and reset outputs. The **SP690A/802L/ 805L** generate a reset when the supply voltage drops below 4.65V. The **SP692A/802M/805M** generate a reset below 4.40V.

The **SP690A/692A/802L/802M/805L/805M** are ideally suited for applications in automotive systems, intelligent instruments, and battery-powered computers and controllers. All designs into an environment where it is critical to monitor the power supply to the  $\mu$ P and it's related digital components will find the **SSP690A/692A/802L/802M/805L/805M** ideal.

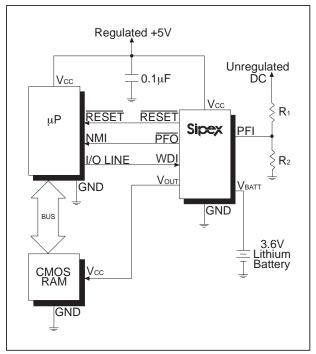


Figure 12. Typical Operating Circuit

#### THEORY OF OPERATION

The **SP690A/692A/802L/802M/805L/805M** microprocessor ( $\mu$ P) supervisory circuits monitor the power supplied to digital circuits such as microprocessors, microcontrollers, or memory. The series is an ideal solution for portable, battery-powered equipment that requires power supply monitoring. Implementing this series will reduce the number of components and overall complexity. The watchdog functions of this product family will continuously oversee the operational status of a system. The operational features and benefits of the **SP690A/692A/802L/802M/805L/805M** are described in more detail below.

#### **Reset Output**

The microprocessor's ( $\mu$ P's) reset input starts the  $\mu$ P in a known state. When the  $\mu$ P is in an unknown state, it should be held in reset. The **SP690A/SP692A/SP802** assert reset during power-up and prevent code execution errors during power-down or brownout conditions.

On power-up, once  $V_{CC}$  reaches 1V, RESET is guaranteed to be a logic low. As  $V_{CC}$  rises, RESET remains low. When  $V_{CC}$  exceeds the reset threshold, RESET will remain low for 200ms, *Figure 9*. If a brownout condition occurs and  $V_{CC}$  dips below the reset threshold, RESET is triggered. Each time RESET is triggered, it stays low for the reset pulse width interval. If a brownout condition interrupts a previously initiated reset pulse, the reset pulse continues for another 200ms. On power-down, once  $V_{CC}$  goes below the threshold, RESET is guaranteed to be logic low until  $V_{CC}$  drops below 1V.

**RESET** is also triggered by a watchdog timeout. If WDI remains either high or low for a period that exceeds the watchdog timeout period (1.6 sec), **RESET** pulses low for 200mS. As long as **RESET** is asserted, the watchdog timer remains clear. When **RESET** comes high, the watchdog resumes timing and must be serviced within 1.6sec. If WDI is tied high or low, a **RESET** pulse is triggered every 1.8sec ( $t_{WD}$  plus  $t_{RS}$ ). The **SP805L/M** active-high RESET output is the inverse of the **SP690A/SP692A/SP802** RE-SET output, and is valid with  $V_{CC}$  down to 1V. Some  $\mu$ P's, such as Intel's 80C51, require an active-high reset pulse.

#### Watchdog Input

The watchdog circuit monitors the  $\mu$ P's activity. If the  $\mu$ P does not toggle the watchdog input (WDI) within 1.6sec, a reset pulse is triggered. The internal 1.6sec timer is cleared by either a reset pulse or by floating the WDI input. As long as RESET is asserted or the WDI input is floating, the timer remains cleared and does not count. As soon as RESET is released and WDI is driven high or low, the timer starts counting. It can detect pulses as short as 50ns.

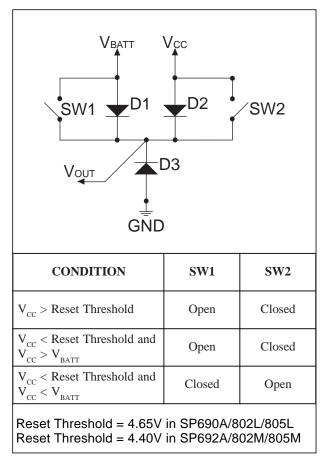


Figure 13. BACKUP-BATTERY Switchover Block Diagram

#### **Power-Fail Comparator**

The Power-Fail Comparator can be used as an under-voltage detector to signal the failing of a power supply (it is completely separate from the rest of the circuitry and does not need to be dedicated to this function). The PFI input is compared to an internal 1.25V reference. If PFI is less than 1.25V,  $\overline{PFO}$  goes low. The external voltage divider drives PFI to sense the unregulated DC input to the +5V regulator. The voltage-divider ratio can be chosen such that the voltage at PFI falls below 1.25V just before the +5V regulator drops out.  $\overline{PFO}$  then triggers an interrupt which signals the  $\mu P$  to prepare for power-down.

