

±250V Input Range G = 1, 10, Micropower, Difference Amplifier

FEATURES

- Pin Selectable Gain of 1 or 10
- High Common Mode Voltage Range: 85V Window ($V_S = 5V$, 0V) $\pm 250V$ ($V_S = \pm 15V$)
- Common Mode Rejection Ratio: 70dB Min
- Input Protection to ±350V
- Gain Error: 0.28% Max
- PSRR: 82dB Min
- High Input Impedance: 2MΩ Differential, 500kΩ Common Mode
- Micropower: 120µA Max Supply Current
- Wide Supply Range: 2.7V to 36V
- -3dB Bandwidth: 100kHz
- Rail-to-Rail Output
- 8-Pin SO Package

APPLICATIONS

- Battery Cell Voltage Monitoring
- High Voltage Current Sensing
- Signal Acquisition in Noisy Environments
- Input Protection
- Fault Protected Front Ends
- Level Sensing
- Isolation

DESCRIPTION

The LT $^{\circ}$ 1990 is a micropower precision difference amplifier with a very high common mode input voltage range. It has pin selectable gains of 1 or 10. The LT1990 operates over a ± 250 V common mode voltage range on a ± 15 V supply. The inputs are fault protected from common mode voltage transients up to ± 350 V and differential voltages up to ± 500 V. The LT1990 is ideally suited for both high side and low side current or voltage monitoring.

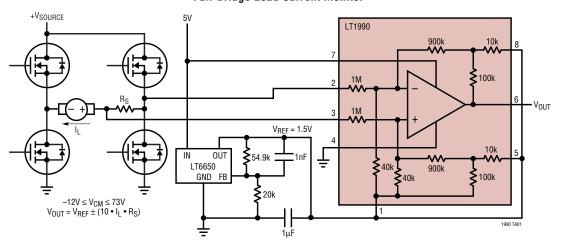
On a single 5V supply, the LT1990 has an adjustable 85V input range, 70dB min CMRR and draws less than 120μ A supply current. The rail-to-rail output maximizes the dynamic range, especially important for single supplies as low as 2.7V.

The LT1990 is specified for single 3V, 5V and ± 15 V supplies over both commercial and industrial temperature ranges. The LT1990 is available in the 8-pin SO package.

T, LT, LTC and LTM are registered trademarks of Linear Technology Corporation. All other trademarks are the property of their respective owners.

TYPICAL APPLICATION

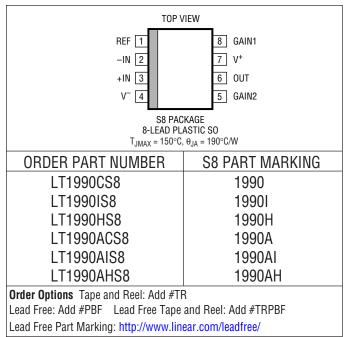
Full-Bridge Load Current Monitor



ABSOLUTE MAXIMUM RATINGS

(Notes 1, 2)	
Total Supply Voltage (V+ to V ⁻)	36V
Input Voltage Range	
Continuous	±250V
Transient (0.1s)	±350V
Differential	±500V
Output Short-Circuit Duration (Note 3)	Indefinite
Operating Temperature Range (Note 4)	
LT1990C	–40°C to 85°C
LT1990I	–40°C to 85°C
LT1990H	-40°C to 125°C
Specified Temperature Range (Note 5)	
LT1990C	
LT1990I	–40°C to 85°C
LT1990H	-40°C to 125°C
Junction Temperature	150°C
Storage Temperature Range	−65°C to 150°C

PACKAGE/ORDER INFORMATION



Consult LTC Marketing for parts specified with wider operating temperature ranges.

3V/5V ELECTRICAL CHARACTERISTICS

Lead Temperature (Soldering, 10 sec.)......300°C

 $V_S = 3V$, 0V; $V_S = 5V$, 0V; $R_L = 10k$, $V_{CM} = V_{REF} = half supply$, G = 1, 10, $T_A = 25$ °C, unless otherwise noted. (Note 6)

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
G	Gain	Pins 5 and 8 = Open Pins 5 and 8 = V _{REF}		1 10		
ΔG	Gain Error	V _{OUT} = 0.5V to (+Vs) -0.75V LT1990, G = 1 LT1990A, G = 1 G = 10, V _S = 5V, 0V		0.4 0.07 0.2	0.6 0.28 0.8	% % %
GNL	Gain Nonlinearity	V _S = 5V, 0V; V _{OUT} = 0.5V to 4.25V G = 1 G = 10		0.001 0.01	0.005	% %
V _{CM}	Input Voltage Range	Guaranteed by CMRR $V_S = 3V$, 0V; $V_{REF} = 1.25V$ $V_S = 5V$, 0V; $V_{REF} = 1.25V$ $V_S = 5V$, 0V; $V_{REF} = 2.5V$	-5 -5 -38		25 80 47	V V V
CMRR	Common Mode Rejection Ratio RTI (Referred to Input)	$V_S = 3V$, 0V (Note 7) $V_{CM} = -5V$ to 25V, $V_{REF} = 1.25V$ LT1990 LT1990A	60 70	68 75		dB dB
		$V_S = 5V$, 0V $V_{CM} = -5V$ to 80V, $V_{REF} = 1.25V$ LT1990 LT1990A	60 70	68 75		dB dB
		$V_S = 5V$, 0V (Note 7) $V_{CM} = -38V$ to 47V, $V_{REF} = 2.5V$ LT1990 LT1990A	60 70	68 75		dB dB



 $V_S = 3V$, 0V; $V_S = 5V$, 0V; $R_L = 10k$, $V_{CM} = V_{REF} = half$ supply, G = 1, 10, $T_A = 25$ °C, unless otherwise noted. (Note 6)

