

$\begin{array}{c} \textbf{AK5393}\\ \textbf{Enhanced Dual Bit } \Delta\Sigma \ 96 \textbf{kHz 24-Bit ADC} \end{array}$

GENERAL DESCRIPTION

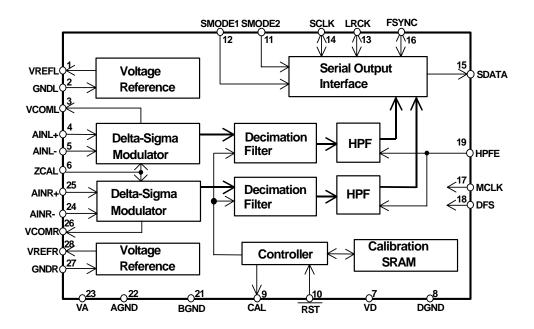
The AK5393 is a 24bit, 128x oversampling 2ch A/D Converter for professional digital audio systems. The modulator in the AK5393 uses the new developed Enhanced Dual Bit architecture. This new architecture achieves the wide dynamic range, while keeping much the same superior distortion characteristics as conventional Single Bit way. The AK5393 performs 117dB dynamic range, so the device is suitable for professional studio equipment such as digital mixer, digital VTR etc.

FEATURES

- Enhanced Dual Bit ADC
 Sampling Rate: 1kHz~108kHz
 Full Differential Insurts
- Full Differential Inputs
- □ S/(N+D): 105dB
- DR: 117dB
- □ S/N: 117dB
- □ High Performance Linear Phase Digital Anti-Alias filter
 - Passband: 0~21.768kHz(@fs=48kHz)
 - Ripple: 0.001dB
 - Stopband: 110dB

Digital HPF & Offset Calibration for Offset Cancel

- □ Power Supply: 5V±5%(Analog), 3~5.25V(Digital)
- □ Power Dissipation: 470mW
- □ Package: 28pin SOP
- □ AK5392 Pin compatible



- 1 -

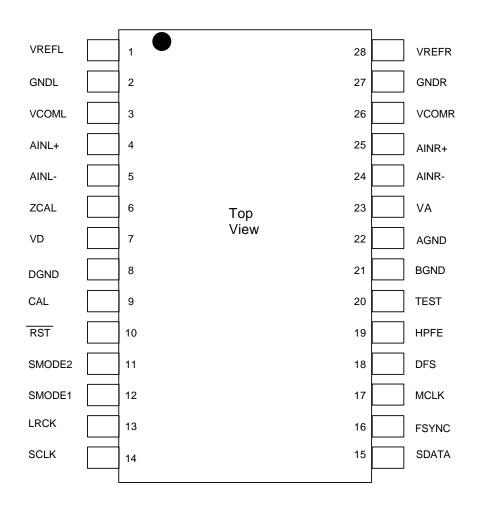
M0038-E-04

2000/4

Ordering Guide

AK5393-VS	$-10 \sim +70^{\circ}C$	28pin SOP
AKD5393	AK5393 Evaluat	ion Board

Pin Layout



■ Compatibility with AK5392

	AK5392	AK5393
Pin 18	CMODE	DFS
fs (max)	54kHz	108kHz
MCLK (DFS ="L"@fs=48kHz)	256fs/384fs	256fs
MCLK (DFS ="H"@fs=96kHz)	N/A	128fs