When  $V_{BATT}$  connects to  $V_{OUT}$ , the power-fail comparator is turned off and PFO is forced low to conserve backup-battery power.

#### **Backup-Battery Switchover**

In the event of a brownout or power failure, it may be necessary to preserve the contents of RAM. With a backup battery installed at  $V_{BATT}$ , the RAM is assured to have power if  $V_{CC}$  fails. As long as  $V_{CC}$  exceeds the reset threshold,  $V_{OUT}$  connects to  $V_{CC}$  through a 0.6 $\Omega$  PMOS power switch. Once  $V_{CC}$  falls below the reset threshold,  $V_{CC}$  or  $V_{BATT}$ , whichever is higher, switches to  $V_{OUT}$ .  $V_{BATT}$  connects to  $V_{OUT}$ through a 5 $\Omega$  switch only when  $V_{CC}$  is below the reset threshold and  $V_{BATT}$  is greater than  $V_{CC}$ .

When  $V_{CC}$  exceeds the reset threshold, it is connected to  $V_{OUT}$ , regardless of the voltage applied to  $V_{BATT}$  *Figure 13*. During this time, the diode (D1) between  $V_{BATT}$  and  $V_{OUT}$  will conduct current from  $V_{BATT}$  to  $V_{OUT}$  if  $V_{BATT}$  is more than .6V above  $V_{OUT}$ .

When  $V_{BATT}$  connects to  $V_{OUT}$ , backup mode is activated and the internal circuitry will be powered from the battery *Figure 14*. When  $V_{CC}$  is just below  $V_{BATT}$ , in the backup mode the current drawn from  $V_{BATT}$  will be typically  $30\mu A$ . When  $V_{CC}$  drops to more than 1V below  $V_{BATT}$ , the internal switchover comparator shuts off and the supply current falls to less than  $0.6\mu A$ .

SIGNAL	STATUS		
V <sub>cc</sub>	Disconnected from $V_{OUT}$		
V <sub>OUT</sub>	Connected to $V_{\text{BATT}}$ through an internal 8 $\Omega$ PMOS switch		
V <sub>batt</sub>	Connected to $V_{OUT}$ . Current drawn from the battery is less than 0.6µA, as long as $V_{CC} < V_{BATT}$ - 1V.		
PFI	Power-fail comparator is disabled.		
PFO	Logic low		
RESET	Logic low		
RESET	Logic high (SP805 only)		
WDI	Watchdog timer is disabled		

Figure 14. Input and Output Status in Battery-Backup Mode. To enter the Battery-Backup mode,  $V_{CC}$  must be less than the Reset threshold and less than  $V_{BATT}$ .

#### Using a High Capacity Capacitor as a Backup Power Source

VBATT has the same operating voltage range as VCC, and the battery-switchover threshold voltages are typically +20mV centered at VBATT, allowing use of a capacitor and a simple charging circuit as a backup source (see *Figure 16*).

PART NUMBER	MAXIMUM BACKUP-BATTERY VOLTAGE [V]
SP690A SP802L SP805L	4.80
SP692A SP802M SP805M	4.55

Figure 15. Allowable BACKUP-BATTERY Voltages

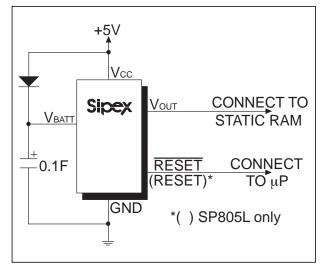


Figure 16. Backup Power Source Using High Capacity Capacitor with SP690A/802L/805L and a +5V ±5% Supply

If VCC is above the reset threshold and VBATT is 0.5V above VCC, current flows to VOUT and VCC from VBATT until the voltage at VBATT is less than 0.5V above VCC.

Leakage current through the capacitor charging diode and the **SP690A/SP802L/SP805L** internal power diode eventually discharges the capacitor to VCC. Also, if VCC and VBATT start from 0.5V above the reset threshold and power is lost at VCC, the capacitor on VBATT discharges through VCC until VBATT reaches the reset threshold; the **SP690A/SP802L/SP805L** then switches to battery-backup mode.