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
V _{OS}	Offset Voltage, RTI	G = 1, 10		0.8	3	mV
e _n	Input Noise Voltage, RTI	f ₀ = 0.1Hz to 10Hz		22		μV _{P-P}
	Noise Voltage Density, RTI	$f_0 = 1kHz$		1		μV/√Hz
R _{IN}	Input Resistance	Differential Common Mode		2 0.5		MΩ MΩ
PSRR	Power Supply Rejection Ratio, RTI	$V_S = 2.7V$ to 12.7V, $V_{CM} = V_{REF} = 1.25V$	80	92		dB
	Minimum Supply Voltage	Guaranteed by PSRR		2.4	2.7	V
Is	Supply Current	(Note 8)		105	120	μА
V_{OL}	Output Voltage Swing LOW	-IN = V ⁺ , +IN = Half Supply (Note 8)		30	50	mV
V _{OH}	Output Voltage Swing HIGH	-IN = 0V, $+IN = Half SupplyV_S = 3V, 0V, Below V^+V_S = 5V, 0V, Below V^+$		100 120	150 175	mV mV
I _{SC}	Output Short-Circuit Current	Short to GND (Note 9) Short to V ⁺ (Note 9)	4 13	8 20		mA mA
BW	Bandwidth (–3dB)	G = 1 G = 10		100 6.5		kHz kHz
SR	Slew Rate	$G = 1$, $V_S = 5V$, $0V$, $V_{OUT} = 0.5V$ to $4.5V$		0.5		V/µs
	Settling Time to 0.01%	4V Step, G = 1, V _S = 5V, 0V		45		μЅ
AV _{REF}	Reference Gain to Output	G = 1 G = 10		1 ± 0.000 1 ± 0.000		

The ullet denotes the specifications which apply over the temperature range of $0^{\circ}C \leq T_A \leq 70^{\circ}C$. $V_S = 3V$, 0V; $V_S = 5V$, 0V; $R_L = 10k$, $V_{CM} = V_{REF} = half$ supply, G = 1, 10, unless otherwise noted. (Notes 4, 6)

SYMBOL	PARAMETER	CONDITIONS		LT1990C/LT1990I Min typ max			UNITS
ΔG	Gain Error	$V_{OUT} = 0.5V \text{ to } (+V_S) - 0.75V$ LT1990, G = 1 LT1990A, G = 1 G = 10	•			0.65 0.33 0.90	% % %
G/T	Gain vs Temperature	G = 1 (Note 10) G = 10 (Note 10)	•		2 7	10 20	ppm/°C ppm/°C
V _{CM}	Input Voltage Range	Guaranteed by CMRR $V_S = 3V$, $0V$, $V_{REF} = 1.25V$ $V_S = 5V$, $0V$, $V_{REF} = 1.25V$ $V_S = 5V$, $0V$, $V_{REF} = 2.5V$	•	-5 -5 -37		25 80 48	V V V
CMRR	Common Mode Rejection Ratio, RTI	V _S = 3V, 0V (Note 7) V _{CM} = -5V to 25V, V _{REF} = 1.25V LT1990 LT1990A		58 68			dB dB
		$V_S = 5V$, 0V $V_{CM} = -5V$ to 80V, $V_{REF} = 1.25V$ LT1990 LT1990A	:	58 68			dB dB
		V_S = 5V, 0V (Note 7) V_{CM} = -38V to 47V, V_{REF} = 2.5V LT1990 LT1990A	:	58 68			dB dB



The ullet denotes the specifications which apply over the temperature range of $0^{\circ}C \leq T_A \leq 70^{\circ}C$. $V_S = 3V$, 0V; $V_S = 5V$, 0V; $R_L = 10k$, $V_{CM} = V_{REF} = half$ supply, G = 1, 10, unless otherwise noted. (Notes 4, 6)

				LT1	990C/LT1	9901	
SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
V _{OS}	Input Offset Voltage, RTI	V _S = 3V, 0V G = 1, 10	•			4.1	mV
		V _S = 5V, 0V G = 1, 10	•			4.1	mV
V _{OS} /T	Input Offset Voltage Drift, RTI	(Note 10)	•		5	22	μV/°C
V _{OSH}	Input Offset Voltage Hysteresis, RTI	(Note 11)	•		230		μV
PSRR	Power Supply Rejection Ratio, RTI	V _S = 2.7V to 12.7V V _{CM} = V _{REF} = 1.25V G = 1, 10	•	78			dB
	Minimum Supply Voltage	Guaranteed by PSRR	•			2.7	V
I _S	Supply Current	(Note 8)	•			150	μА
V_{OL}	Output Voltage Swing LOW	-IN = V ⁺ , +IN = Half Supply (Note 8)	•			60	mV
V _{OH}	Output Voltage Swing HIGH	-IN = 0V, $+IN = Half SupplyV_S = 3V, 0V, Below V^+V_S = 5V, 0V, Below V^+$	•			180 205	mV mV
I _{SC}	Output Short-Circuit Current	Short to GND (Note 9) Short to V ⁺ (Note 9)	•	3 11			mA mA

The ullet denotes the specifications which apply over the temperature range of $-40^{\circ}C \leq T_A \leq 85^{\circ}C$. $V_S = 3V$, 0V; $V_S = 5V$, 0V; $R_L = 10k$, $V_{CM} = V_{REF} = half$ supply, G = 1, 10, unless otherwise noted. (Notes 4, 6)