	PIN/FUNCTION							
No.	Pin Name	I/O	Function					
1	VREFL	0	Lch Reference Voltage Pin, 3.75V Normally connected to GNDL with a 10µF electrolytic capacitor and					
			a 0.1μ F ceramic capacitor.					
2	GNDL	-	Lch Reference Ground Pin, 0V					
3	VCOML	0	Lch Common Voltage Pin, 2.75V					
4	AINL+	Ι	Lch Analog positive input Pin					
5	AINL-	Ι	Lch Analog negative input Pin					
6	ZCAL	Ι	Zero Calibration Control Pin					
			This pin controls the calibration reference signal.					
			"L" :VCOML and VCOMR					
			"H" : Analog Input Pins (AINL±, AINR±)					
7	VD	-	Digital Power Supply Pin, 3.3V					
8	DGND	-	Digital Ground Pin, 0V					
9	CAL	0	Calibration Active Signal Pin					
			"H" means the offset calibration cycle is in progress. Offset calibration starts when RST goes "H". CAL goes "L" after 8704 LRCK cycles for DFS="L",					
			when KST goes T1. CAL goes L and 8704 LKCK cycles for DTS= L, 17408 LRCK cycles for DFS ="H".					
10	RST	Ι	Reset Pin					
	NO I		When "L", Digital section is powered-down. Upon returning "H", an					
			offset calibration cycle is started. An offset calibration cycle should always					
			be initiated after power-up.					
11	SMODE2	Ι	Serial Interface Mode Select Pin					
12	SMODE1	Ι	MSB first, 2's compliment.					
			SMODE2 SMODE1 MODE LRCK					
			L L Slave mode : MSB justified : H/L					
			L H Master mode : Similar to I^2S : H/L					
			H L Slave mode : I^2S : L/H					
12	LDCV	T/C	H H Master mode : I^2S : L/H					
13	LRCK	I/O	Left/Right Channel Select Clock Pin					
			LRCK goes "H" at SMODE2="L" and "L" at SMODE2="H" during reset when SMODE1 "H".					
			when SMODEL H.					

Downloaded from $\underline{Elcodis.com}$ electronic components distributor

14	SCLK	I/O	Serial Data Clock Pin
			Data is clocked out on the falling edge of SCLK.
			Slave mode:
			SCLK requires more than 48fs clock.
			Master mode:
			SCLK outputs a 128fs(DFS="L") or 64fs(DFS="H") clock.
			SCLK stays "L" during reset.
15	SDATA	0	Serial Data Output Pin
			MSB first, 2's complement. SDATA stays "L" during reset.
16	FSYNC	I/O	Frame Synchronization Signal Pin
			Slave mode:
			When "H", the data bits are clocked out on SDATA. In I ² S mode, FSYNC is
			don't care.
			Master mode:
			FSYNC outputs 2fs clock. FSYNC stays "L" during reset.
17	MCLK	Ι	Master Clock Input Pin
			256fs at DFS="L", 128fs at DFS="H".
18	DFS	Ι	Double Speed Sampling Mode Pin
			"L": Normal Speed
			"H": Double Speed
19	HPFE	Ι	High Pass Filter Enable Pin
			"L": Disable
			"H": Enable
20	TEST	Ι	Test Pin (pull-down pin)
			Should be connected to GND.
21	BGND	-	Substrate Ground Pin, 0V
22	AGND	-	Analog Ground Pin, 0V
23	VA	-	Analog Supply Pin, 5V
24	AINR-	Ι	Rch Analog negative input Pin
25	AINR+	Ι	Rch Analog positive input Pin
26	VCOMR	0	Rch Common Voltage Pin, 2.75V
27	GNDR	-	Rch Reference Ground Pin, 0V
28	VREFR	0	Rch Reference Voltage Pin, 3.75V
			Normally connected to GNDR with a $10\mu F$ electrolytic capacitor and a $0.1\mu F$
			ceramic capacitor

Note: All digital inputs should not be left floating.

Г

ABSOLUTE MAXIMUM RATINGS								
(AGND,BGND,DGND=0V; Note 1)								
Parameter	Symbol	min	max	Units				
Power Supplies: Analog	VA	-0.3	6.0	V				
Digital	VD	-0.3	6.0	V				
BGND-DGND (Note 2)	ΔGND	-	0.3	V				
Input Current, Any Pin Except Supplies	IIN	-	±10	mA				
Analog Input Voltage	VINA	-0.3	VA+0.3	V				
Digital Input Voltage	VIND	-0.3	VD+0.3	V				
Ambient Temperature (power applied)	Та	-10	70	°C				
Storage Temperature	Tstg	-65	150	°C				

Notes: 1. All voltages with respect to ground.

