Micropower, Rail to Rail Input Current Sense Amplifier with Voltage Output

ISL28006

The ISL28006 is a micropower, uni-directional high-side and low-side current sense amplifier featuring a proprietary rail-to-rail input current sensing amplifier. The ISL28006 is ideal for high-side current sense applications where the sense voltage is usually much higher than the amplifier supply voltage. The device can be used to sense voltages as high as 28V when operating from a supply voltage as low as 2.7V. The micropower ISL28006 consumes only 50µA of supply current when operating from a 2.7V to 28V supply.

The ISL28006 features a common-mode input voltage range from 0V to 28V. The proprietary architecture extends the input voltage sensing range down to 0V, making it an excellent choice for low-side ground sensing applications. The benefit of this architecture is that a high degree of total output accuracy is maintained over the entire 0V to 28V common mode input voltage range.

The ISL28006 is available in fixed (100V/V, 50V/V, 20V/V and Adjustable) gains in the space saving 5 Ld SOT-23 package and the 6 Ld SOT-23 package for the adjustable gain part. The parts operate over the extended temperature range from -40°C to +125°C.

Features

- Low Power Consumption 50µA,Typ
- Supply Range. . . . . 2.7V to 28V
- Wide Common Mode Input . . . . 0V to 28V
- Gain Versions
  - ISL28006-100. . . . . 100V/V
  - ISL28006-50. 50V/V
  - ISL28006-20. 20V/V
  - ISL28006-ADJ. . . . . . . ADJ (Min Gain = 20V/V)
- Operating Temperature Range . . .-40°C to +125°C
- Packages. . . . . 5 Ld SOT-23, 6 Ld SOT-23

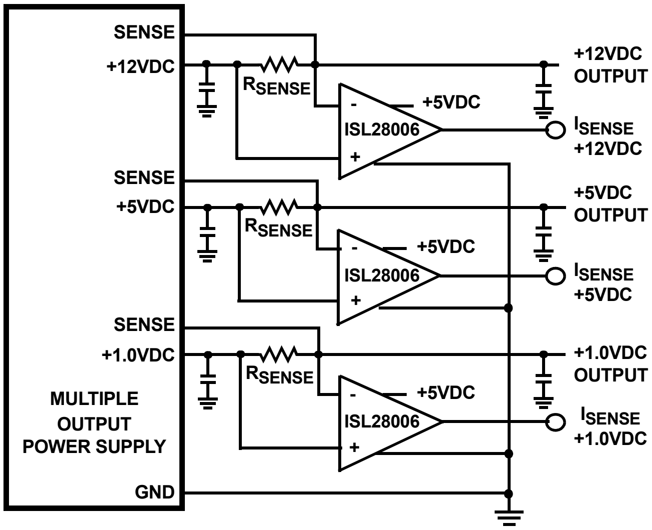
Applications* (see page 23)

- Power Management/Monitors
- Power Distribution and Safety
- DC/DC, AC/DC Converters
- Battery Management/Charging
- Automotive Power Distribution

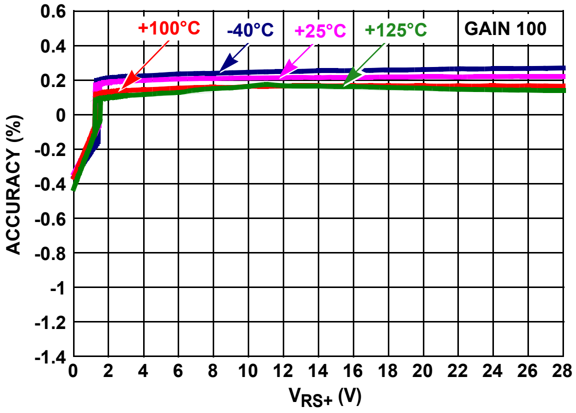
Related Literature* (see page 23)

- See AN1532 for "ISL28006 Evaluation Board User's Guide"

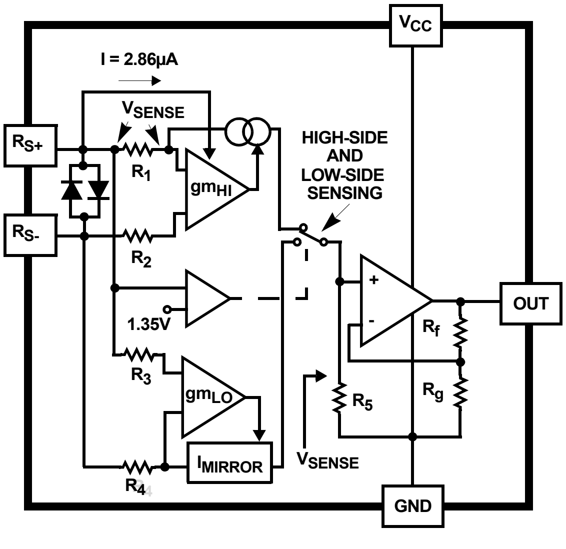
Typical Application



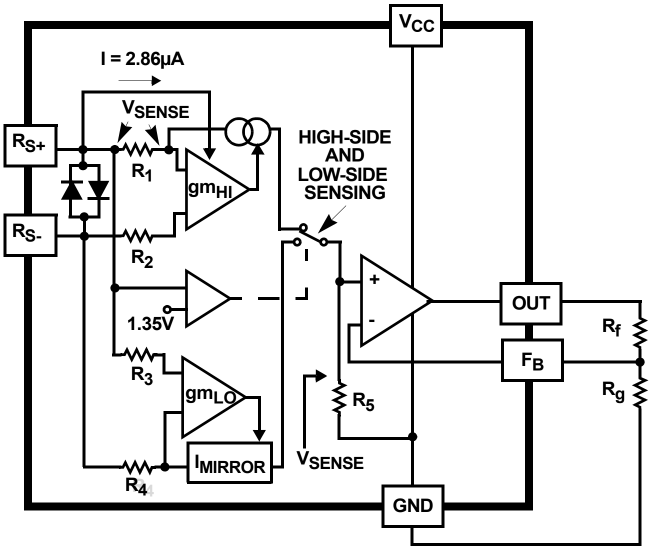
Gain Accuracy vs V_{RS+} = 0V to 28V



Block Diagram



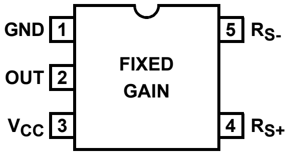
FIXED GAIN PARTS



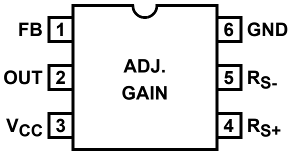
ADJUSTABLE GAIN PART

Pin Configurations

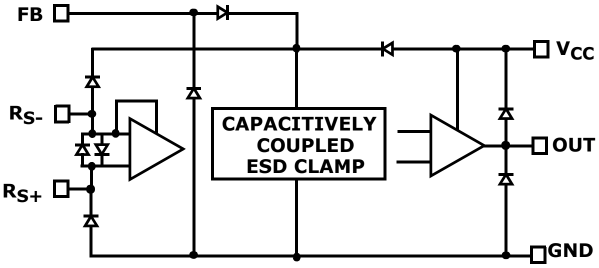
ISL28006-100, 50, 20
(5 LD SOT-23)
TOP VIEW



ISL28006-ADJ
(6 LD SOT-23)
TOP VIEW



Pin Descriptions

ISL28006-100, 50, 20 (5 LD SOT-23)	ISL28006-ADJ (6 LD SOT-23)	PIN NAME	DESCRIPTION
1	6	GND	Power Ground
	1	FB	Input Pin for External Resistors
2	2	OUT	Amplifier Output
3	3	V_{CC}	Positive Power Supply
4	4	R_{S+}	Sense Voltage Non-inverting Input
5	5	R_{S-}	Sense Voltage Inverting Input
			

Ordering Information

PART NUMBER (Notes 1, 2, 3)	GAIN	PART MARKING (Note 4)	PACKAGE Tape & Reel (Pb-Free)	PKG. DWG. #
ISL28006FH100Z-T7	100V/V	BDJA	5 Ld SOT-23	P5.064A
ISL28006FH100Z-T7A	100V/V	BDJA	5 Ld SOT-23	P5.064A
ISL28006FH50Z-T7	50V/V	BDHA	5 Ld SOT-23	P5.064A
ISL28006FH50Z-T7A	50V/V	BDHA	5 Ld SOT-23	P5.064A
ISL28006FH20Z-T7	20V/V	BDGA	5 Ld SOT-23	P5.064A
ISL28006FH20Z-T7A	20V/V	BDGA	5 Ld SOT-23	P5.064A
ISL28006FHADJZ-T7	ADJ	BDFA	6 Ld SOT-23	P6.064
ISL28006FHADJZ-T7A	ADJ	BDFA	6 Ld SOT-23	P6.064
ISL28006FH-100EVAL1Z	100V/V Evaluation Board			
ISL28006FH-50EVAL1Z	50V/V Evaluation Board			
ISL28006FH-20EVAL1Z	20V/V Evaluation Board			
ISL28006FH-ADJEVAL1Z	Adjustable Evaluation Board			

NOTES:

1. Please refer to TB347 for details on reel specifications.
2. These Intersil Pb-free plastic packaged products employ special Pb-free material sets, molding compounds/die attach materials, and 100% matte tin plate plus anneal (e3 termination finish, which is RoHS compliant and compatible with both SnPb and Pb-free soldering operations). Intersil Pb-free products are MSL classified at Pb-free peak reflow temperatures that meet or exceed the Pb-free requirements of IPC/JEDEC J STD-020.
3. For Moisture Sensitivity Level (MSL), please see device information page for ISL28006. For more information on MSL please see techbrief TB363.
4. The part marking is located on the bottom of the part.

Absolute Maximum Ratings

Max Supply Voltage 28V
Max Differential Input Current 20mA
Max Differential Input Voltage . . . . ±0.5V
Max Input Voltage (R_{S+}, R_{S-}, FB). . . . . . . GND - 0.5V to 30V
Max Input Current for Input Voltage <GND - 0.5V . . . . ±20mA
Output Short-Circuit Duration . . . . Indefinite
Di-Electrically Isolated PR40 Process. . . . . Latch-up free
ESD Rating
  Human Body Model (Tested per JESD22-A114F) 4kV
  Machine Model (Tested per EIA/JESD22-A115-A). . . . 200V
  Charged Device Model (Tested per JESD22-C101D) . . . 1.5kV

Thermal Information

Thermal Resistance (Typical)	θ_{JA} (°C/W)	θ_{JC} (°C/W)
5 Ld SOT-23 (Notes 5, 6)	190	90
6 Ld SOT-23 (Notes 5, 6)	180	90

Maximum Storage Temperature Range . . . -65°C to +150°C
Maximum Junction Temperature (T_{JMAX}) . . . . +150°C
Pb-Free Reflow Profile . . . . .see link below
http://www.intersil.com/pbfree/Pb-FreeReflow.asp

Recommended Operating Conditions

Ambient Temperature Range (T_A) . . . . -40°C to +125°C

CAUTION: Do not operate at or near the maximum ratings listed for extended periods of time. Exposure to such conditions may adversely impact product reliability and result in failures not covered by warranty.

NOTES:

5. θ_{JA} is measured with the component mounted on a high effective thermal conductivity test board in free air. See Tech Brief TB379 for details.
6. For θ_{JC}, the "case temp" location is taken at the package top center.

Electrical Specifications

V_{CC} = 12V, V_{RS+} = 0V to 28V, V_{SENSE} = 0V, R_{LOAD} = 1MΩ, T_A = +25°C unless otherwise specified. **Boldface limits apply over the operating temperature range, -40°C to +125°C. Temperature data established by characterization**.

PARAMETER	DESCRIPTION	CONDITIONS	MIN (Note 7)	TYP	MAX (Note 7)	UNIT
V_{OS} (Input Offset Voltage)	Gain = 100 (Notes 8, 9)	V_{CC} = V_{RS}+ = 12V, V_{SENSE} = 20mV to 100mV	-250	60	250	µV
			-300		**300**	µV
		V_{CC} = 12V, V_{RS}+ = 0.2V, V_{SENSE} = 20mV to 100mV	-2.5	-1.2	2.5	mV
			-2.8		**2.8**	mV
	Gain = 50, Gain = 20 (Notes 8, 9)	V_{CC} = V_{RS}+ = 12V, V_{SENSE} = 20mV to 100mV	-300	60	300	µV
			-450		**450**	µV
		V_{CC} = 12V, V_{RS}+ = 0.2V, V_{SENSE} = 20mV to 100mV	-2.8	-1.2	2.8	mV
			-3.2		**3.2**	mV
	Adjustable, Gain = 21 R_f = 100kΩ, R_g = 5kΩ (Notes 8, 9)	V_{CC} = V_{RS}+ = 12V, V_{SENSE} = 20mV to 100mV	-300	60	300	µV
			-450		**450**	µV
		V_{CC} = 12V, V_{RS}+ = 0.2V, V_{SENSE} = 20mV to 100mV	-3.1	-1.2	3.1	mV
			-3.4		**3.4**	mV
I_{RS}+, I_{RS} -	Leakage Current	V_{CC} = 0V, V_{RS+} = 28V		0.041	1.2	µA
					1.5	µA

Electrical Specifications

V_{CC} = 12V, V_{RS+} = 0V to 28V, $\mathbf{V_{SENSE}}$ = 0V, R_{LOAD} = 1MΩ, T_A = +25°C unless otherwise specified. **Boldface limits apply over the operating temperature range, -40°C to +125°C. Temperature data established by characterization**. **(Continued)**