OVERDOL				LT1	LT1990C/LT1990I		
SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
ΔG	Gain Error	V _{OUT} = 0.5V to (+V _S) - 0.75V LT1990, G = 1 LT1990A, G = 1 G = 10	•			0.67 0.35 0.95	% % %
G/T	Gain vs Temperature	G = 1 (Note 10) G = 10 (Note 10)	•		2 7	10 20	ppm/°C ppm/°C
V _{CM}	Input Voltage Range	Guaranteed by CMRR $V_S = 3V$, 0V, $V_{REF} = 1.25V$ $V_S = 5V$, 0V, $V_{REF} = 1.25V$ $V_S = 5V$, 0V, $V_{REF} = 2.5V$	•	-5 -5 -37		25 80 48	V V V
CMRR	Common Mode Rejection Ratio, RTI	V _S = 3V, 0V (Note 7) V _{CM} = -5V to 25V, V _{REF} = 1.25V LT1990 LT1990A	7)			dB dB	
		V _S = 5V, 0V V _{CM} = -5V to 80V, V _{REF} = 1.25V LT1990 LT1990A	•	57 67			dB dB
		V _S = 5V, 0V (Note 7) V _{CM} = -38V to 47V, V _{REF} = 2.5V LT1990 LT1990A	•	57 67			dB dB

LINEAR TECHNOLOGY

The ullet denotes the specifications which apply over the temperature range of $-40^{\circ}C \leq T_A \leq 85^{\circ}C$. $V_S = 3V$, 0V; $V_S = 5V$, 0V; $R_L = 10k$, $V_{CM} = V_{REF} = half$ supply, G = 1, 10, unless otherwise noted. (Notes 4, 6)

SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
V _{OS}	Input Offset Voltage, RTI	V _S = 3V, 0V					
		G = 1, 10	•			4.5	mV
		$V_{S} = 5V, 0V$					
		G = 1, 10	•			4.5	mV
V _{OS} /T	Input Offset Voltage Drift, RTI	(Note 10)	•		5	22	μV/°C
V _{OSH}	Input Offset Voltage Hysteresis, RTI	(Note 11)	•		230		μV
PSRR	Power Supply Rejection Ratio, RTI	V _S = 2.7V to 12.7V					
		V _{CM} = V _{REF} = 1.25V	•	76			dB
	Minimum Supply Voltage	Guaranteed by PSRR	•			2.7	V
Is	Supply Current	(Note 8)	•			170	μА
V_{OL}	Output Voltage Swing LOW	-IN = V ⁺ , +IN = Half Supply (Note 8)	•			70	mV
V _{OH}	Output Voltage Swing HIGH	-IN = 0V, +IN = Half Supply					
		$V_S = 3V$, 0V, Below V ⁺	•			200	mV
		$V_S = 5V$, 0V, Below V^+	•			275	mV
I _{SC}	Output Short-Circuit Current	Short to GND (Note 9)	•	2			mA
		Short to V ⁺ (Note 9)	•	8			mA

3V/5V ELECTRICAL CHARACTERISTICS

The ullet denotes the specifications which apply over the temperature range of $-40^{\circ}C \le T_A \le 125^{\circ}C$. $V_S = 3V$, 0V; $V_S = 5V$, 0V; $R_L = 10k$, $V_{CM} = V_{REF} = half$ supply, G = 1, 10, unless otherwise noted. (Notes 4, 6)

SYMBOL	PARAMETER	CONDITIONS		MIN	LT1990H TYP	MAX	UNITS
ΔG	Gain Error	V _{OUT} = 0.5V to (+V _S) - 0.75V LT1990, G = 1 LT1990A, G = 1 G = 10	•	- MIIN		0.69 0.37 0.97	% % %
G/T	Gain vs Temperature	G = 1 (Note 10) G = 10 (Note 10)	•		2 7	10 20	ppm/°C ppm/°C
V _{CM}	Input Voltage Range	Guaranteed by CMRR $V_S = 3V$, 0V, $V_{REF} = 1.25V$ $V_S = 5V$, 0V, $V_{REF} = 1.25V$ $V_S = 5V$, 0V, $V_{REF} = 2.5V$	•	-5 -5 -37		25 80 48	V V V
CMRR	Common Mode Rejection Ratio, RTI	V _S = 3V, 0V (Note 7) V _{CM} = -5V to 25V, V _{REF} = 1.25V LT1990 LT1990A	•	56 66			dB dB
		$V_S = 5V$, 0V $V_{CM} = -5V$ to 80V, $V_{REF} = 1.25V$ LT1990 LT1990A	•	56 66			dB dB
		$V_S = 5V$, 0V (Note 7) $V_{CM} = -38V$ to 47V, $V_{REF} = 2.5V$ LT1990 LT1990A	•	56 66			dB dB



The ullet denotes the specifications which apply over the temperature range of $-40^{\circ}C \le T_A \le 125^{\circ}C$. $V_S = 3V$, 0V; $V_S = 5V$, 0V; $R_L = 10k$, $V_{CM} = V_{REF} = half$ supply, G = 1, 10, unless otherwise noted. (Notes 4, 6)

SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
V _{OS}	Input Offset Voltage, RTI	V _S = 3V, 0V					
		G = 1, 10	•			5.2	mV
		$V_S = 5V, 0V$					
		G = 1, 10	•			5.2	mV
V _{OS} /T	Input Offset Voltage Drift, RTI	(Note 10)	•		5	22	μV/°C
V _{OSH}	Input Offset Voltage Hysteresis, RTI	(Note 11)	•		250		μV
PSRR	Power Supply Rejection Ratio, RTI	V _S = 2.7V to 12.7V					
		$V_{CM} = V_{REF} = 1.25V$	•	75			dB
	Minimum Supply Voltage	Guaranteed by PSRR	•			2.7	V
Is	Supply Current	(Note 8)	•			200	μА
V_{OL}	Output Voltage Swing LOW	-IN = V ⁺ , +IN = Half Supply (Note 8)	•			80	mV
V _{OH}	Output Voltage Swing HIGH	-IN = 0V, +IN = Half Supply					
		$V_S = 3V$, 0V, Below V ⁺	•			230	mV
		$V_S = 5V$, 0V, Below V ⁺	•			275	mV
I _{SC}	Output Short-Circuit Current	Short to GND (Note 9)	•	1			mA
		Short to V ⁺ (Note 9)	•	5			mA