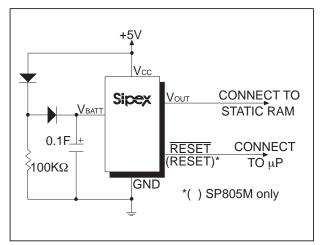


Figure 17. Backup Power Source Using High Capacity Capacitor with SP692A/802M/805M and a +5V ±10% Supply

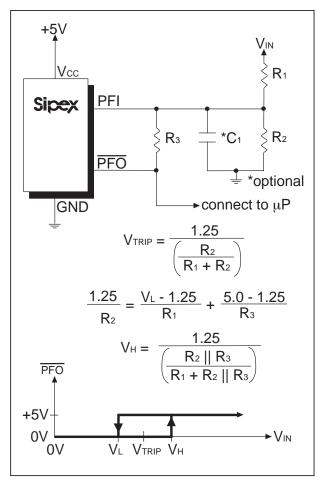


Figure 18. Adding Hysteresis to the POWER-FAIL Comparator

## Allowable Backup Power-Source Batteries

Lithium batteries work very well as backup batteries due to very low self-discharge rate and high energy density. Single lithium batteries with open-circuit voltages of 3.0V to 3.6V are ideal. Any battery with an open-circuit voltage less than the minimum reset threshold plus 0.3V can be connected directly to the  $V_{BATT}$  input of this series with no additional circuitry; see FIGURE 12. However, batteries with open-circuit voltages that are greater than this value cannot be used for backup, as current is sourced into  $V_{OUT}$  through the diode (*D1 in Figure 13*) when  $V_{CC}$  is close to the reset threshold.

# Operation Without a Backup Power Source

If a backup power source is not used, ground  $V_{BATT}$  and connect  $V_{OUT}$  to  $V_{CC}$ . Since there is no need to switch over to any backup power source,  $V_{OUT}$  does not need to be switched. A direct connection to  $V_{CC}$  eliminates any voltage drops across the switch which may push  $V_{OUT}$  below  $V_{CC}$ .

#### **Replacing the Backup Battery**

The backup battery can be removed while  $V_{CC}$  remains valid, without danger of triggering RESET/RESET. As long as  $V_{CC}$  stays above the reset threshold, battery-backup mode cannot be entered.

# Adding Hysteresis to the Power-Fail Comparator

Hysteresis adds a noise margin to the power-fail comparator and prevents repeated triggering of PFO when  $V_{IN}$  is close to its trip point. *Figure 18* shows how to add hysteresis to the power-fail comparator. Select the ratio of  $R_1$  and  $R_2$  such that PFI sees 1.25V when  $V_{IN}$  falls to its trip point ( $V_{TRIP}$ ).  $R_3$  adds the hysteresis. It will typically be an order of magnitude greater (about 10 times) than  $R_1$  or  $R_2$ . The current through  $R_1$ and  $R_2$  should be at least 1µA to ensure that the 25nA (max) PFI input current does not shift the trip point.  $R_3$  should be larger than 10K $\Omega$  so it does not load down the PFO pin. Capacitor C1 adds additional noise rejection.

#### Monitoring a Negative Voltage

The power-fail comparator can be used to monitor a negative supply rail using the circuit of *Figure* 19. When the negative rail is valid, PFO is low. When the negative supply voltage drops, PFO goes high. This circuit's accuracy is affected by the PFI threshold tolerance, the  $V_{CC}$ voltage, and the resistors, R1 and R2.

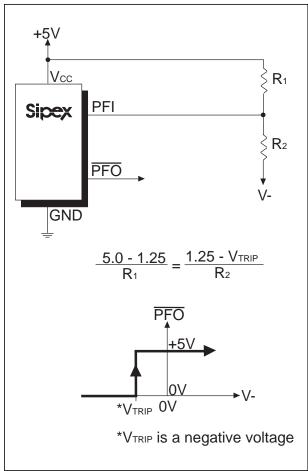


Figure 19. Monitoring a Negative Voltage

#### Interfacing to Microprocessors with Bidirectional Reset Pins

Microprocessors with bidirectional reset pins, such as the Motorola 68HC11 series, can contend with this series' RESET output. If, for example, the RESET output is driven high and the  $\mu$ P wants to pull it low, indeterminate logic levels may result. To correct this, connect a 4.7K $\Omega$  resistor between the RESET output and the  $\mu$ P reset I/O, as in *Figure 20*. Buffer the RESET output to other system components.