PARAMETER	DESCRIPTION	CONDITIONS	MIN (Note 7)	TYP	MAX (Note 7)	UNIT
I_{RS}+ (+ Input Bias Current)	Gain = 100	V_{RS}+ = 2V, V_{SENSE} = 5mV		4.7	6	µA
					7	µA
		V_{RS}+ = 0V, V_{SENSE} = 5mV	-500	-432		nA
			-600			nA
	Gain = 50, Gain = 20	V_{RS}+ = 2V, V_{SENSE} = 5mV		4.7	6	µA
					8	µA
		V_{RS}+ = 0V, V_{SENSE} = 5mV	-700	-432		nA
			-840			nA
	ADJ Gain = 101 R_f = 100kΩ, R_g = 1kΩ	V_{RS}+ = 2V, V_{SENSE} = 5mV		4.7	6	µA
					7	µA
		V_{RS}+ = 0V, V_{SENSE} = 5mV	-500	-432		nA
			-600			nA
I_{RS} - (- Input Bias Current)	G = 100, 50, 20, ADJ	V_{RS}+ = 2V, V_{SENSE} = 5mV		5	50	nA
					75	nA
		V_{RS}+ = 0V, V_{SENSE} = 5mV	-125	-45		nA
			-130			nA
CMRR	Common Mode Rejection Ratio	V_{RS}+ = 2V to 28V	**105**	**115**		dB
PSRR	Power Supply Rejection Ratio	V_{CC} = 2.7V to 28V, V_{RS}+ = 2V	**90**	**105**		dB
V_{FS}	Full-scale Sense Voltage	V_{CC} = 28V, V_{RS+} = 0.2V, 12V	200			mV
G (Gain)	(Note 8)	ISL28006-100		100		V/V
		ISL28006-50		50		V/V
		ISL28006-20		20		V/V
		ISL28006-ADJ	20			V/V
G_A (Gain Accuracy)	Gain = 100 (Note 10)	V_{CC} = V_{RS}+ = 12V, V_{SENSE} = 20mV to 100mV	-0.2		0.7	%
			-1		**1**	%
		V_{CC} = 12V, V_{RS}+ = 0.1V, V_{SENSE} = 20mV to 100mV		-0.25		%
	Gain = 50, Gain = 20 (Note 10)	V_{CC} = V_{RS}+ = 12V, V_{SENSE} = 20mV to 100mV	-0.35		0.7	%
			-1		**1**	%
		V_{CC} = 12V, V_{RS}+ = 0.1V, V_{SENSE} = 20mV to 100mV	-2.2	-0.33	2.2	%
			-2.3		**2.3**	%
	ADJ Gain = 21 R_f = 100kΩ, R_g = 5kΩ (Note 10)	V_{CC} = V_{RS}+ = 12V, V_{SENSE} = 20mV to 100mV	-0.65		1	%
			-1		**1.05**	%
		V_{CC} = 12V, V_{RS}+ = 0.1V, V_{SENSE} = 20mV to 100mV	-2.2	-0.33	2.2	%
			-2.3		**2.3**	

Electrical Specifications

V_{CC} = 12V, V_{RS+} = 0V to 28V, $\mathbf{V_{SENSE}}$ = 0V, R_{LOAD} = 1MΩ, T_A = +25°C unless otherwise specified. **Boldface limits apply over the operating temperature range, -40°C to +125°C. Temperature data established by characterization**. **(Continued)**

PARAMETER	DESCRIPTION	CONDITIONS	MIN (Note 7)	TYP	MAX (Note 7)	UNIT
V_{OA} (Total Output Accuracy)	Gain = 100 (Note 11)	V_{CC} = V_{RS}+ = 12V, V_{SENSE} = 100mV	-0.7		0.7	%
			-0.9		**0.9**	%
		V_{CC} = 12V, V_{RS}+ = 0.1V, V_{SENSE} = 100mV		-1.25		%
	Gain = 50, Gain = 20 (Note 11)	V_{CC} = V_{RS}+ = 12V, V_{SENSE} = 100mV	-0.7		0.7	%
			-0.9		**0.9**	%
		V_{CC} = 12V, V_{RS}+ = 0.1V, V_{SENSE} = 100mV	-4.7	-1.41	1.8	%
			-5.2		**2.3**	%
	ADJ Gain = 21 R_f = 100kΩ, R_g = 5kΩ (Note 11)	V_{CC} = V_{RS}+ = 12V, V_{SENSE} = 100mV	-0.7		1.05	%
			-0.9		**1.2**	%
		V_{CC} = 12V, V_{RS}+ = 0.1V, V_{SENSE} = 100mV	-4.7	-1.41	1.8	%
			-5.2		**2.3**	%
V_{OH}	Output Voltage Swing, High V_{CC} - V_{OUT}	I_O = -500µA, V_{CC} = 2.7V, V_{SENSE} = 100mV, V_{RS}+ = 2V		39	50	mV
V_{OL}	Output Voltage Swing, Low V_{OUT}	I_O = 500µA, V_{CC} = 2.7V V_{SENSE} = 0V, V_{RS}+ = 2V		30	50	mV
R_{OUT}	Output Resistance	V_{CC} = V_{RS}+ = 12V, V_{SENSE} = 100mV I_{OUT} = 10µA to 1mA		6.5		Ω
I_{SC+}	Short Circuit Sourcing Current	V_{CC} = V_{RS}+ = 5V, R_L = 10Ω		4.8		mA
I_{SC-}	Short Circuit Sinking Current	V_{CC} = V_{RS}+ = 5V, R_L = 10Ω		8.7		mA
I_S	Gain = 100	V_{RS}+ > 2V, V_{SENSE} = 5mV		50	59	µA
					62	µA
	Gain = 50, 20,	V_{RS}+ > 2V, V_{SENSE} = 5mV		50	62	µA
					63	µA
	ADJ Gain = 21 R_f = 100kΩ, R_g = 5kΩ	V_{RS}+ > 2V, V_{SENSE} = 5mV		50	62	µA
					63	µA
V_{CC}	Supply Voltage	Guaranteed by PSRR	**2.7**		**28**	V
Slew Rate	Gain = 100	Pulse on R_{S+} pin, V_{OUT} = 8V_{P-P} (Figure 65)	0.58	0.76		V/µs
	Gain = 50	Pulse on R_{S+} pin, V_{OUT} = 8V_{P-P} (Figure 65)	0.58	0.67		V/µs
	Gain = 20	Pulse on R_{S+} pin, V_{OUT} = 3.5V_{P-P} (Figure 65)	0.50	0.67		V/µs
	ADJ Gain = 21 R_f = 100kΩ, R_g = 5kΩ	Pulse on R_{S+} pin, V_{OUT} = 3.5V_{P-P} (Figure 65)	0.50	0.67		V/µs

Electrical Specifications

V_{CC} = 12V, V_{RS+} = 0V to 28V, $\mathbf{V_{SENSE}}$ = 0V, R_{LOAD} = 1MΩ, T_A = +25°C unless otherwise specified. **Boldface limits apply over the operating temperature range, -40°C to +125°C. Temperature data established by characterization**. **(Continued)**

PARAMETER	DESCRIPTION	CONDITIONS	MIN (Note 7)	TYP	MAX (Note 7)	UNIT
BW_{-3dB}	Gain = 100	V_{RS}+ = 12V, 0.1V, V_{SENSE} = 100mV		110		kHz
	Gain = 50	V_{RS}+ = 12V, 0.1V, V_{SENSE} = 100mV		160		kHz
	Gain = 20	V_{RS}+ = 12V, 0.1V, V_{SENSE} = 100mV		180		kHz
	ADJ, Gain = 101 (Figure 57)	V_{RS}+ = 12V, 0.1V, V_{SENSE} = 100mV, R_f = 100kΩ, R_g = 1kΩ		40		kHz
	ADJ, Gain = 51 (Figure 57)	V_{RS}+ = 12V, V_{SENSE} = 100mV, R_f = 100kΩ, R_g = 2kΩ		78		kHz
		V_{RS}+ = 0.1V, V_{SENSE} = 100mV, R_f = 100kΩ, R_g = 2kΩ		122		kHz
	ADJ, Gain = 21 (Figure 57)	V_{RS}+ = 12V, V_{SENSE} = 100mV, R_f = 100kΩ, R_g = 5kΩ		131		kHz
		V_{RS}+ = 0.1V, V_{SENSE} = 100mV, R_f = 100kΩ, R_g = 5kΩ		237		kHz
t_S	Output Settling Time to 1% of Final Value	V_{CC} = V_{RS}+ = 12V, V_{OUT} = 10V step, V_{SENSE} > 7mV		15		µs
		V_{CC} = V_{RS}+ = 0.2V, V_{OUT} = 10V step, V_{SENSE} > 7mV		20		µs
	Capacitive-Load Stability	No sustained oscillations		300		pF
$t_{S\ Power\text{-}up}$	Power-Up Time to 1% of Final Value	V_{CC} = V_{RS}+ = 12V, V_{SENSE} = 100mV		15		µs
		V_{CC} = 12V, V_{RS}+ = 0.2V V_{SENSE} = 100mV		50		µs
	Saturation Recovery Time	V_{CC} = V_{RS}+ = 12V, V_{SENSE} = 100mV, overdrive		10		µs

NOTES:

7. Parameters with MIN and/or MAX limits are 100% tested at +25°C, unless otherwise specified. Temperature limits established by characterization and are not production tested.
8. DEFINITION OF TERMS:
   - $V_{SENSE}A = V_{SENSE}$ @ 100mV
   - $V_{SENSE}B = V_{SENSE}$ @ 20mV
   - $V_{OUT}A = V_{OUT}$ @ $V_{SENSE}A$ = 100mV
   - $V_{OUT}B = V_{OUT}$ @ $V_{SENSE}B$ = 20mV
   - $G = \text{GAIN} = \left(\frac{V_{OUT}A - V_{OUT}B}{V_{SENSE}A - V_{SENSE}B}\right)$
9. V_{OS} is extrapolated from the gain measurement. $V_{OS} = V_{SENSE}A - \frac{V_{OUT}A}{G}$
10. % Gain Accuracy = $G_A = \left(\frac{G_{MEASURED} - G_{EXPECTED}}{G_{EXPECTED}}\right) \times 100$
11. Output Accuracy % VOA = $\left(\frac{VOUT_{MEASURED} - VOUT_{EXPECTED}}{VOUT_{EXPECTED}}\right) \times 100$ where $V_{OUT} = V_{SENSE}$ X GAIN and V_{SENSE} = 100mV

Typical Performance Curves V_{CC} = 12V, R_L = 1M, unless otherwise specified.