±15V ELECTRICAL CHARACTERISTICS

 $V_S = \pm 15V$, $R_L = 10k$, $V_{CM} = V_{REF} = 0V$, G = 1, $I_A = 25$ °C, unless otherwise noted. (Note 6)

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
G	Gain	Pins 5 and 8 = Open Pins 5 and 8 = V _{REF}		1 10		
ΔG	Gain Error	V _{OUT} = ±10V LT1990, G = 1 LT1990A, G = 1 G = 10		0.4 0.07 0.2	0.6 0.28 0.8	% % %
GNL	Gain Nonlinearity	V _{OUT} = ±10V G = 1 G = 10		0.0008 0.005	0.002 0.02	%
V _{CM}	Input Voltage Range	Guaranteed by CMRR	-250		250	V
CMRR	Common Mode Rejection Ratio, RTI	V _{CM} = -250V to 250V LT1990 LT1990A	60 70	68 75		dB dB
V _{OS}	Offset Voltage, RTI	G = 1, 10		0.9	5.2	mV
e _n	Input Noise Voltage, RTI	f ₀ = 0.1Hz to 10Hz		22		μV _{P-P}
	Noise Voltage Density, RTI	$f_0 = 1kHz$		1		μV/√Hz
R _{IN}	Input Resistance	Differential Common Mode		2 0.5		MΩ MΩ
PSRR	Power Supply Rejection Ratio, RTI	$V_S = \pm 1.35 V$ to $\pm 18 V$	82	100		dB
	Minimum Supply Voltage	Guaranteed by PSRR		±1.2	±1.35	V
Is	Supply Current			140	180	μА
V _{OUT}	Output Voltage Swing		±14.5	±14.79		V





 V_S = $\pm 15V$, R_L = 10k, V_{CM} = V_{REF} = 0V, G = 1, 10, T_A = 25°C, unless otherwise noted. (Note 6)

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
I _{SC}	Output Short-Circuit Current	Short to V ⁻ Short to V ⁺	6 15	9 22		mA mA
BW	Bandwidth	G = 1 G = 10		105 7		kHz kHz
SR	Slew Rate	G = 1, V _{OUT} = ±10V	0.3	0.55		V/µs
	Settling Time to 0.01%	10V Step, G = 1		60		μS
AV _{REF}	Reference Gain to Output	G = 1 G = 10		1 ± 0.000 1 ± 0.007		

The ullet denotes the specifications which apply over the temperature range of 0°C \leq T_A \leq 70°C. V_S = \pm 15V, R_L = 10k, V_{CM} = V_{REF} = 0V, G = 1, 10, unless otherwise noted. (Notes 4, 6)

				LT19	990C/LT	19901	
SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
ΔG	Gain Error	V _{OUT} = ±10V LT1990, G = 1 LT1990A, G = 1 G = 10	•			0.65 0.33 0.9	% % %
GNL	Gain Nonlinearity	V _{OUT} = ±10V G = 1 G = 10	•			0.0025 0.025	% %
G/T	Gain vs Temperature	G = 1 (Note 10) G = 10 (Note 10)	•		2 7	10 20	ppm/°C ppm/°C
V _{CM}	Input Voltage Range	Guaranteed by CMRR	•	-250		250	V
CMRR	Common Mode Rejection Ratio, RTI	V _{CM} = -250V to 250V LT1990 LT1990A	:	59 68			dB dB
V _{OS}	Input Offset Voltage, RTI	G = 1, 10	•			6.2	mV
V _{OS} /T	Input Offset Voltage Drift, RTI	(Note 10)	•		5	22	μV/°C
V _{OSH}	Input Offset Voltage Hysteresis, RTI	(Note 11)	•		250		μV
PSRR	Power Supply Rejection Ratio, RTI	$V_S = \pm 1.35 V \text{ to } \pm 18 V$	•	80			dB
	Minimum Supply Voltage	Guaranteed by PSRR	•			±1.35	V
Is	Supply Current		•			230	μА
V _{OUT}	Output Voltage Swing		•	±14.4			V
I _{SC}	Output Short-Circuit Current	Short to V ⁺	•	5 13			mA mA
SR	Slew Rate	G = 1, V _{OUT} = ±10V	•	0.25			V/µs

The ullet denotes the specifications which apply over the temperature range of $-40^{\circ}C \leq T_A \leq 85^{\circ}C$. $V_S = \pm 15V$, $R_L = 10k$, $V_{CM} = V_{REF} = 0V$, G = 1, 10, unless otherwise noted. (Notes 4, 6)