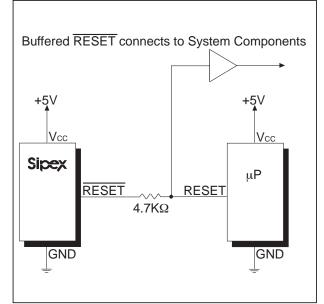
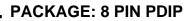
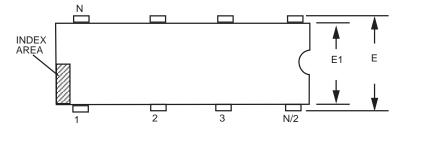
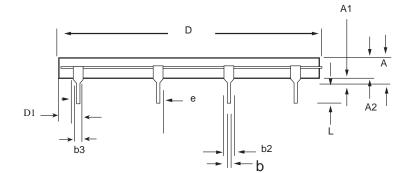
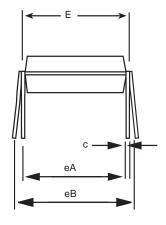


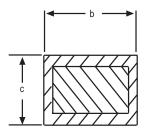
Figure 20. Interfacing to Microprocessors with Bidirectional RESET I/O





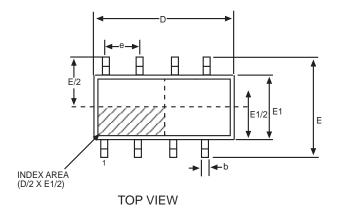


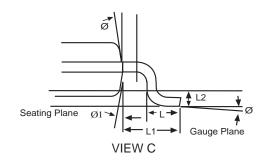


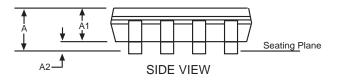


8 PIN PDIP JEDEC MS-001 (BA) Variation				
SYMBOL	MIN	NOM	MAX	
A	-	-	0.21	
A1	0.15	-	-	
A2	0.115	0.13	0.195	
b	0.014	0.018	0.022	
b2	0.045	0.06	0.07	
b3	0.3	0.039	0.045	
С	0.008	0.01	0.014	
D	0.355	0.365	0.4	
D1	0.005	-	-	
E	0.3	0.31	0.325	
E1	0.24	0.25	0.28	
e	.100 BSC			
eA	.300 BSC			
eB	-	_	0.43	
L	0.115	0.13	0.15	

Note: Dimensions in (mm)

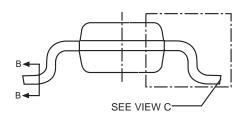


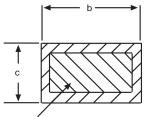




8 Pin NSOIC JEDEC MO-012 (AA) Variation				
SYMBOL	MIN	NOM	MAX	
A	1.35	-	1.75	
A1	0.1	-	0.25	
A2	1.25	-	1.65	
b	0.31	-	0.51	
с	0.17	-	0.24	
D	4.90 BSC			
E	6.00 BSC			
E1	3.90 BSC			
е	1.27 BSC			
L	0.4	-	1.27	
L1	1.04 REF			
L2	0.25 BSC			
ø	00	-	8°	
ø1	5°	-	15°	

Note: Dimensions in (mm)





BASE METAL SECTION B-B WITH PLATING

ORDERING INFORMATION			
Model	Temperature Range	Package Types	
SP690ACN	Temperature Range 0°C to +70°C	8-Pin NSOIC	
	0°C to +70°C		
SP690ACP	0°C to +70°C	8-Pin PDIP	
SP690AEN	40°C to +85°C	8-Pin NSOIC	
SP690AEN/TR	40°C to +85°C	8-Pin NSOIC	
	40°C to +85°C		
SP692ACN	0°C to +70°C	8-Pin NSOIC	
SP692ACN/TR	0°C to +70°C	8-Pin NSOIC	
SP692ACP	0°C to +70°C	8-Pin PDIP	
SP692AEN	40°C to +85°C	8-Pin NSOIC	
SP692AEN/TR	-40°C to +85°C	8-Pin NSOIC	
	-40°C to +85°C		
SP802LCN	0°C to +70°C	8-Pin NSOIC	
	0°C to +70°C		
	0°C to +70°C		
	40°C to +85°C		
	-40°C to +85°C		
	-40°C to +85°C		
01 002221			
SP802MCN	0°C to +70°C	8-Pin NSOIC	
	0°C to +70°C		
	0°C to +70°C		
	-40°C to +85°C		
	40°C to +85°C		
SP805LCN	0°C to +70°C	8-Pin NSOIC	
	0°C to +70°C		
	0°C to +70°C		
	-40°C to +85°C		
	40°C to +85°C		
	-40°C to +85°C		
SP805MCN	0°C to +70°C	8-Pin NSOIC	
	0°C to +70°C		
	0°C to +70°C		
	-40°C to +85°C		
	-40°C to +85°C		
	-40°C to +85°C		

Available in lead free packaging. To order add "-L" suffix to part number. Example: SP802LCN/TR = standard; SP802LCN-L/TR = lead free

/TR = Tape and Reel Pack quantity 2,500 for NSOIC.

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