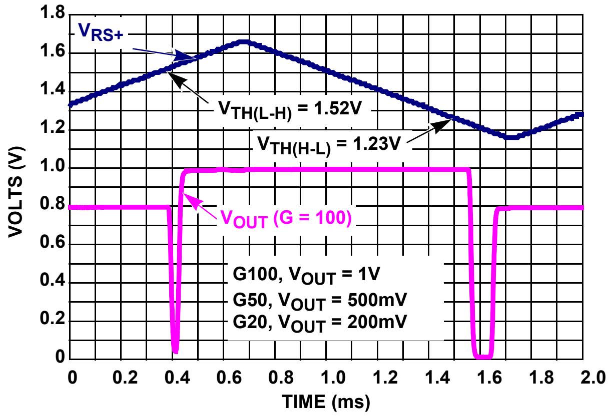


FIGURE 1. HIGH-SIDE and LOW-SIDE THRESHOLD VOLTAGE $V_{RS+(L-H)}$ and $V_{RS+(H-L)}$, V_{SENSE} = 10mV

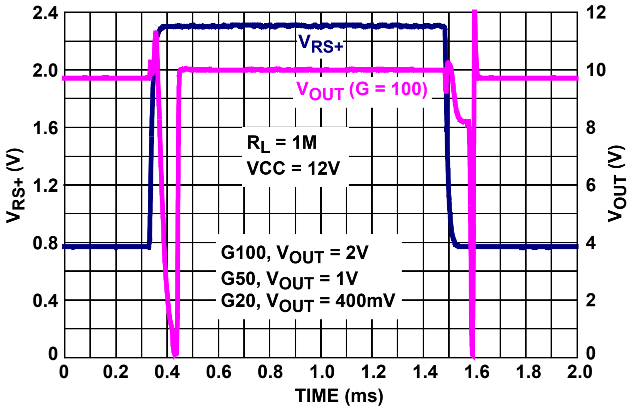


FIGURE 2. V_{OUT} vs V_{RS+}, V_{SENSE} = 20mV TRANSIENT RESPONSE

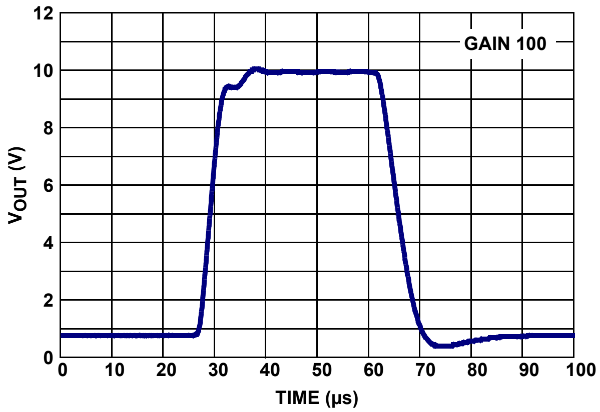


FIGURE 3. LARGE SIGNAL TRANSIENT RESPONSE V_{RS+} = 0.2V, V_{SENSE} = 100mV

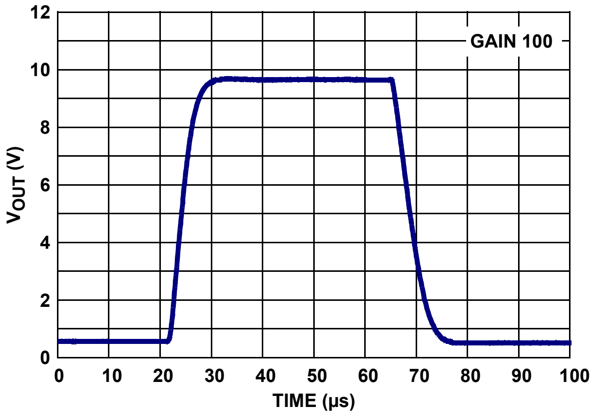


FIGURE 4. LARGE SIGNAL TRANSIENT RESPONSE V_{RS+} = 12V, V_{SENSE} = 100mV

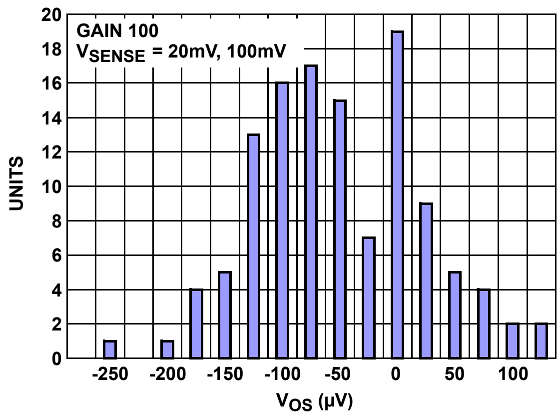


FIGURE 5. V_{OS} (µV) DISTRIBUTION AT +25°C, V_{RS+} = 12V, QUANTITY: 100

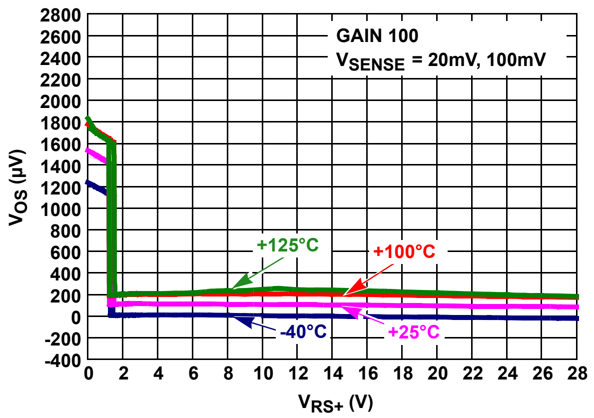


FIGURE 6. V_{OS} vs V_{RS+}

Typical Performance Curves V_{CC} = 12V, R_L = 1M, unless otherwise specified. **(Continued)**

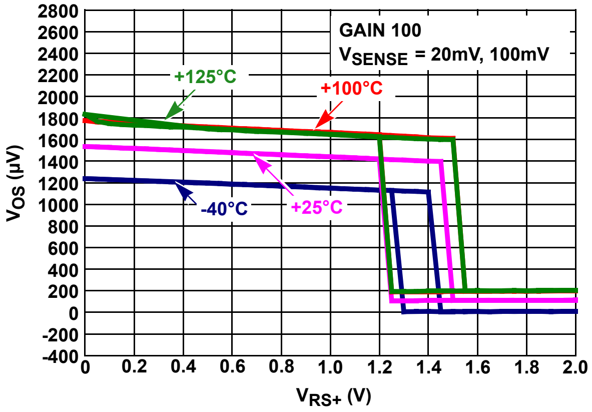


FIGURE 7. V_{OS} vs V_{RS+}

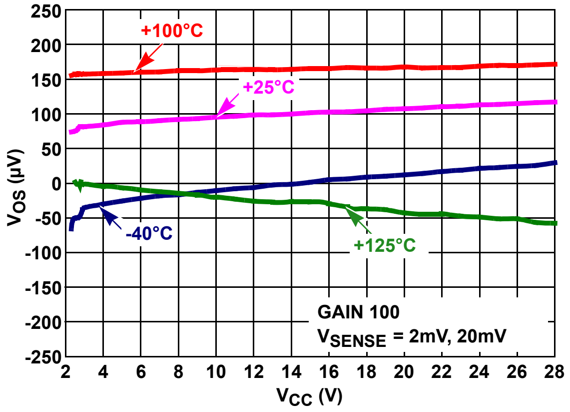


FIGURE 8. V_{OS} vs V_{CC}, V_{RS+}= 12V

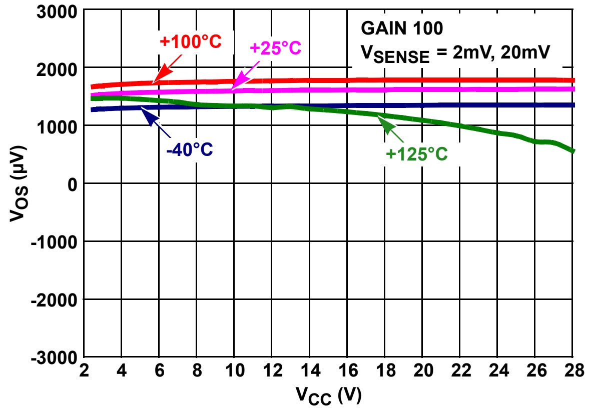


FIGURE 9. V_{OS} vs V_{CC}, V_{RS+} = 0.1V

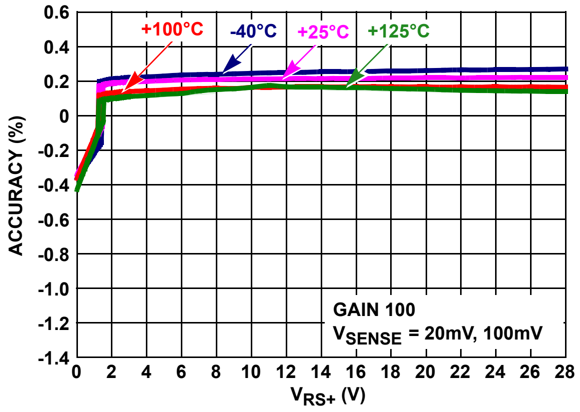


FIGURE 10. GAIN ACCURACY vs V_{RS+} = 0V TO 28V

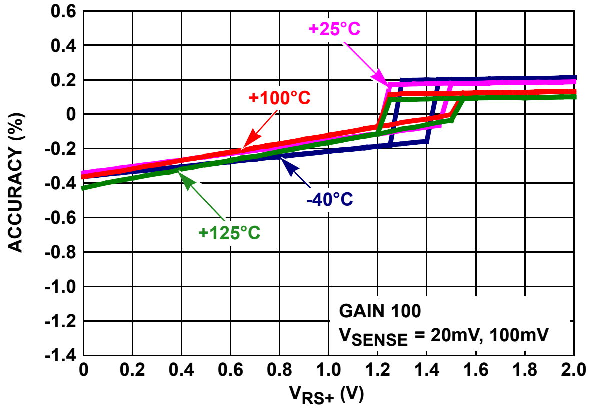


FIGURE 11. GAIN ACCURACY vs V_{RS+} = 0V TO 2V

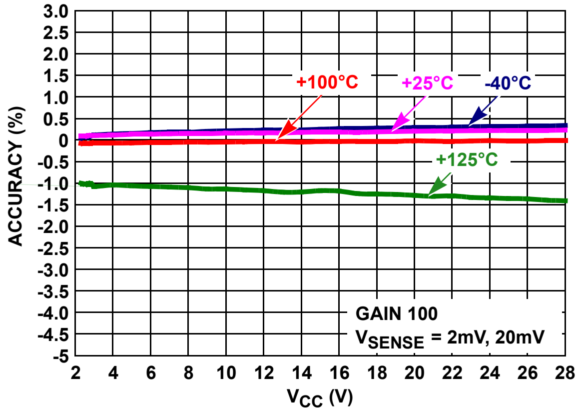


FIGURE 12. GAIN ACCURACY vs V_{CC}, V_{RS+} = 12V

Typical Performance Curves V_{CC} = 12V, R_L = 1M, unless otherwise specified. **(Continued)**

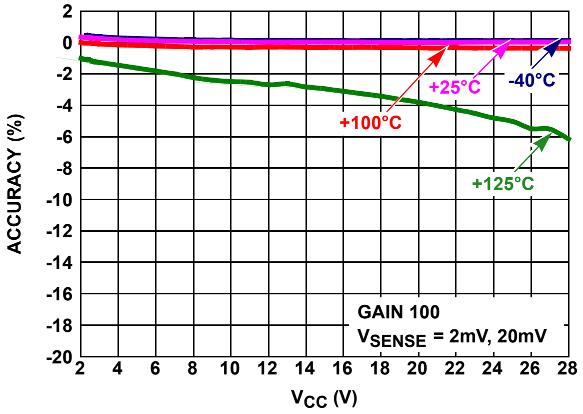


FIGURE 13. GAIN ACCURACY vs V_{CC}, V_{RS+} = 0.1V

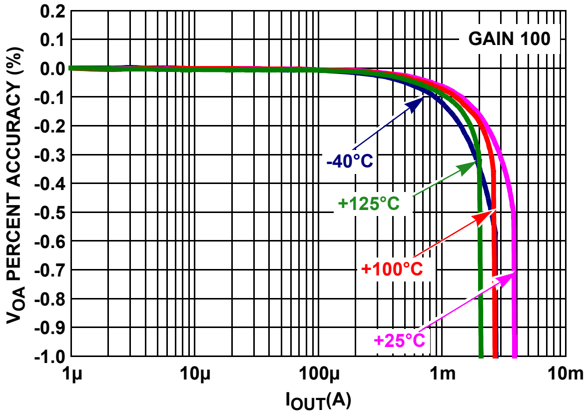


FIGURE 14. NORMALIZED V_{OA} vs I_{OUT}

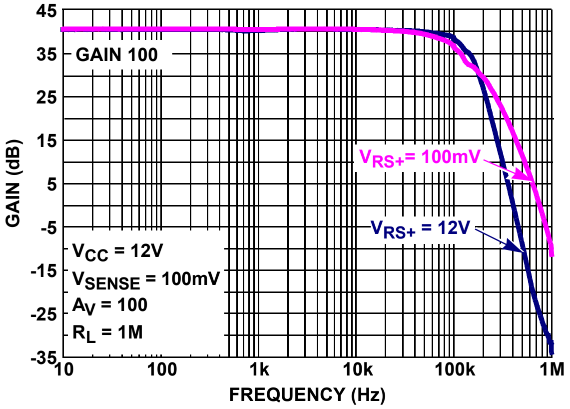


FIGURE 15. GAIN vs FREQUENCY V_{RS+} = 100mV/12V, V_{SENSE} = 100mV, V_{OUT} = 50mV_{P-P}

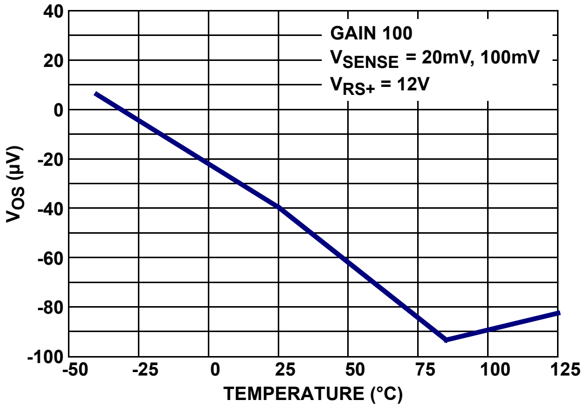


FIGURE 16. V_{OS} (µV) vs TEMPERATURE

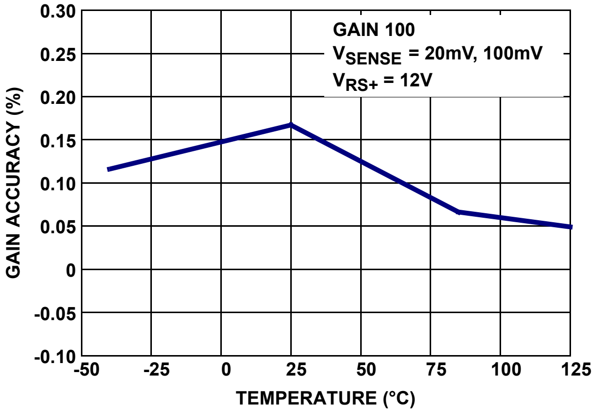


FIGURE 17. GAIN ACCURACY (%) vs TEMPERATURE

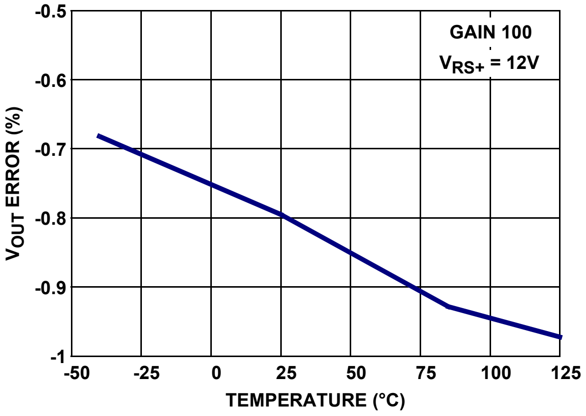


FIGURE 18. V_{OUT} ERROR (%) vs TEMPERATURE

Typical Performance Curves V_{CC} = 12V, R_L = 1M, unless otherwise specified. **(Continued)**