2. AGND, BGND and DGND must be connected to the same analog ground plane.

WARNING: Operation at or beyond these limits may result in permanent damage to the device. Normal operation is not guaranteed at these extremes.

	RECOMMENDED OPERATING CONDITIONS					
(AGND,BGND,DGND=0V; Note 1)						
Parameter Symbol min typ max Units						
Power Supplies:	Analog	VA	4.75	5.0	5.25	V
(Note 3)	Digital	VD	3.0	3.3	5.25	V

Notes: 1. All voltages with respect to ground.

3. The power up sequence between VA and VD is not critical.

* AKM assumes no responsibility for the usage beyond the conditions in this data sheet.

ANALOG CHARACTERISTICS

(Ta=25°C; VA=5.0V; VD=3.3V; AGND,BGND,DGND=0V; fs=48kHz; Signal Frequency=1kHz;

24bit Output; Measurement frequency=10Hz~20kHz; unless otherwise specified)

Parameter			min	typ	max	Units
Resolution	Resolution				24	Bits
Analog Input Characteristics:						• •
S/(N+D)	fs=48kHz	-1dBFS	98	105		dB
		-20dBFS	-	94		dB
		-60dBFS	-	54		dB
	fs=96kHz	-1dBFS	96	103		dB
	BW=40kHz	-20dBFS	-	85		dB
		-60dBFS	-	45		dB
Dynamic Range	e (-60dBFS with A-W	Veighted)	112	117		dB
S/N (A-	Weighted)		112	117		dB
Interchannel Iso	olation		110	120		dB
Interchannel Gain Mismatch			0.1	0.5	dB	
Gain Drift				150	ppm/°C	
Offset Error	after calibration, HI	PF=OFF		±200	±1000	LSB ₂₄
	after calibration, HI	PF=ON		±1		LSB ₂₄
Offset Drift	(HI	PF=OFF)	-	±10	-	LSB ₂₄ /°C
Offset Calibrati	on Range (H	PF=OFF)		±50		mV
Input Voltage	(AIN+)-(AIN-)		±2.3	±2.45	±2.6	V
Input Impedanc	e		2.4	4		kΩ
Power Supplies	S					
Power Supply C	Current					
VA				90	130	mA
VD (fs=48kHz; DFS="L")				6	9	mA
(fs=96kHz; DFS="H")		9	14	mA
Power Dissipati	on			470	680	mW
Power Supply R	Rejection (Note 4))		70		dB

Note: 4. DC to 26kHz. 110dB(typ) beyond 26kHz.

FILTER CHARACTERISTICS(fs=48kHz)							
(Ta=25°C; VA=5.0V±5%	; VD=3.0~5.	25V; fs=48	8kHz, DFS=	="L")			
Parameter			Symbol	min	typ	max	Units
ADC Digital Filter(Deci	mation LPF):					
Passband	()	Note 5)	PB	0		21.768	kHz
Stopband	1)	Note 5)	SB	26.232			kHz
Passband Ripple			PR			±0.001	dB
Stopband Attenuation	(N	Note 6)	SA	110			dB
Group Delay Distortion			ΔGD		0		us
Group Delay	(N	Note 7)	GD		38.7		1/fs
ADC Digital Filter(HPF):						
Frequency response	(Note 5)	-3dB	FR		1.0		Hz
		-0.1dB			6.5		Hz

FILTER CHARACTERISTICS(fs=96kHz)							
(Ta=25°C; VA=5.0V±5	%; VD=3.0/	~5.25V; fs	=96kHz, DF	S="H")			
Parameter			Symbol	min	typ	max	Units
ADC Digital Filter(Decir	nation LPF):					
Passband	(Note	: 5)	PB	0		43.536	kHz
Stopband	(Note	e 5)	SB	52.464			kHz
Passband Ripple			PR			±0.003	dB
Stopband Attenuation	(Note	8)	SA	110			dB
Group Delay Distortion		ΔGD		0		us	
Group Delay	(Note	7)	GD		38.8		1/fs
ADC Digital Filter(HPF)	:						
Frequency response (N	Note 5)	-3dB	FR		2.0		Hz
		-0.1dB			13.0		Hz

Notes: 5. The passband and stopband frequencies scale with fs.