SYMBOL	PARAMETER	CONDITIONS		LT1 MIN	990C/LT	1990I MAX	UNITS
ΔG	Gain Error	V _{OUT} = ±10V LT1990, G = 1 LT1990A, G = 1 G = 10	•	MIN	- 1117	0.67 0.35 0.95	% % %
GNL	Gain Nonlinearity	V _{OUT} = ±10V G = 1 G = 10	•			0.003 0.03	% %
G/T	Gain vs Temperature	G = 1 (Note 10) G = 10 (Note 10)	•		2 7	10 20	ppm/°C
V _{CM}	Input Voltage Range	Guaranteed by CMRR	•	-250		250	V
CMRR	Common Mode Rejection Ratio, RTI	V _{CM} = -250V to 250V LT1990 LT1990A	•	58 67			dB dB
V _{OS}	Input Offset Voltage, RTI	G = 1, 10	•			6.7	mV
V _{OS} /T	Input Offset Voltage Drift, RTI	(Note 10)	•		5	22	μV/°C
V _{OSH}	Input Offset Voltage Hysteresis, RTI	(Note 11)	•		250		μV
PSRR	Power Supply Rejection Ratio, RTI	$V_S = \pm 1.35 V \text{ to } \pm 18 V$	•	78			dB
	Minimum Supply Voltage	Guaranteed by PSRR	•			±1.35	V
Is	Supply Current		•			280	μА
V _{OUT}	Output Voltage Swing		•	±14.3			V
I _{SC}	Output Short-Circuit Current	Short to V ⁻ Short to V ⁺	•	3 10			mA mA
SR	Slew Rate	G = 1, V _{OUT} = ±10V	•	0.2			V/µs

±15V ELECTRICAL CHARACTERISTICS

The ullet denotes the specifications which apply over the temperature range of $-40^{\circ}C \leq T_A \leq 125^{\circ}C$. $V_S = \pm 15V$, $R_L = 10k$, $V_{CM} = V_{REF} = 0V$, G = 1, 10, unless otherwise noted. (Notes 4, 6)

SYMBOL	PARAMETER	CONDITIONS		MIN	LT1990 TYP	H MAX	UNITS
ΔG	Gain Error	V _{OUT} = ±10V LT1990, G = 1 LT1990A, G = 1 G = 10	•			0.69 0.37 0.97	% % %
GNL	Gain Nonlinearity	V _{OUT} = ±10V G = 1 G = 10	•			0.0035 0.035	%
G/T	Gain vs Temperature	G = 1 (Note 10) G = 10 (Note 10)	•		2 7	10 20	ppm/°C ppm/°C
V _{CM}	Input Voltage Range	Guaranteed by CMRR	•	-250		250	V
CMRR	Common Mode Rejection Ratio, RTI	V _{CM} = -250V to 250V LT1990 LT1990A	•	57 66			dB dB
V _{OS}	Input Offset Voltage, RTI	G = 1, 10	•			7.4	mV
V _{OS} /T	Input Offset Voltage Drift, RTI	(Note 10)	•		5	22	μV/°C



The ullet denotes the specifications which apply over the temperature range of $-40^{\circ}C \le T_A \le 125^{\circ}C$. $V_S = \pm 15V$, $R_L = 10k$, $V_{CM} = V_{REF} = 0V$, G = 1, 10, unless otherwise noted. (Notes 4, 6)

SYMBOL	PARAMETER	CONDITIONS		MIN	LT1990H TYP MAX	UNITS
V _{OSH}	Input Offset Voltage Hysteresis, RTI	(Note 11)	•		250	μV
PSRR	Power Supply Rejection Ratio, RTI	$V_S = \pm 1.35 V \text{ to } \pm 18 V$	•	77		dB
	Minimum Supply Voltage	Guaranteed by PSRR	•		±1.35	V
Is	Supply Current		•		330	μΑ
V _{OUT}	Output Voltage Swing		•	±14.2		V
I _{SC}	Output Short-Circuit Current	Short to V ⁻ Short to V ⁺	•	1.5 7		mA mA
SR	Slew Rate	G = 1, V _{OUT} = ±10V	•	0.1		V/µs

Note 1: Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. Exposure to any Absolute Maximum Rating condition for extended periods may affect device reliability and lifetime.

Note 2: ESD (Electrostatic Discharge) sensitive device. Extensive use of ESD protection devices are used internal to the LT1990, however, high electrostatic discharge can damage or degrade the device. Use proper ESD handling precautions.

Note 3: A heat sink may be required to keep the junction temperature below absolute maximum.

Note 4: The LT1990C/LT1990I are guaranteed functional over the operating temperature range of -40° C to 85°C. The LT1990H is guaranteed functional over the operating temperature range of -40° C to 125°C.

Note 5: The LT1990C is guaranteed to meet the specified performance from 0° C to 70° C and is designed, characterized and expected to meet specified performance from -40° C to 85° C but is not tested or QA

sampled at these temperatures. The LT1990I is guaranteed to meet specified performance from -40° C to 85° C. The LT1990H is guaranteed to meet specified performance from -40° C to 125° C.

Note 6: G = 10 limits are guaranteed by correlation to G = 1 tests and gain error tests at G = 10.

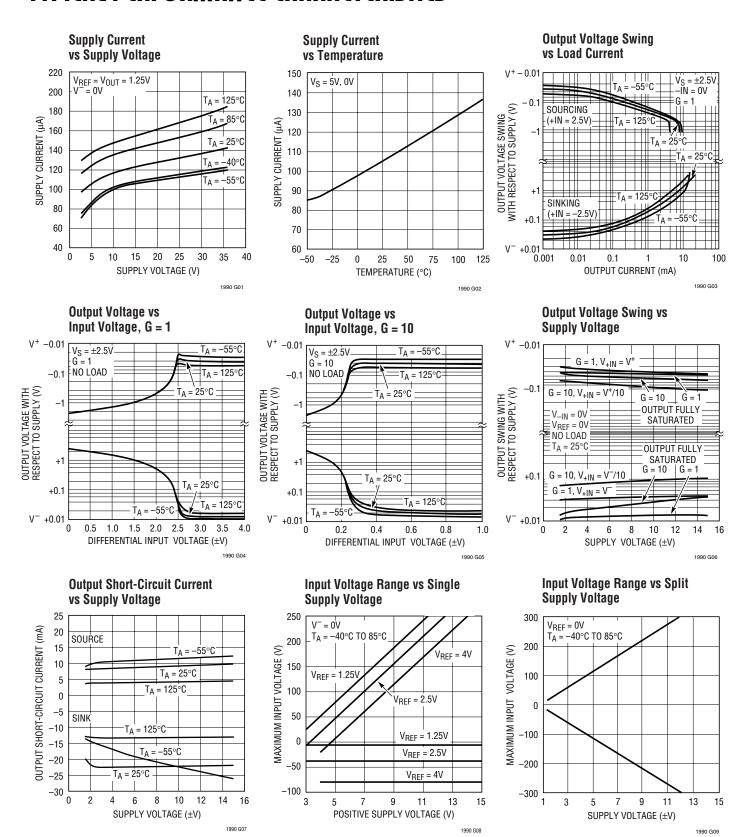
Note 7: Limits are guaranteed by correlation to -5V to 80V CMRR tests.