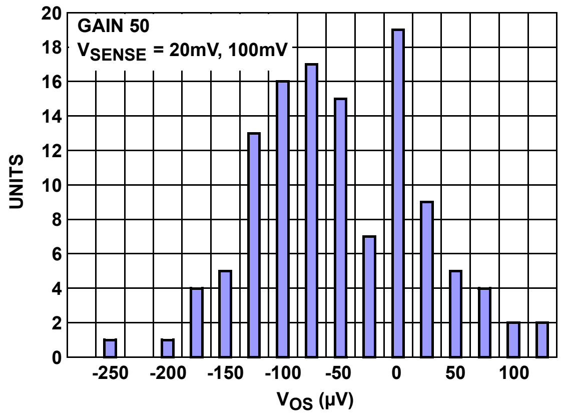


FIGURE 19. V_{OS} (µV) DISTRIBUTION AT +25°C, V_{RS+} = 12V, QUANTITY: 100

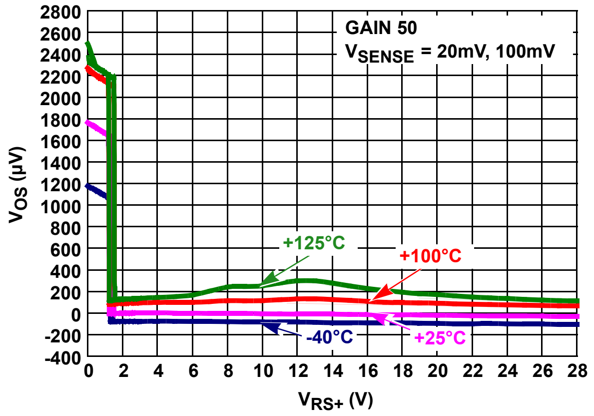


FIGURE 20. V_{OS} vs V_{RS+}

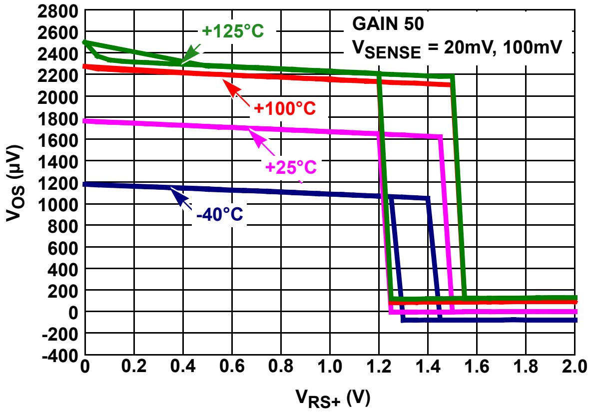


FIGURE 21. V_{OS} vs V_{RS+}

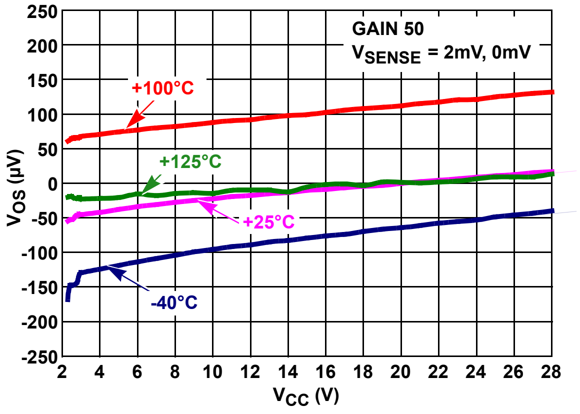


FIGURE 22. V_{OS} vs V_{CC}, V_{RS+} = 12V

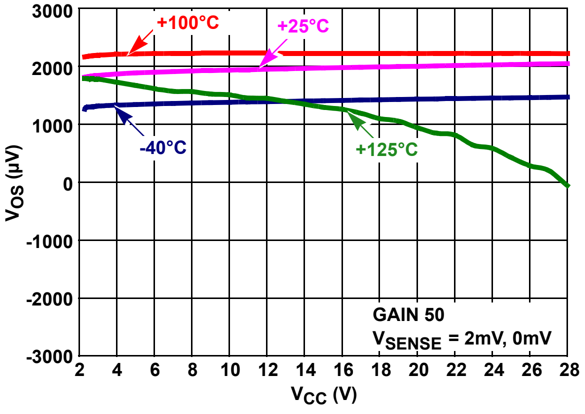


FIGURE 23. V_{OS} vs V_{CC}, V_{RS+} = V_{RS+} = 0.1V

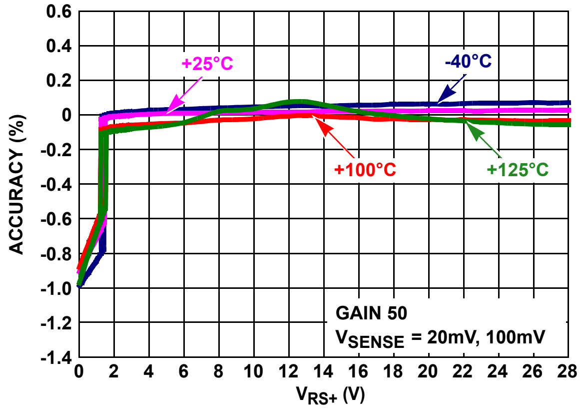


FIGURE 24. GAIN ACCURACY vs V_{RS+} = 0V TO 28V

Typical Performance Curves V_{CC} = 12V, R_L = 1M, unless otherwise specified. **(Continued)**

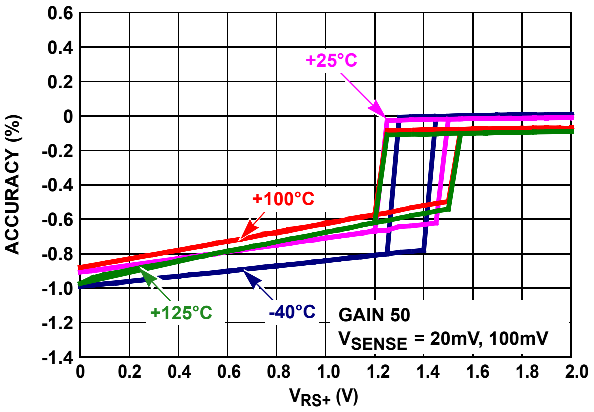


FIGURE 25. GAIN ACCURACY vs V_{RS+} = 0V TO 2V

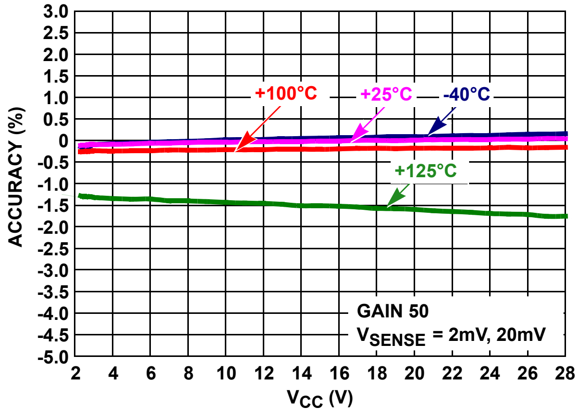


FIGURE 26. GAIN ACCURACY vs V_{CC}, HIGH-SIDE

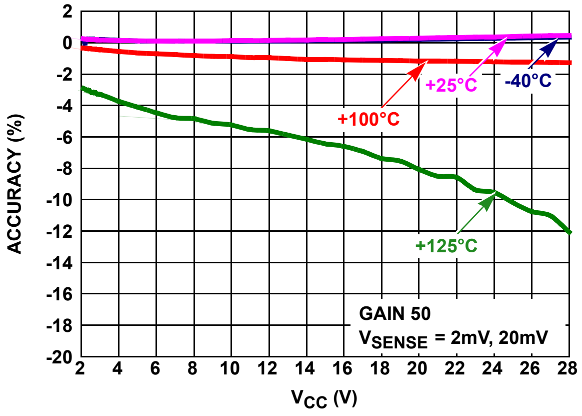


FIGURE 27. GAIN ACCURACY vs V_{CC}, LOW-SIDE

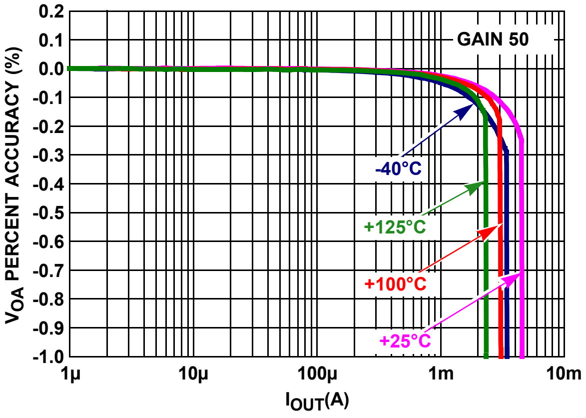


FIGURE 28. NORMALIZED V_{OA} vs I_{OUT}

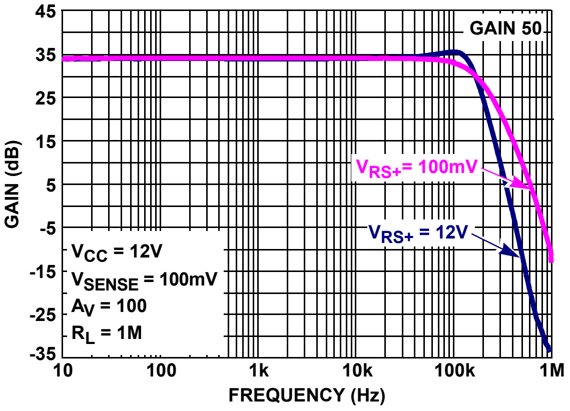


FIGURE 29. GAIN vs FREQUENCY V_{RS+} = 100mV/12V, V_{SENSE} = 100mV, V_{OUT} = 50mV$_{P-P}$

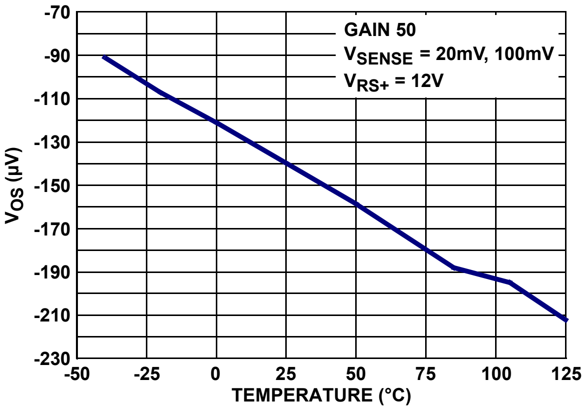


FIGURE 30. V_{OS} (µV) vs TEMPERATURE

Typical Performance Curves V_{CC} = 12V, R_L = 1M, unless otherwise specified. **(Continued)**

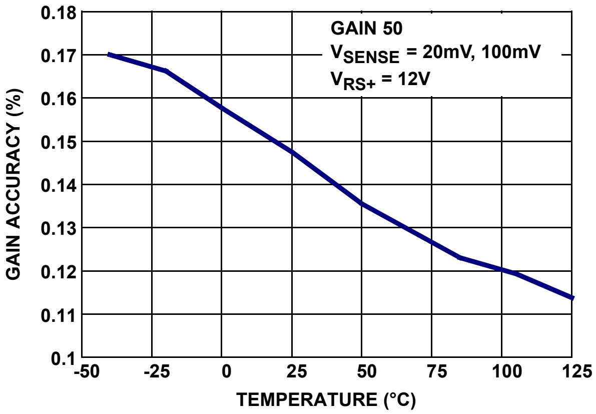


FIGURE 31. GAIN ACCURACY (%) vs TEMPERATURE

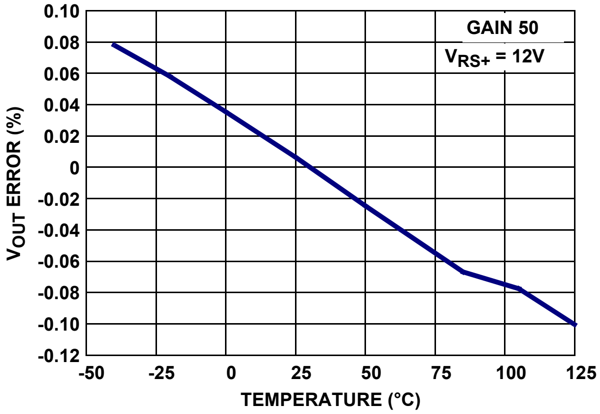


FIGURE 32. V_{OUT} ERROR (%) vs TEMPERATURE

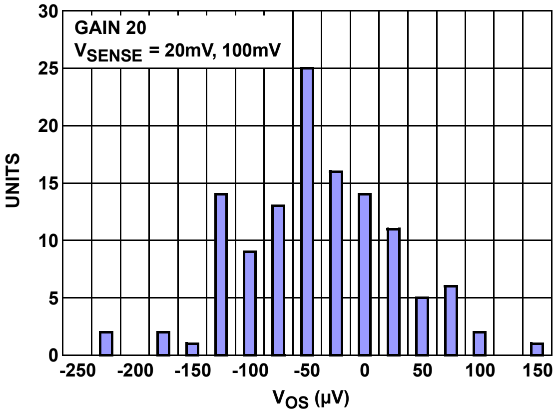


FIGURE 33. V_{OS} (µV) DISTRIBUTION AT +25°C, V_{RS+} = 12V, QUANTITY: 100

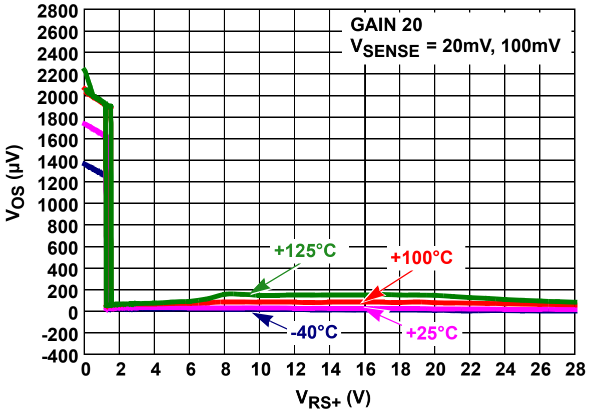


FIGURE 34. V_{OS} vs V_{RS+}

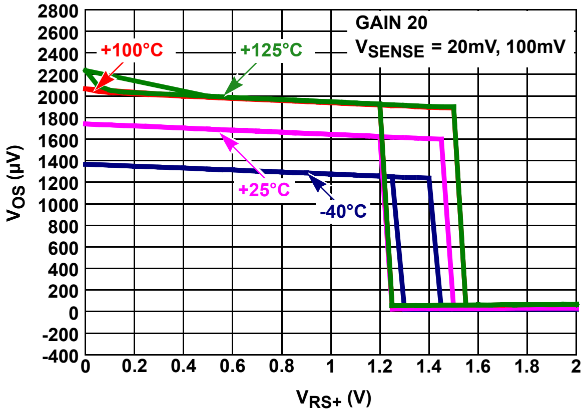


FIGURE 35. V_{OS} vs V_{RS+}

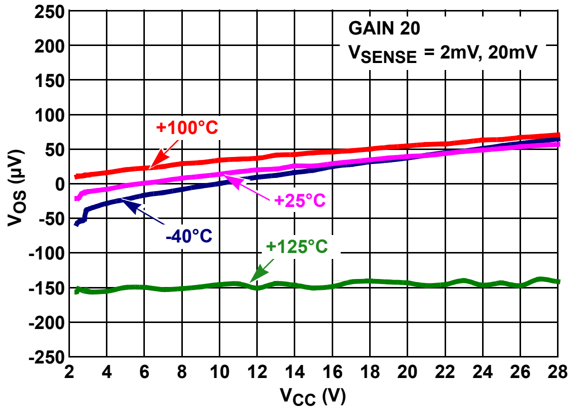


FIGURE 36. V_{OS} vs V_{CC}, V_{RS+} = 12V

Typical Performance Curves V_{CC} = 12V, R_L = 1M, unless otherwise specified. **(Continued)**