- 6. The analog modulator samples the input at 6.144MHz for an output word rate of 48kHz. There is no rejection of input signals which are multiples of the sampling frequency (that is: there is no rejection for n x 6.144MHz \pm 21.768kHz, where n=1,2,3...).
- 7. The calculating delay time which occurred by digital filtering. This time is from the input of analog signal to setting the 24bit data of both channels to the output register. 40.7/fs(DFS="L"),40.8/fs(DFS="H")typ. at HPF:ON.
- 8. The analog modulator samples the input at 6.144MHz for an output word rate of 96kHz. There is no rejection of input signals which are multiples of the sampling frequency (that is: there is no rejection for n x 6.144MHz \pm 43.536kHz, where n=1,2,3...).

DIGITAL CHARACTERISTICS

$(Ta=25^{\circ}C; VA=5.0V\pm5\%; VD=3.0 \sim 5.25V)$	
--	--

Parameter	Symbol	min	typ	max	Units
High-Level Input Voltage	VIH	70%VD	-	-	V
Low-Level Input Voltage	VIL	-	-	30% VD	V
High-Level Output Voltage Iout=-20µA	VOH	VD-0.1	-	-	V
Low-Level Output Voltage Iout=20µA	VOL	-		0.1	V
Input Leakage Current	Iin	-	-	±10	μA

SWITCHING CHARACTERISTICS (Ta=25°C; VA=5.0V±5%; VD=3.0 ~ 5.25V; C_L=20pF) Parameter Symbol min typ max Units **Control Clock Frequency** Master Clock fCLK 0.256 256fs: 12.288 13.824 MHz Pulse width Low tCLKL 29 ns Pulse width High tCLKH 29 ns Serial Data Output Clock (SCLK) fSLK 6.144 6.912 MHz Channel Select Clock (LRCK) fs 1 48 108 kHz duty cycle 25 75 % **Serial Interface Timing** (Note 9) Slave Mode(SMODE1="L") tSLK 144.7 SCLK Period ns SCLK Pulse width Low tSLKL 65 ns Pulse width High tSLKH 65 ns SCLK falling to LRCK Edge (Note 10) tSLR -45 45 ns LRCK Edge to SDATA MSB Valid tDLR 45 ns SCLK falling to SDATA Valid tDSS 45 ns SCLK falling to FSYNC Edge tSF 45 -45 ns Master Mode(SMODE1="H") fSLK 128fs SCLK Frequency (DFS="L") Hz SCLK Frequency (DFS="H") **fSLK** 64fs Hz duty cycle 50 % FSYNC Frequency **fFSYNC** 2fs Hz duty cycle 50 % tSLR SCLK falling to LRCK Edge -20 20 ns LRCK Edge to FSYNC rising tLRF 1 tslk tDSS 45 SCLK falling to SDATA Valid ns SCLK falling to FSYNC Edge tSF -20 20 ns**Reset/Calibration timing** RST Pulse width tRTW 150 ns RST falling to CAL rising tRCR 50 ns 8704 RST rising to CAL falling (Note 11) tRCF 1/fsRST rising to SDATA Valid (Note 11) tRTV 8960 1/fs

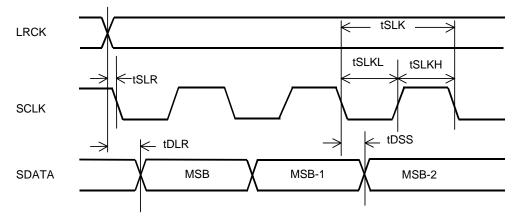
Notes: 9. Refer to Serial Data interface.