Note 8: $V_S = 3V$ limits are guaranteed by correlation to $V_S = 5V$ and $V_S = \pm 15V$ tests.

Note 9: $V_S = 5V$ limits are guaranteed by correlation to $V_S = 3V$ and $V_S = \pm 15V$ tests.

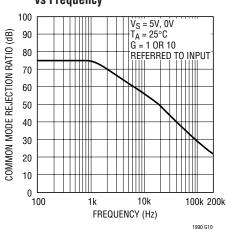
Note 10: This parameter is not 100% tested.

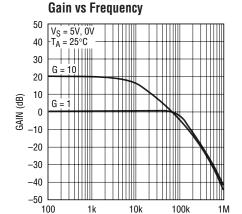
Note 11: Hysteresis in offset voltage is created by package stress that differs depending on whether the IC was previously at a higher or lower temperature. Offset voltage hysteresis is always measured at 25°C, but the IC is cycled to 85°C I-grade (70°C C-grade or 125°C H-grade) or -40°C I/H-grade (0°C C-grade) before successive measurement.





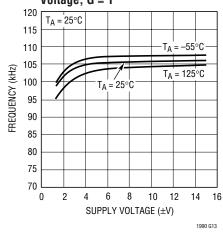




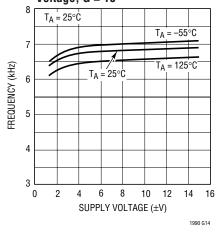


FREQUENCY (Hz)