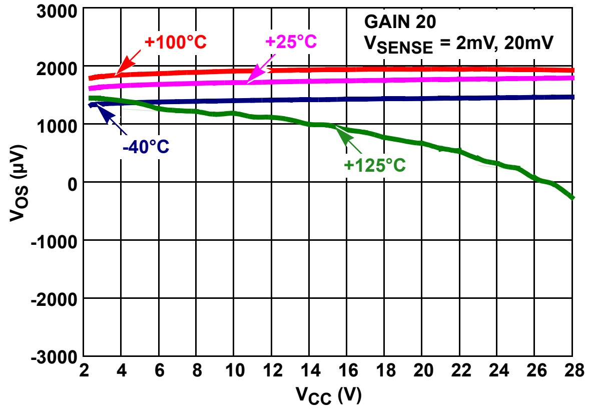


FIGURE 37. V_{OS} vs V_{CC}, V_{RS+} = 0.1V

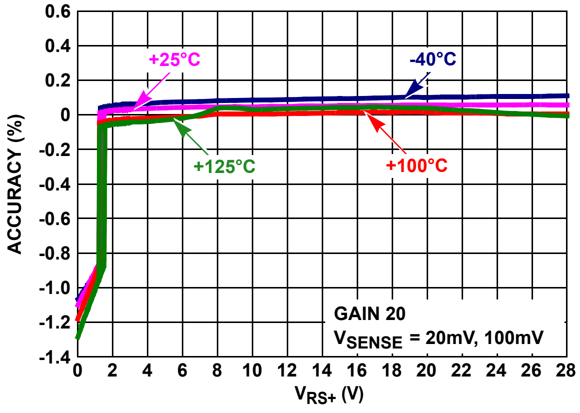


FIGURE 38. GAIN ACCURACY vs V_{RS+} = 0V TO 28V

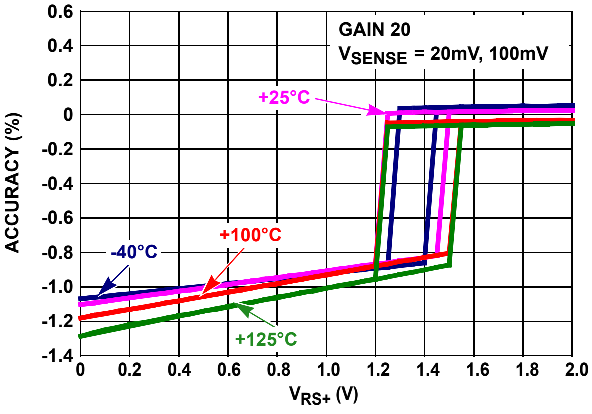


FIGURE 39. GAIN ACCURACY vs V_{RS+} = 0V TO 2V

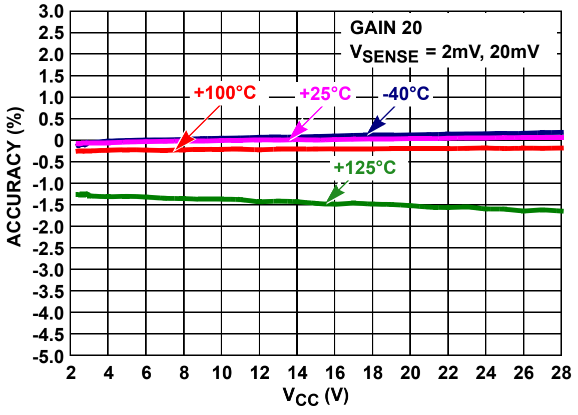


FIGURE 40. GAIN ACCURACY vs V_{CC}, HIGH-SIDE

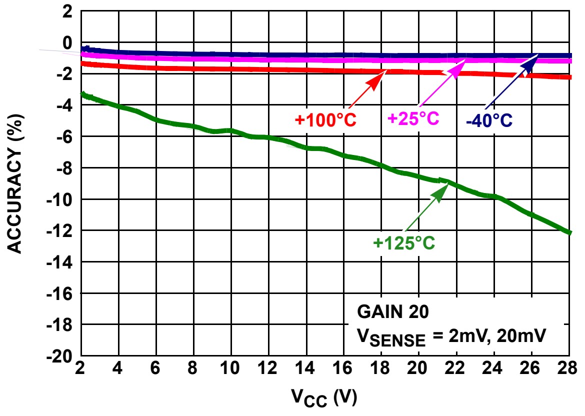


FIGURE 41. GAIN ACCURACY vs V_{CC}, LOW-SIDE

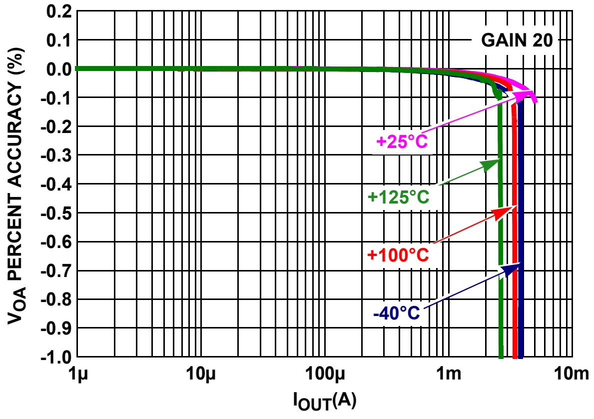


FIGURE 42. NORMALIZED V_{OA} vs I_{OUT}

Typical Performance Curves V_{CC} = 12V, R_L = 1M, unless otherwise specified. **(Continued)**

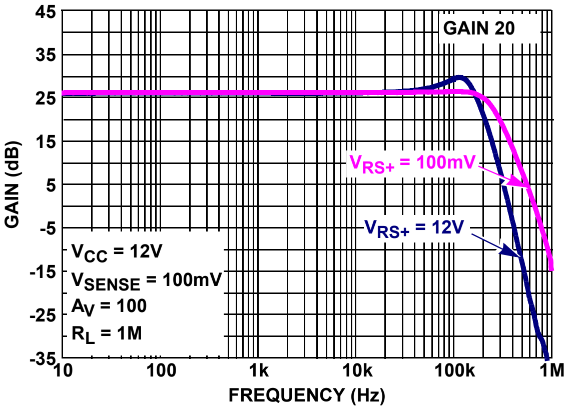


FIGURE 43. GAIN vs FREQUENCY V_{RS+} = 100mV/12V, V_{SENSE} = 100mV, V_{OUT} = 50mV$_{P-P}$

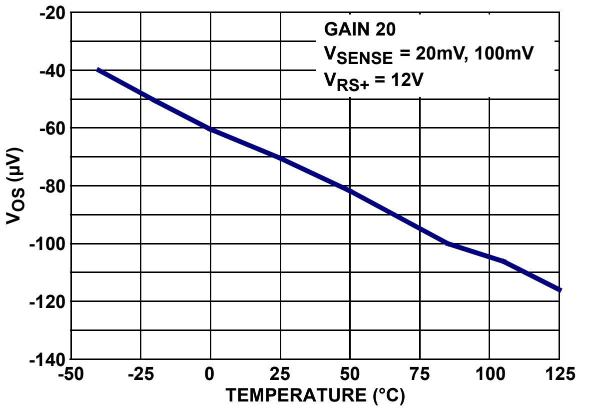


FIGURE 44. V_{OS} (µV) vs TEMPERATURE

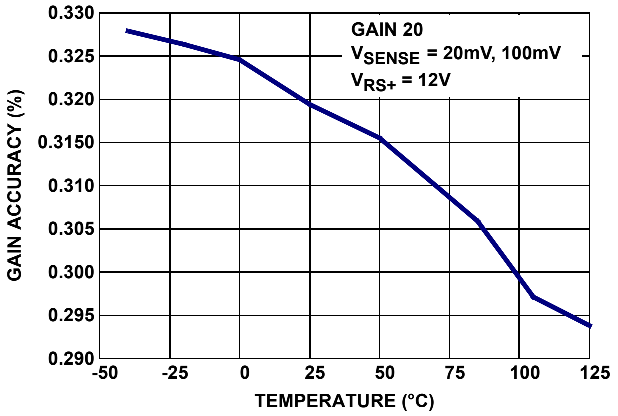


FIGURE 45. GAIN ACCURACY (%) vs TEMPERATURE

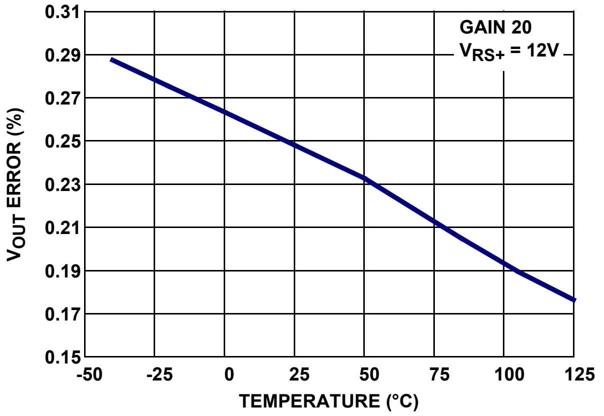


FIGURE 46. V_{OUT} ERROR (%) vs TEMPERATURE

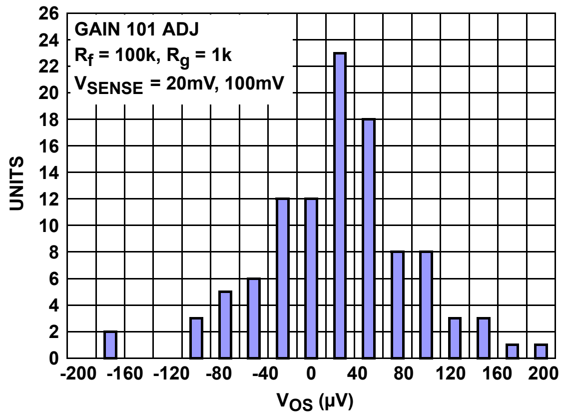


FIGURE 47. V_{OS} (µV) DISTRIBUTION AT +25°C, V_{RS+} = 12V, QUANTITY: 100

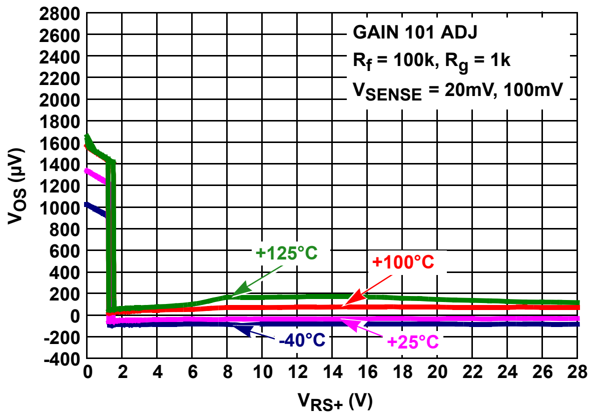


FIGURE 48. V_{OS} vs V_{RS+}

Typical Performance Curves V_{CC} = 12V, R_L = 1M, unless otherwise specified. (Continued)