10. Specified LRCK edges not to coincide with the rising edges of SCLK.

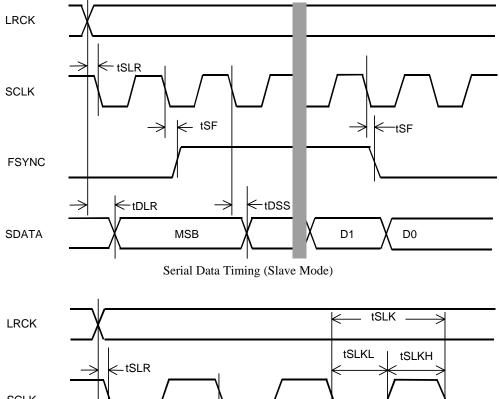
11. The number of the LRCK rising edges after RST brought high at DFS="L". The value is in master mode. In slave mode it becomes one LRCK clock(1/fs) longer. When DFS="H", tRCF=17408 and tRTV=17920.

M0038-E-04

Timing Diagram

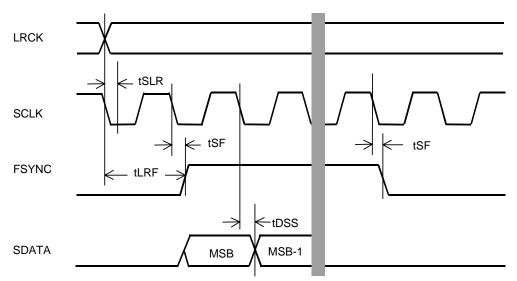


Serial Data Timing (Slave Mode, FSYNC="H")

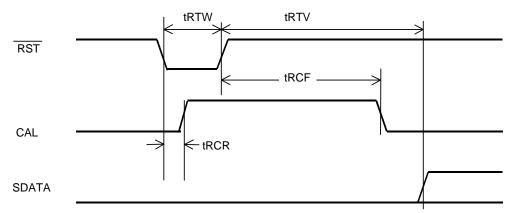


SCLK

Serial Data Timing (I²S Slave Mode, FSYNC = Don't Care)



Serial Data Timing (Master Mode & I²S Master Mode, DFS ="L")



Reset & Calibration Timing

OPERATION OVERVIEW

System Clock Input

The external clocks which are required to operate the AK5393 are MCLK, LRCK(fs), SCLK. MCLK should be synchronized with LRCK but the phase is free of care. MCLK should be 256fs in normal sampling mode(DFS="L") and double sampling mode needs 128fs as MCLK. Table 2 illustrates standard audio word rates and corresponding frequencies used in the AK5393.

As the AK5393 includes the phase detect circuit for LRCK, the AK5393 is reset automatically when the synchronization is out of phase by changing the clock frequencies. Therefore, the reset is only needed for power-up.

All external clocks must be present unless $\overline{RST} = "L"$, otherwise excessive current may result from abnormal operation of internal dynamic logic.

Speed	Normal(DFS ="L")	Double(DFS ="H")
LRCK (max)	54kHz	108kHz
SCLK	~128fs	~64fs
MCLK	256fs	128fs

fs	MCLK	SCLK
32.0kHz	8.1920MHz	4.0960MHz
44.1kHz	11.2896MHz	5.6448MHz
48.0kHz	12.2880MHz	6.1440MHz
96.0kHz	12.2880MHz	6.1440MHz