-3dB Bandwidth vs Supply Voltage, G = 1

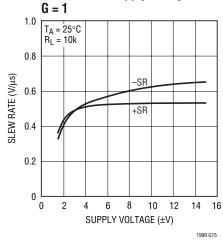


-3dB Bandwidth vs Supply Voltage, G = 10

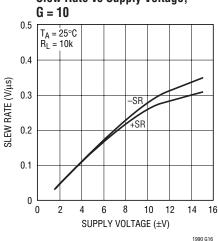


Slew Rate vs Supply Voltage,

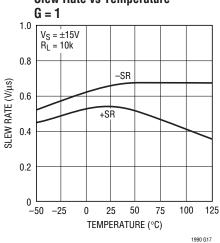
1990 G12



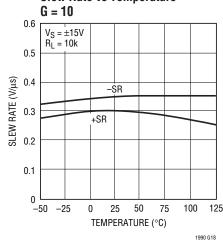
Slew Rate vs Supply Voltage,

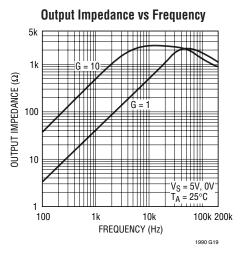


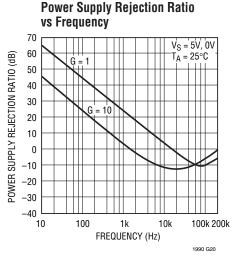
Slew Rate vs Temperature

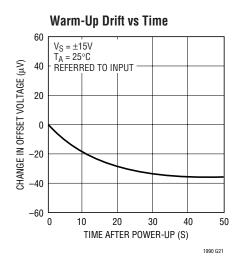


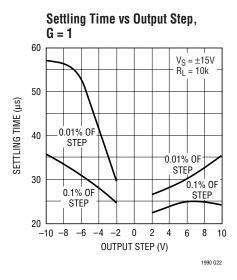
Slew Rate vs Temperature

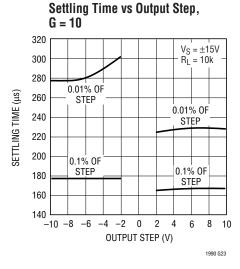


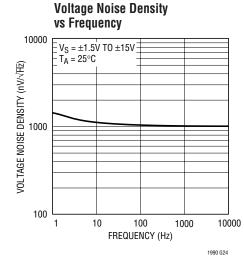


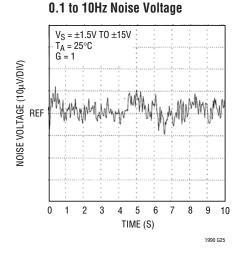


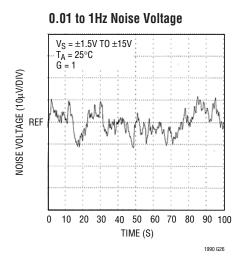


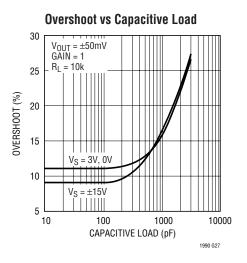




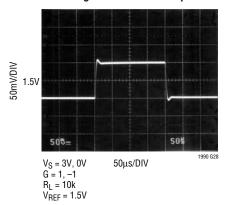




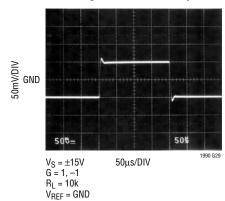




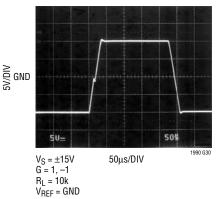
Small Signal Transient Response



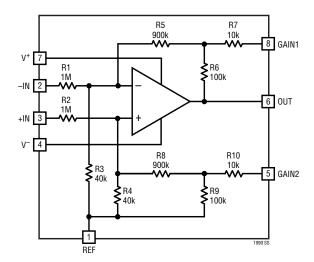
Small Signal Transient Response



Large Signal Transient Response



BLOCK DIAGRAM



PIN FUNCTIONS

REF (Pin 1): Reference Input. Sets the output level when the difference between the inputs is zero.

-IN (Pin 2): Inverting Input. Connects a $1M\Omega$ resistor to the op amp's inverting input. Designed to permit high voltage operation.

+IN (Pin 3): Noninverting Input. Connects a $1M\Omega$ resistor to the op amp's noninverting input. Designed to permit high voltage operation.

V⁻ (**Pin 4**): Negative Power Supply. Can be either ground (in single supply applications) or a negative voltage (in split supply applications).

GAIN2 (Pin 5): Gain = 10 Select Input. Configures the

amplifier for a gain of 10 when connected to the GAIN1 pin and the REF pin. The gain is equal to one when both GAIN2 and GAIN1 are open. See Applications section for additional functions.

OUT (Pin 6): Output. $V_{OUT} = G \cdot (V_{+IN} - V_{-IN}) + V_{REF}$, in the basic configuration.

 V^+ (Pin 7): Positive Power Supply. Can range from 2.7V to 36V above the V^- voltage.

GAIN1 (Pin 8): Gain = 10 Select Input. Configures the amplifier for a gain of 10 when connected to the GAIN2 pin and the REF pin. The gain is equal to one when both GAIN1 and GAIN2 are open. See Applications section for additional functions.



APPLICATIONS INFORMATION

Primary Features

The LT1990 is a complete gain-block solution for high input common mode voltage applications, incorporating a low power precision operational amplifier providing railto-rail output swing along with on-chip precision thin-film resistors for high accuracy. The Block Diagram shows the internal architecture of the part. The on-chip resistors form a modified difference amplifier including a reference port for introducing offset or other additive waveforms. With pin-strapping alone either unity gain or gain of 10 is produced with high precision. The resistor network is designed to produce internal common-mode voltage division of 27 so that a very large input range is available compared to the power supply voltage(s) used by the LT1990 itself. The LT1990 is ideally suited to situations where relatively small signals need to be extracted from high voltage circuits, as is the case in many current monitoring instrumentation applications for example. With the ability to accept a range of input voltages well outside the limits of the local power rails and its greater than $1 \mathrm{M}\Omega$ input impedances, development of precision low power over-the-top and under-the-bottom instrumentation designs is greatly simplified with the LT1990 single chip solution over conventional discrete implementations.

Classic Difference Amplifier

Used in the basic difference amplifier topology where the gain G is pin-strap configurable to be unity or ten, the following relationship is realized:

$$V_0 = G \bullet (V_{+IN} - V_{-IN}) + V_{RFF}$$

To operate in unity gain, the GAIN1 and GAIN2 pins are left disconnected. For G = 10 operation, the GAIN1 and GAIN2 pins are simply connected to the REF pin.

The input common mode range capability is up to ± 250 V, governed by the following relationships:

For G = 1 and G = 10 where GAIN1 and GAIN2 are only tied together (not grounded,etc):

$$V_{CM+} \le 27 \bullet V^+ - 26 \bullet V_{REF} - 23$$

$$V_{CM-} \ge 27 \bullet V^- - 26 \bullet V_{RFF} + 27$$

For G = 10 where GAIN1 and GAIN2 are tied to a common potential V_{GAIN} :

$$V_{CM+} \le 27 \cdot V^{+} - 26 \cdot V_{REF} - 23 - V_{GAIN}$$

$$V_{CM-} \ge 27 \bullet V^- - 26 \bullet V_{RFF} + 27 - V_{GAIN}$$

For split supplies over about $\pm 11V$, the full $\pm 250V$ common mode range is normally available (with V_{REF} a small fraction of the supply). With lower supply voltages, an appropriate selection of V_{REF} can tailor the input common mode range to a specific requirement. As an example, the following low supply voltage scenarios are readily implemented with the LT1990:

Supply	V _{REF}	V _{CM} Range
+3V	1.25V	-5V to 25V (e.g. 12V automotive environment)
+5V	1.25V	-5V to 80V (e.g. 42V automotive environment)
+5V	4.00V	-77V to 8V (e.g. telecom environment; use downward signaling)

Configuring Other Gains

An intermediate gain G ranging between 1 and 10 may be produced by placing an adjustable resistance between the GAIN1 and GAIN2 pins according to the following nominal relationship:

$$R_{GAIN} \approx (180k/(G-1)) - 20k$$

While the expression is exact, the value is approximate because the absolute resistance of the internal network could vary on a unit-to-unit basis by as much as $\pm 30\%$ from the nominal figures and the external gain resistance is required to accommodate that deviation. Once adjusted, however, the gain stability is excellent by virtue of the $-30\text{ppm}/^{\circ}\text{C}$ typical temperature coefficient offered by the on-chip thin-film resistor process.