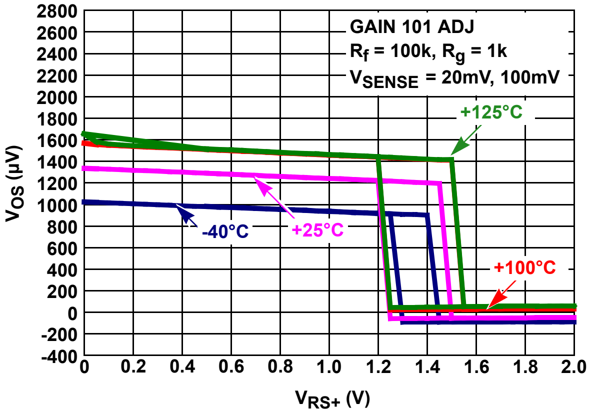


FIGURE 49. V_{OS} vs V_{RS+}

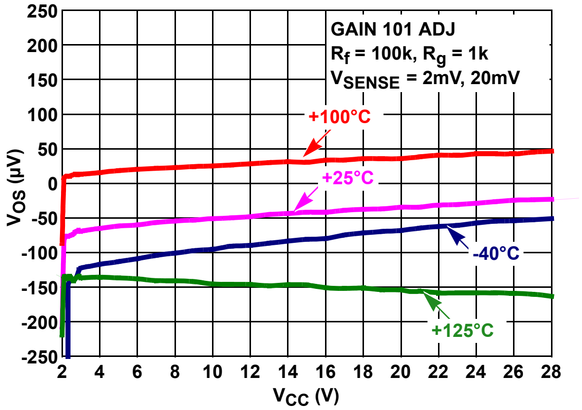


FIGURE 50. V_{OS} vs V_{CC}, HIGH-SIDE

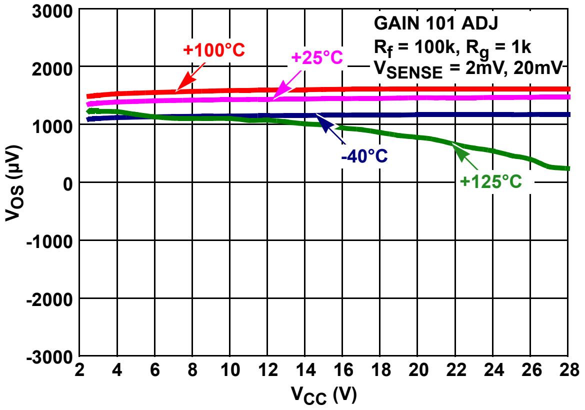


FIGURE 51. V_{OS} vs V_{CC}, LOW-SIDE

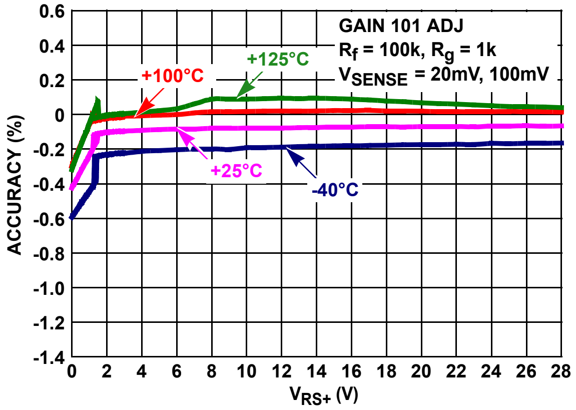


FIGURE 52. GAIN ACCURACY vs V_{RS+} = 0V TO 28V

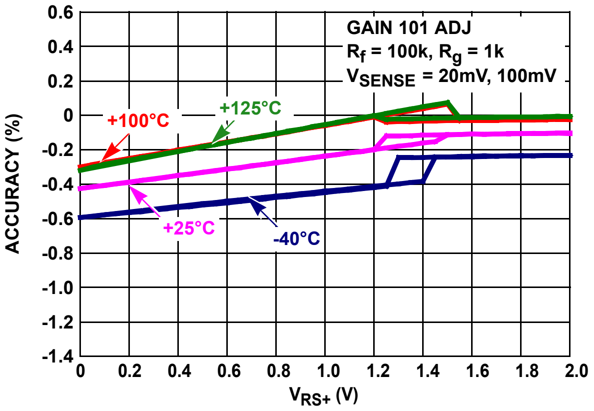


FIGURE 53. GAIN ACCURACY vs V_{RS+} = 0V TO 2V

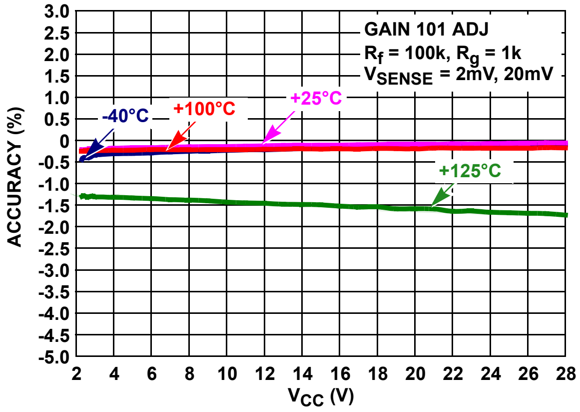


FIGURE 54. GAIN ACCURACY vs V_{CC}, V_{RS+} = 12V

Typical Performance Curves V_{CC} = 12V, R_L = 1M, unless otherwise specified. **(Continued)**

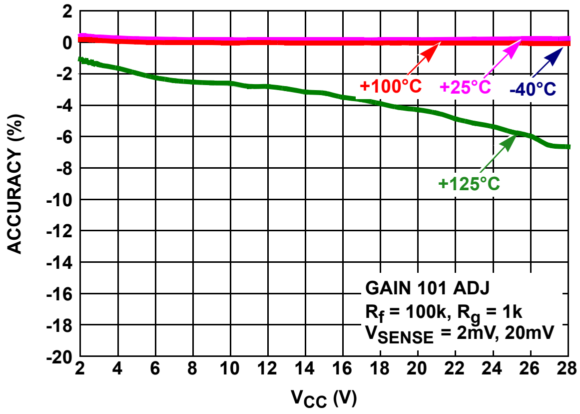


FIGURE 55. GAIN ACCURACY vs V_{CC}, V_{RS+} = 0.1V

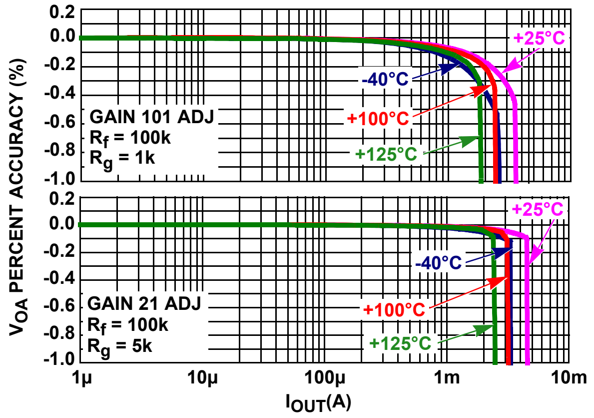


FIGURE 56. NORMALIZED V_{OA} vs I_{OUT}

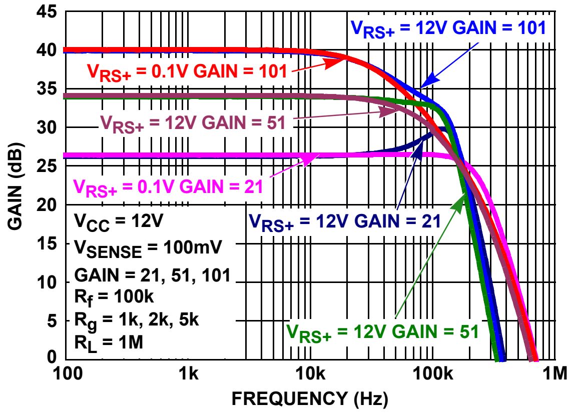


FIGURE 57. GAIN vs FREQUENCY V_{RS+} = 100mV/12V, V_{SENSE} = 100mV, V_{OUT} = 50mV_{P-P}

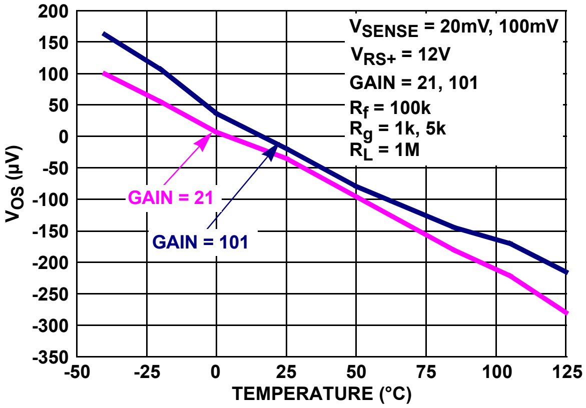


FIGURE 58. V_{OS} (µV) vs TEMPERATURE

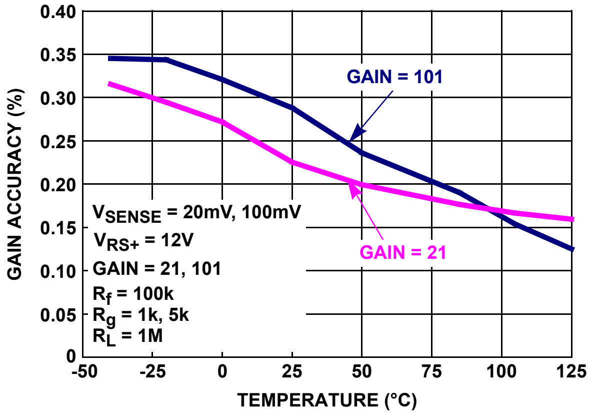


FIGURE 59. GAIN ACCURACY (%) vs TEMPERATURE

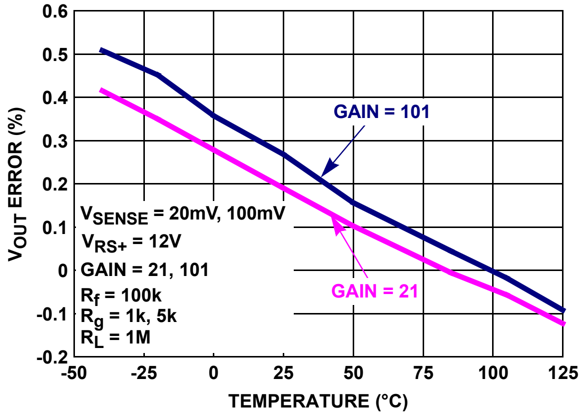


FIGURE 60. V_{OUT} ERROR (%) vs TEMPERATURE

Test Circuits and Waveforms

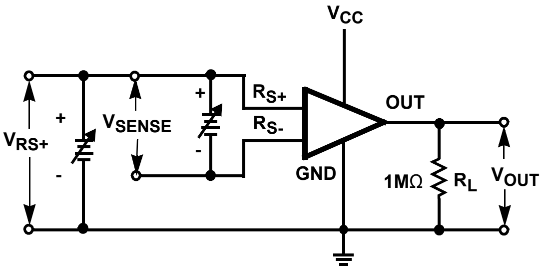


FIGURE 61. I_S, V_{OS}, V_{OA}, CMRR, PSRR, GAIN ACCURACY

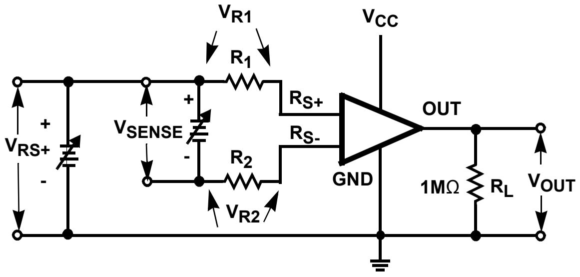


FIGURE 62. INPUT BIAS CURRENT, LEAKAGE CURRENT

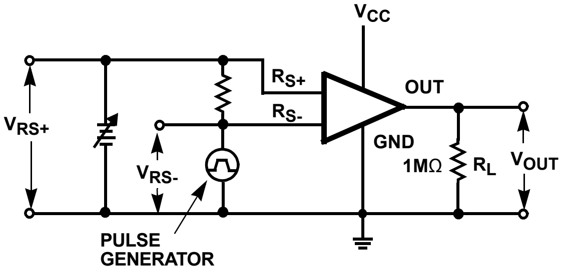


FIGURE 63. t_S, SATURATION RECOVERY TIME

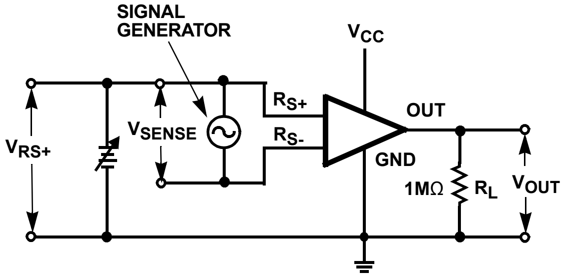


FIGURE 64. GAIN vs FREQUENCY

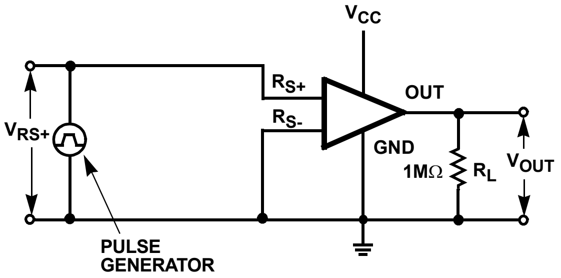


FIGURE 65. SLEW RATE

Applications Information

Functional Description

The ISL28006-20, ISL28006-50 and ISL28006-100 are single supply, uni-directional current sense amplifiers with fixed gains of 20V/V, 50V/V and 100V/V respectively. The ISL28006-ADJ is single supply, uni-directional current sense amplifier with an adjustable gain via external resistors (see Figure 70). The ISL28006-ADJ is stable for gains of 20 and higher.

The ISL28006 is a 2-stage amplifier. Figure 66 shows the active circuitry for high-side current sense applications where the sense voltage is between 1.35V to 28V. Figure 67 shows the active circuitry for ground sense applications where the sense voltage is between 0V to 1.35V.

The first stage is a bi-level trans-conductance amp and level translator. The gm stage converts the low voltage drop (V_{SENSE}) sensed across an external milli-ohm sense resistor, to a current (@ gm = 21.3µA/V). The trans-conductance amplifier forces a current through R_1 resulting to a voltage drop across R_1 that is equal to the sense voltage (V_{SENSE}). The current through R_1 is mirrored across R_5 creating a ground-referenced voltage at the input of the second amplifier equal to V_{SENSE}.