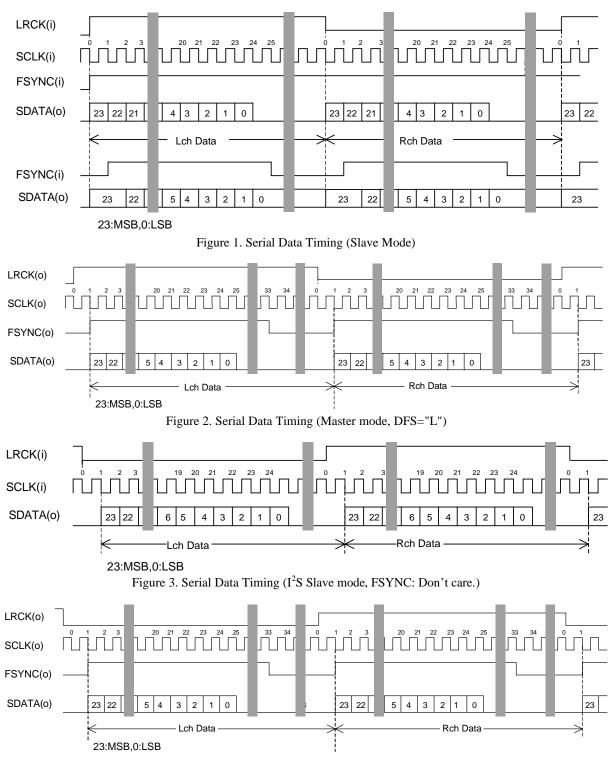
Table 2. Examples of System Clock Frequency

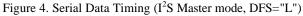
Serial Data Interface

The AK5393 supports four serial data formats which can be selected via SMODE1 and SMODE2 pins(Table 3). The data format is MSB-first, 2's complement.

Figure	SMODE2	SMODE1	Mode	LRCK
Figure 1	L	L	Slave Mode	Lch = H, Rch = L
Figure 2	L	Н	Master Mode	Lch =H, Rch =L
Figure 3	Н	L	I ² S Slave Mode	Lch =L, Rch =H
Figure 4	Н	Н	I ² S Master Mode	Lch =L, Rch =H

Table 3. Serial I/F Format





Offset Calibration

When RST pin goes to "L", the digital section is powered-down. Upon returning "H", an offset calibration cycle is started. An offset calibration cycle should always be initiated after power-up.

During the offset calibration cycle, the digital section of the part measures and stores the values of calibration input of each channel in registers. The calibration input value is subtracted from all future outputs. The calibration input may be obtained from either the analog input pins (AIN+/-) or the VCOM pins depending on the state of the ZCAL pin. With ZCAL "H", the analog input pin voltages are measured, and with ZCAL "L", the VCOM pin voltages are measured. The CAL output is "H" during calibration.

■ Digital High Pass Filter

The AK5393 also has a digital high pass filter for DC offset cancel. The cut-off frequency of the HPF is 1Hz at fs=48kHz and also scales with sampling rate(fs).

SYSTEM DESIGN

Figure 5 and 6 show the system connection diagram. An evaluation board[AKD5393] is available which demonstrates the optimum layout, power supply arrangements and measurement results.

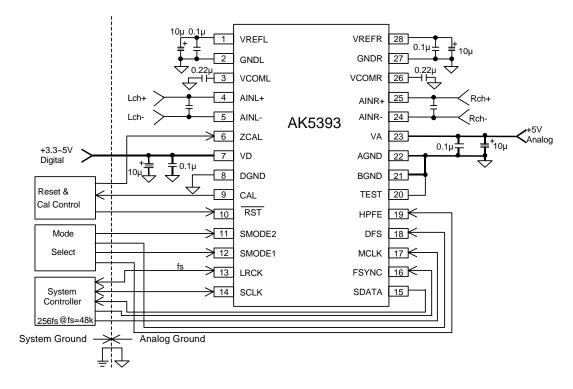


Figure 5. Typical Connection Diagram

Notes:

- LRCK = fs, SCLK=64fs.
- Power lines of VA and VD should be distributed separately from the point with low impedance of regulator etc.
- GND, BGND and DGND must be connected to the same analog ground plane.
- All input pins except pull-down/pull-up pins should not be left floating.