Preserving and Enhancing Common Mode Rejection

The basic difference amplifier topology of the LT1990 requires that source impedances seen by the input pins +IN and -IN, should be matched to within a few tens of ohms to avoid increasing common mode induced errors beyond the basic production limits of the part. Known source imbalances beyond that level should be compensated for by the addition of series resistance to the lower-impedance source. Also the source impedance of a signal connected to the REF pin must be on the order of a few ohms or less to preserve the high accuracy of the LT1990.



APPLICATIONS INFORMATION

While the LT1990 comes from the factory with an excellent CMRR, some precision applications with a large applied common mode voltage may require a method to trim out residual common mode error. This is easily accomplished by adding series resistance to each input, +IN and -IN, such that an adjustable resistance difference of $\pm 1 k\Omega$ is provided. This is most easily realized by adding a fixed $1k\Omega$ in series with one of the inputs, and a $2k\Omega$ trimmer in series with the other as shown in Figure 1. The trim range of this configuration is $\pm 0.1\%$ for the internal gain resistor matching, so a much more finely resolved correction is available using the LT1990 than is realizable with ordinary discrete solutions. In applications where the input common mode voltage is relatively constant and large (perhaps at or beyond the supply range), this same configuration can be treated as an offset adjustment.

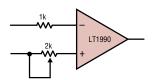


Figure 1. Optional CMRR Trim

Dual Differential-Input Arithmetic Block

The internal resistor network topology of the LT1990 allows the GAIN1 and GAIN2 pins to be used as another differential input in addition to the normal +IN and -IN port. This can be a very useful function for implementing servo-loop differential error amplifiers, for example. In this mode of operation, the output is governed by the

following relationship:

$$V_0 = 10 \cdot (V_{+IN} - V_{-IN} + V_{GAIN2} - V_{GAIN1}) + V_{REF}$$

Unlike the main inputs, the GAIN1 and GAIN2 pins are clamped by substrate diodes and ESD structures, thus the operating voltage range of these pins is limited to $V^- - 0.2V$ to $V^- + 36V$. If the GAIN inputs are brought beyond the operating input range, care must be taken to limit the input currents to less than 10mA to prevent damage to the device. For best results in this mode of operation the common mode voltage of the GAIN1 and GAIN2 pins should be equal to the REF pin voltage. Also, since the gain setting resistors associated with the GAIN1 and GAIN2 inputs are in the $10k\Omega$ area, low source impedances are particularly important to preserve the precision of the LT1990.

This dual differential input mode of operation is used in the circuit as shown in Figure 2.

This circuit is a high efficiency H-bridge driver that is PWM modulated to provide a controlled current to an electromagnet coil. Since the common mode voltage of the current sense resistor R_S varies with operating current and the coil properties, a differential feedback is required. In this application, it was desirable to allow the control input to utilize the wide common mode range port (+IN and -IN) so that constraints on input referencing are eliminated. The GAIN1 and GAIN2 pins always operate within the supply range and both ports operate with a gain of 10 to develop the loop error. The LTC1923 provides the loop integrator and PWM functions of the servo.

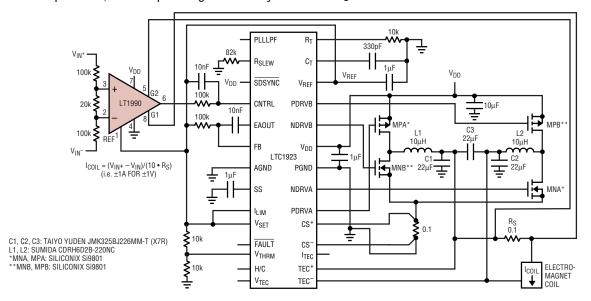


Figure 2. PWM-Based ±1A Electromagnet Current Controller



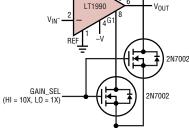
TYPICAL APPLICATION

Telecom Supply Current Monitor

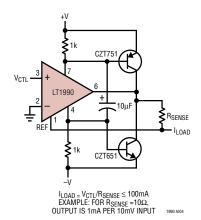
LOAD LT1990 $\begin{array}{l} -77 V \leq V_{CM} \leq 8 V \\ V_{OUT} = V_{REF} - (10 \bullet I_L \bullet R_S) \end{array} \label{eq:volume}$ OUT LT6650 GND FB **≨**20k

Selectable Gain Amplifier

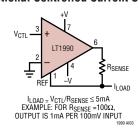
V_{OUT}



Boosted Bidirectional Controlled Current Source

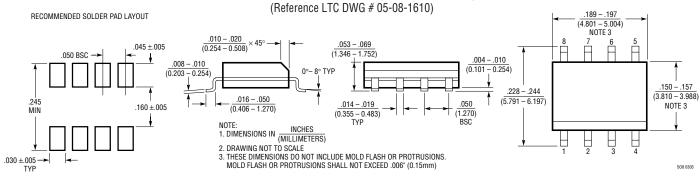


Bidirectional Controlled Current Source



PACKAGE DESCRIPTION

S8 Package 8-Lead Plastic Small Outline (Narrow .150 Inch)



RELATED PARTS

PART NUMBER	DESCRIPTION	COMMENTS
LT1787	Precision High Side Current Sense Amplifier	On-Chip Precision Resistor Array
LT1789	Micropower Instrumentation Amplifier	Micropower, Precision, G = 1 to 1000
LTC1921	Dual –48V Supply and Fuse Monitor	Withstands ±200V Transients
LT1991	High Accuracy Difference Amplifier	Micropower, Precision, Pin Selectable $G = -13$ to 14
LT1995	30MHz, 1000V/µs Gain Selectable Amplifier	Pin Selectable G = -7 to 8
LT6910	Single Supply Programmable Gain Amplifier	Digitally Controlled, SOT-23, G = 0 to 100