The second stage is responsible for the overall gain and frequency response performance of the device. The fixed gains (20, 50, 100) are set with internal resistors R_f and R_g. The variable gain (ADJ) has an additional FB pin and uses external gain resistors to set the gain of the output. For the fixed gain amps the only external component needed is a current sense resistor (typically 0.001Ω to 0.01Ω, 1W to 2W).

The transfer function for the fixed gain parts is given in Equation 1.

$$V_{OUT} = GAIN \times (I_S R_S + V_{OS}) \qquad \text{(EQ. 1)}$$

The transfer function for the adjustable gain part is given in Equation 2.

$$V_{OUT} = \left(1 + \frac{R_F}{R_G}\right)(I_S R_S + V_{OS}) \qquad \text{(EQ. 2)}$$

The input gm stage derives its ~2.86µA supply current from the input source through the R_{S+} terminal as long as the sensed voltage at the R_{S+} pin is >1.35V and the gm_{HI} amplifier is selected. When the sense voltage at R_{S+} drops below the 1.35V threshold, the gm_{LO} amplifier kicks in and the gm_{LO} output current reverses, flowing out of the R_{S-} pin.

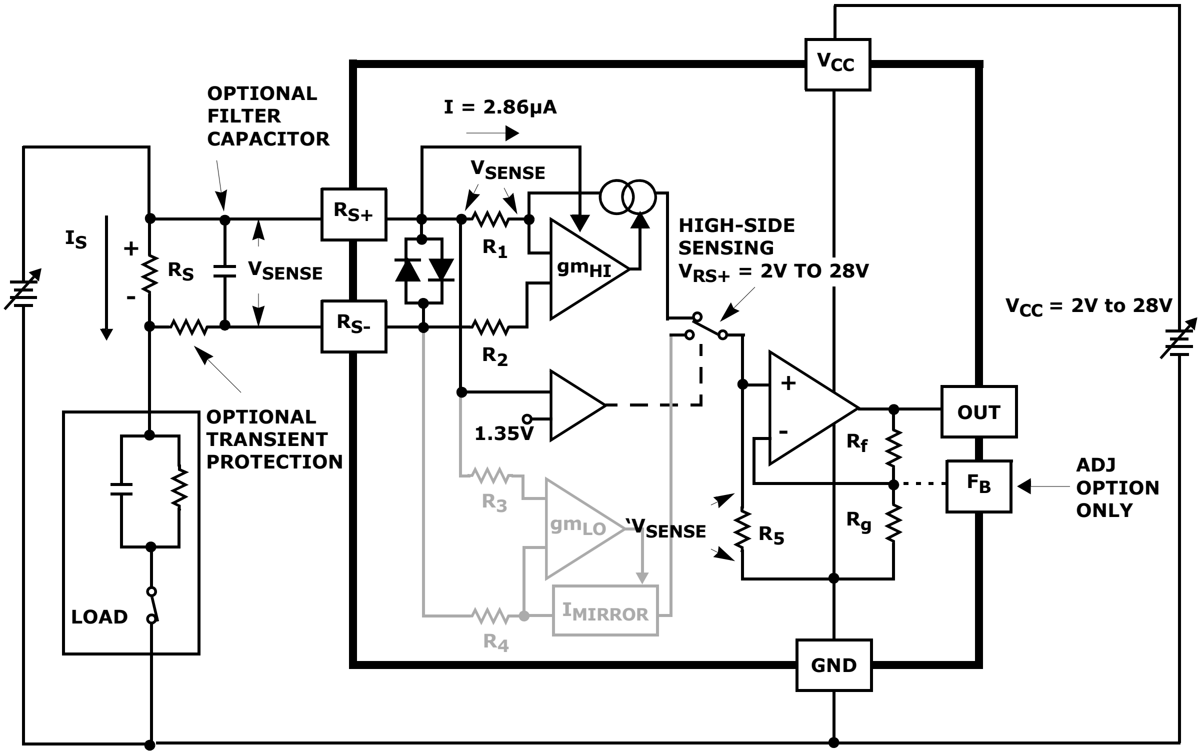


FIGURE 66. HIGH-SIDE CURRENT DETECTION

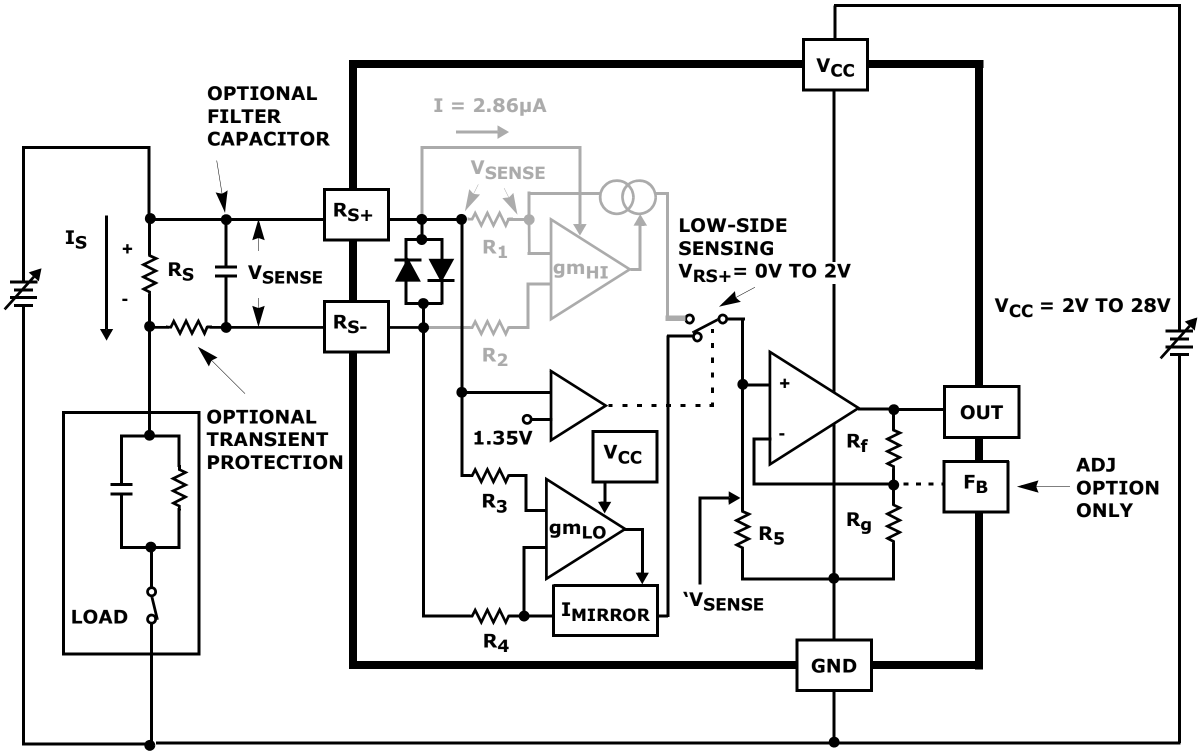


FIGURE 67. LOW-SIDE CURRENT DETECTION

Hysteretic Comparator

The input trans-conductance amps are under control of a hysteretic comparator operating from the incoming source voltage on the R_{S+} pin (Figure 66). The comparator monitors the voltage on R_{S+} and switches the sense amplifier from the low-side gm amp to the high-side gm amplifier whenever the input voltage at R_{S+} increases above the 1.35V threshold. Conversely, a decreasing voltage on the R_{S+} pin, causes the hysteric comparator to switch from the high-side gm amp to the low-side gm amp as the voltage decreases below 1.35V. It is that low-side sense gm amplifier that gives the ISL28006 the proprietary ability to sense current all the way to 0V. Negative voltages on the R_{S+} or R_{S-} are beyond the sensing voltage range of this amplifier.

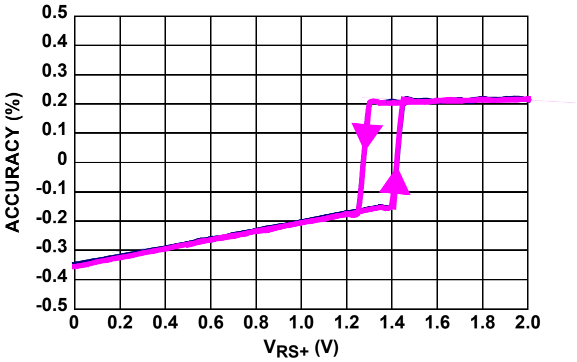


FIGURE 68. GAIN ACCURACY vs V_{RS+} = 0V TO 2V

Typical Application Circuit

Figure 70 shows the basic application circuit and optional protection components for switched-load applications. For applications where the load and the power source is permanently connected, only an external sense resistor is needed. For applications where fast transients are caused by hot plugging the source or load, external protection components may be needed. The external current limiting resistor (R_P) in Figure 70 may be required to limit the peak current through the internal ESD diodes to <20mA. This condition can occur in applications that experience high levels of in-rush current causing high peak voltages that can damage the internal ESD diodes. An R_P resistor value of 100Ω will provide protection for a 2V transient with the maximum of 20mA flowing through the input while adding only an additional 13µV (worse case over-temperature) of V_{OS}. Refer to Equation 3:

$$((R_P \times I_{RS-}) = (100\Omega \times 130nA) = 13\mu V) \qquad \text{(EQ. 3)}$$

Switching applications can generate voltage spikes that can overdrive the amplifier input and drive the output of the amplifier into the rails, resulting in a long overload recover time. Capacitors C_M and C_D filter the common mode and differential voltage spikes.

Error Sources

There are 3 dominant error sources: gain error, input offset voltage error and Kelvin voltage error (see Figure 69). The gain error is dominated by the internal resistance matching tolerances. The remaining errors appear as sense voltage errors at the input to the amplifier. They are V_{OS} of the amplifier and Kelvin voltage errors. If the transient protection resistor is added, an additional V_{OS} error can result from the IxR voltage due to input bias current. The limiting resistor should only be added to the R_{S-} input, due to the high-side gm amplifier (gm_{HI}) sinking several micro amps of current through the R_{S+} pin.

Layout Guidelines

The Kelvin Connected Sense Resistor

The source of Kelvin voltage errors is illustrated in Figure 69. The resistance of 1/2 Oz copper is ~1mΩ per square with a TC of ~3900ppm/°C (0.39%/°C). When you compare this unwanted parasitic resistance with the total 1mΩ to 10mΩ resistance of the sense resistor, it is easy to see why the sense connection must be chosen very carefully. For example, consider a maximum current of 20A through a 0.005Ω sense resistor, generating a V_{SENSE} = 0.1 and a full scale output voltage of 10V (G = 100). Two side contacts of only 0.25 square per contact puts the V_{SENSE} input about 0.5 x 1mΩ away from the resistor end capacitor. If only 10A the 20A total current flows through the kelvin path to the resistor, you get an error voltage of 10mV (10A x 0.5sq x 0.001Ω/sq. = 10mV) added to the 100mV sense voltage for a sense voltage error of 10% (0.110V-0.1)/0.1V)x 100.

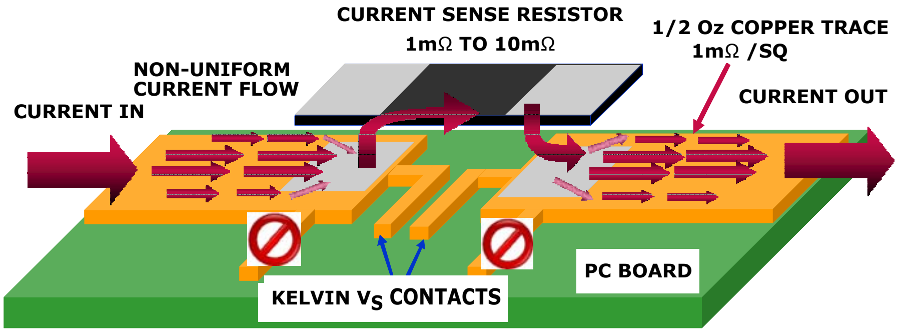


FIGURE 69. PC BOARD CURRENT SENSE KELVIN CONNECTION

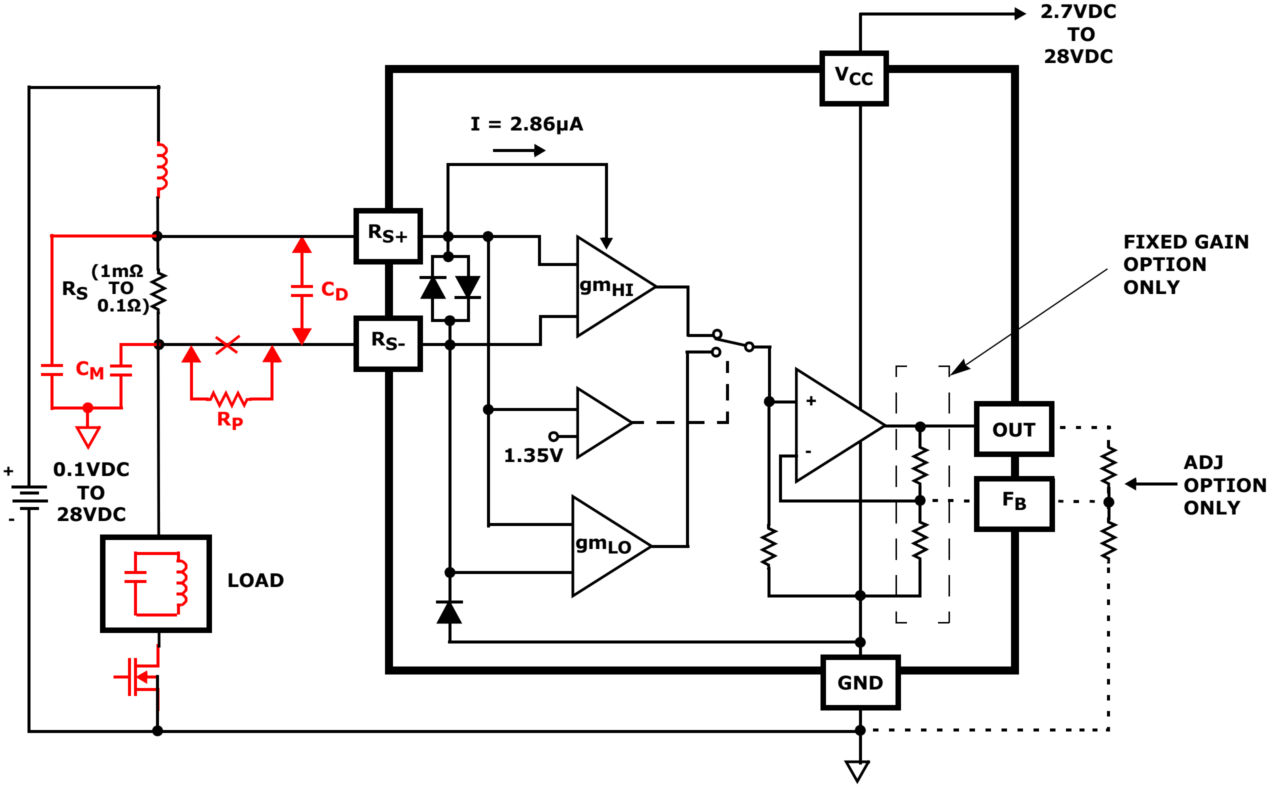


FIGURE 70. TYPICAL APPLICATION CIRCUIT

Overall Accuracy (V_{OA} %)