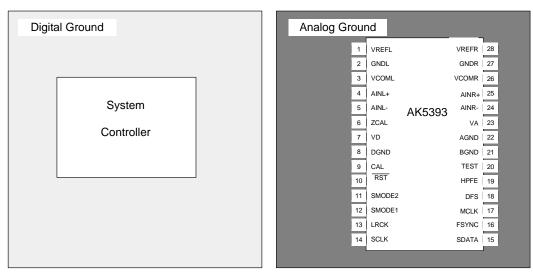


Figure 6 Ground layout

1. Grounding and Power Supply Decoupling

The AK5393 requires careful attention to power supply and grounding arrangements. Analog ground and digital ground should be separate and connected together near to where the supplies are brought onto the printed circuit board. Decoupling capacitors should be as near to the AK5393 as possible, with the small value ceramic capacitor being the nearest.

2. On-chip voltage reference and VCOM

The reference voltage for A/D converter is a differential voltage between the VREFL/R output voltage and the GNDL/R input voltage. The GNDL/R are connected to AGND and a 10 μ F electrolytic capacitor parallel with a 0.1 μ F ceramic capacitor between the VREFL/R and the GNDL/R eliminate the effects of high frequency noise. Especially a ceramic capacitor should be as near to the pins as possible. And all digital signals, especially clocks, should be kept away from the VREFL/R pins in order to avoid unwanted coupling into the AK5393. No load current may be taken from the VREFL/R pins.

VCOM is a common voltage of the analog signal. In order to eliminate the effects of high frequency noise, a 0.22μ F ceramic capacitor should be connected as near to the VCOM pin as possible. And all signals, especially clocks, should be kept away from the VCOM pin in order to avoid unwanted coupling into the AK5393. No load current may be drawn from the VCOM pin.

3. Analog Inputs

Analog signal is differentially input into the modulator via the AIN+ and the AIN- pins. The input voltage is the difference between AIN+ and AIN- pins. The full-scale of each pin is nominally ± 2.45 Vpp(typ). The AK5393 can accept input voltages from AGND to VA. The ADC output data format is 2's complement. The output code is 7FFFFH(@24bit) for input above a positive full scale and 800000H(@24bit) for input below a negative full scale. The ideal code is 000000H (@24bit) with no input signal. The DC offset is removed by the offset calibration.

The AK5393 samples the analog inputs at 128fs(6.144MHz @fs=48kHz,DFS="L"). The digital filter rejects noise above the stop band except for multiples of 128fs. A simple RC filter may be used to attenuate any noise around 128fs and most audio signals do not have significant energy at 128fs.

The AK5393 accepts +5V supply voltage. Any voltage which exceeds the upper limit of VA+0.3V and lower limit of AGND-0.3V and any current beyond 10mA for the analog input pins(AIN+/-) should be avoided. Excessive currents to the input pins may damage the device. Hence input pins must be protected from signals at or beyond these limits. Use caution specially in case of using $\pm 15V$ in other analog circuits.

Figure 7shows an input buffer circuit example 1. This is a full-differential input buffer circuit with an inverted-amp (gain :-10dB). The capacitor of 10nF between AIN+ /- decreases the clock feed through noise of modulator, and composes a 1st order LPF(fc=360kHz) with 22ohm resistor before the capacitor. This circuit also has a 1st order LPF(fc=370kHz) composed of op-amp. In this example, the internal offset is removed by self calibration. The evaluation board should be referred about the detail.

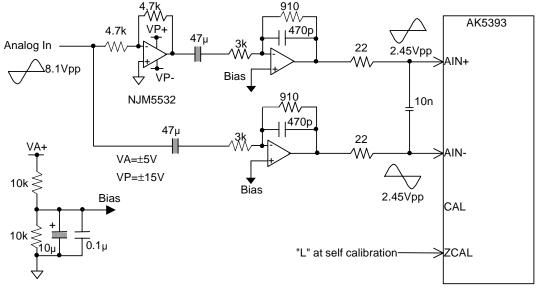


Figure 7 Differential Input Buffer Example 1

Figure 8 shows an input buffer circuit example 2. (1st order HPF; fc=0.66Hz, Table 4, 1st order LPF; fc=590kHz, gain=-14dB, Table 5). The analog signal is able to input through XLR or BNC connectors. (short JP1 and JP2 for BNC input, open JP1 and JP2 for XLR input). The input level of this circuit is +/-12.4Vpp (AK5393: +/-2.45Vpp Typ.).

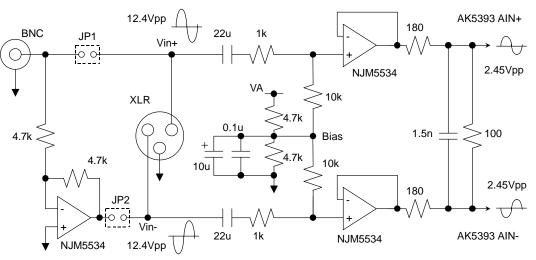


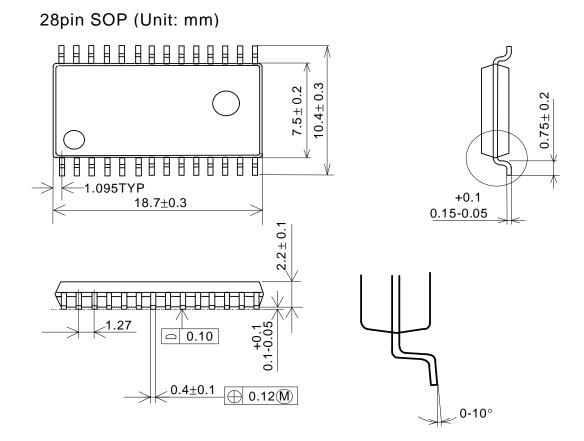
Figure 8 Differential Input Buffer Example 2

	fin	1F	łz	10Hz	
	Frequency Resp	onse -1.5	6dB -().02dB	
	Table 4. F	requency Resp	onse of HP	F	
	fin	20kHz	40kHz	6.144	MHz
Freg	uency Response	-0.005dB	-0.02dB	-15.0	6dB

Table 5. Frequency Response of LPF

M0038-E-04

PACKAGE



■ Package & Lead frame material

Package molding compound:	Epoxy
Lead frame material:	Cu
Lead frame surface treatment:	Solder plate

MARKING AKM JAPAN AK5393VS XXXBYYYYC U

Contents of XXXBYYYYC

XXXB:	Lot # (X : numbers, B : alphabet)
YYYYC:	Data Code (Y : numbers, C : alphabet)

IMPORTANT NOTICE

- These products and their specifications are subject to change without notice. Before considering any use or application, consult the Asahi Kasei Microsystems Co., Ltd. (AKM) sales office or authorized distributor concerning their current status.
- AKM assumes no liability for infringement of any patent, intellectual property, or other right in the application or use of any information contained herein.
- Any export of these products, or devices or systems containing them, may require an export license or other official approval under the law and regulations of the country of export pertaining to customs and tariffs, currency exchange, or strategic materials.
- AKM products are neither intended nor authorized for use as critical components in any safety, life support, or other hazard related device or system, and AKM assumes no responsibility relating to any such use, except with the express written consent of the Representative Director of AKM. As used here:
 - (a) A hazard related device or system is one designed or intended for life support or maintenance of safety or for applications in medicine, aerospace, nuclear energy, or other fields, in which its failure to function or perform may reasonably be expected to result in loss of life or in significant injury or damage to person or property.
 - (b) A critical component is one whose failure to function or perform may reasonably be expected to result, whether directly or indirectly, in the loss of the safety or effectiveness of the device or system containing it, and which must therefore meet very high standards of performance and reliability.
- It is the responsibility of the buyer or distributor of an AKM product who distributes, disposes of, or otherwise places the product with a third party to notify that party in advance of the above content and conditions, and the buyer or distributor agrees to assume any and all responsibility and liability for and hold AKM harmless from any and all claims arising from the use of said product in the absence of such notification.