V_{OA} is defined as the total output accuracy Referred-to-Output (RTO). The output accuracy contains all offset and gain errors, at a single output voltage. Equation 4 is used to calculate the % total output accuracy.

$$V_{OA} = 100 \times \left(\frac{V_{OUT}\text{actual} - V_{OUT}\text{expected}}{V_{OUT}\text{expected}}\right) \qquad \text{(EQ. 4)}$$

where

V_{OUT} Actual = V_{SENSE} x GAIN

Example: Gain = 100, For 100mV V_{SENSE} input we measure 10.1V. The overall accuracy (V_{OA}) is 1% as shown in Equation 5.

$$V_{OA} = 100 \times \left(\frac{10.1 - 10}{10}\right) = 1\% \qquad \text{(EQ. 5)}$$

Power Dissipation

It is possible to exceed the +150°C maximum junction temperatures under certain load and power supply conditions. It is therefore important to calculate the maximum junction temperature (T_{JMAX}) for all applications to determine if power supply voltages, load conditions, or package type need to be modified to remain in the safe operating area. These parameters are related using Equation 6:

$$T_{JMAX} = T_{MAX} + \theta_{JA} x PD_{MAXTOTAL} \qquad \text{(EQ. 6)}$$

where:

- $P_{DMAXTOTAL}$ is the sum of the maximum power dissipation of each amplifier in the package (PD_{MAX})
- PD_{MAX} for each amplifier can be calculated using Equation 7:

$$PD_{MAX} = V_S \times I_{qMAX} + (V_S - V_{OUTMAX}) \times \frac{V_{OUTMAX}}{R_L} \qquad \text{(EQ. 7)}$$

where:

- T_{MAX} = Maximum ambient temperature
- θ_{JA} = Thermal resistance of the package
- PD_{MAX} = Maximum power dissipation of 1 amplifier
- V_{CC} = Total supply voltage
- I_{qMAX} = Maximum quiescent supply current of 1 amplifier
- V_{OUTMAX} = Maximum output voltage swing of the application

R_L = Load resistance

Revision History

The revision history provided is for informational purposes only and is believed to be accurate, but not warranted. Please go to web to make sure you have the latest Rev.

DATE	REVISION	CHANGE
9/2/10	FN6548.4	Added -T7A tape and reel options to Ordering Information Table for all packages.
5/12/10	FN6548.3	Added Note 4 to Part Marking Column in "Ordering Information" on page 3. Corrected hyperlinks in Notes 1 and 3 in "Ordering Information" on page 3.
4/8/10		Removed "Coming Soon" from evaluation boards in "Ordering Information" on page 3.
4/7/10		Added "Related Literature*(see page 23)" on page 1 Updated Package Drawing Number in the "Ordering Information" on page 3 for the 20V, 50V and 100V options from MDP0038 to P50.64A. Revised package outline drawing from MDP0038 to P5.064A on page 24. MDP0038 package contained 2 packages for both the 5 and 6 Ld SOT-23. MDP0038 was obsoleted and the packages were separated and made into 2 separate package outline drawings; P5.064A and P6.064A. Changes to the 5 Ld SOT-23 were to move dimensions from table onto drawing, add land pattern and add JEDEC reference number.
3/10/10	FN6548.2	Releasing adjustable gain option. Added adjustable block diagram (Page 2), Added adjustable gain limits to electrical spec table, added Figures 47 through 60, Added +85°C curves to Figures 6 thru 14, 20 thru 28, 34 thru 42, and Figures 48 thru 56. Modified Figure 70.
2/4/10	FN6548.1	-Page 1: Edited last sentence of paragraph 2. Moved order of GAIN listings from 20, 50, 100 to 100, 50, 20 in the 3rd paragraph. Under Features ....removed "Low Input Offset Voltage 250µV, max" Under Features .... moved order of parts listing from 20, 50, 100 (from top to bottom) to 100, 50, 20. -Page 3: Removed coming soon on ISL28006FH50Z and ISL28006FH20Z and changes the order or listing them to 100, 50, 20. -Page 5: VOA test. Under conditions column ...deleted 20mV to. It now reads ... Vsense = 100mV SR test. Under conditions column ..deleted what was there. It now reads ... Pulse on RS+pin, See Figure 51 -Page 6: ts test. Removed Gain = 100 and Gain = 100V/V in both description and conditions columns respectively. -Page 9: Added VRS+= 12V to Figures 16, 17, 18. -Page 11: Added VRS+= 12V to Figures 30, 31, 32. -Page 13 & 14: Added VRS+= 12V to Figures 44, 45, 46. -Page 14 Added Figure 51 and adjusted figure numbers to account for the added figure. -Figs 8, 26, and 40 change "HIGH SIDE" to "VRS = 12V", where RS is subscript. -Figs 9, 27, and 41 change "LOW SIDE" to "VRS = 0.1V", where RS is subscript.
12/14/09	FN6548.0	Initial Release

Products

Intersil Corporation is a leader in the design and manufacture of high-performance analog semiconductors. The Company's products address some of the industry's fastest growing markets, such as, flat panel displays, cell phones, handheld products, and notebooks. Intersil's product families address power management and analog signal processing functions. Go to www.intersil.com/products for a complete list of Intersil product families.

*For a complete listing of Applications, Related Documentation and Related Parts, please see the respective device information page on intersil.com: ISL28006

To report errors or suggestions for this datasheet, please go to www.intersil.com/askourstaff

FITs are available from our website at http://rel.intersil.com/reports/search.php

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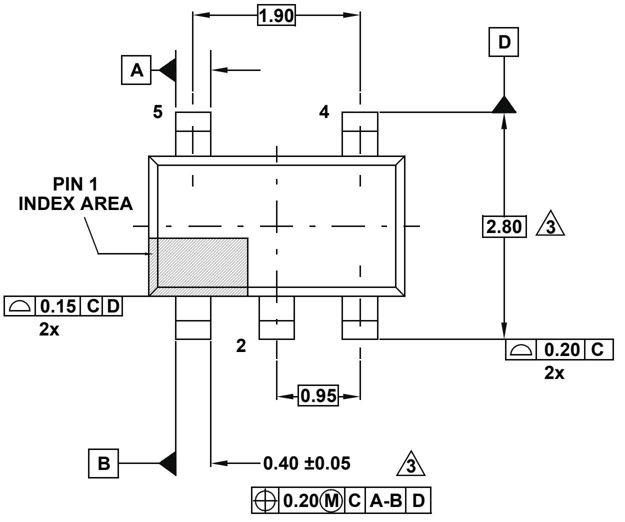
Package Outline Drawing

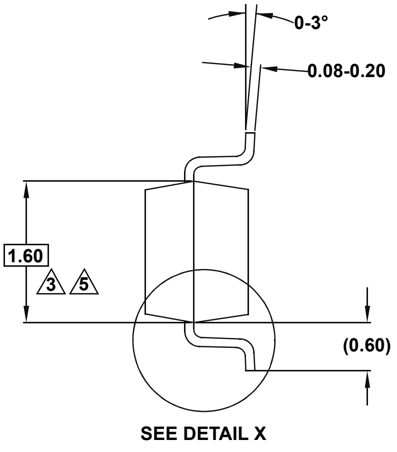
P5.064A

5 LEAD SMALL OUTLINE TRANSISTOR PLASTIC PACKAGE

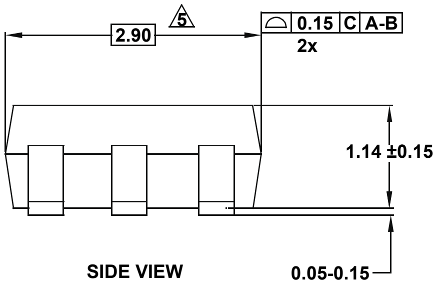
Rev 0, 2/10



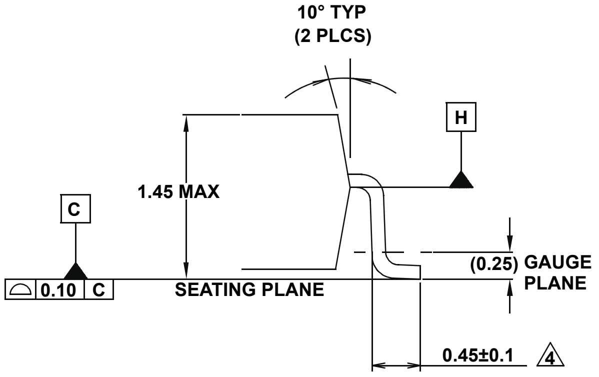
TOP VIEW



END VIEW



SIDE VIEW



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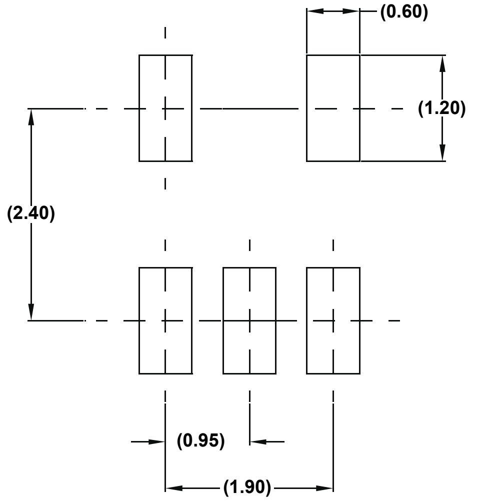


TYPICAL RECOMMENDED LAND PATTERN

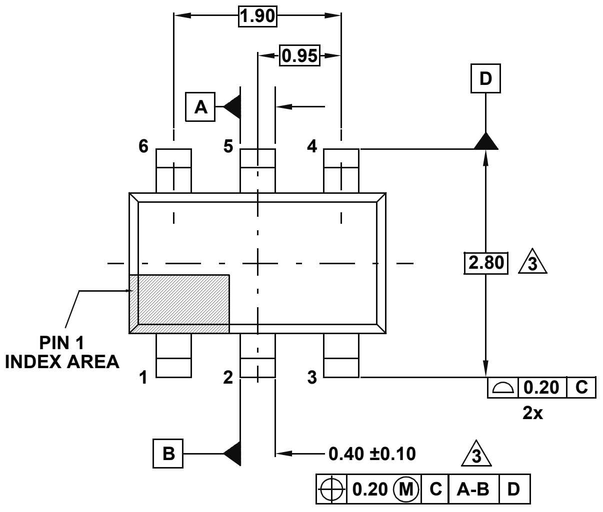
NOTES:

1. Dimensions are in millimeters.
   Dimensions in ( ) for Reference Only.
2. Dimensioning and tolerancing conform to ASME Y14.5M-1994.
3. Dimension is exclusive of mold flash, protrusions or gate burrs.
4. Foot length is measured at reference to guage plane.
5. This dimension is measured at Datum "H".
6. Package conforms to JEDEC MO-178AA.

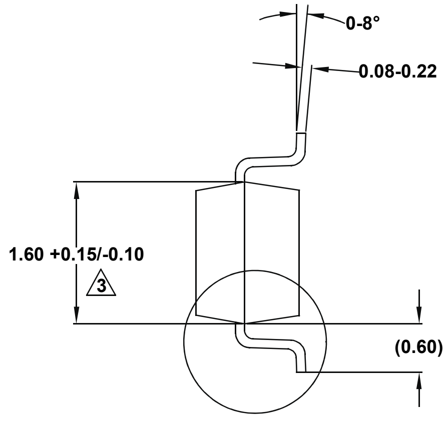
Package Outline Drawing

P6.064

6 LEAD SMALL OUTLINE TRANSISTOR PLASTIC PACKAGE
Rev 4, 2/10

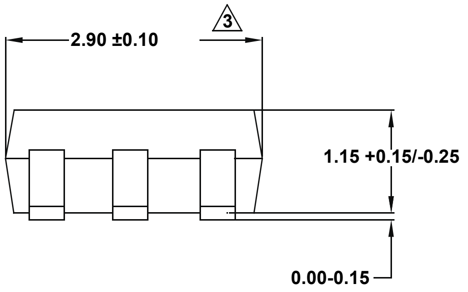


TOP VIEW

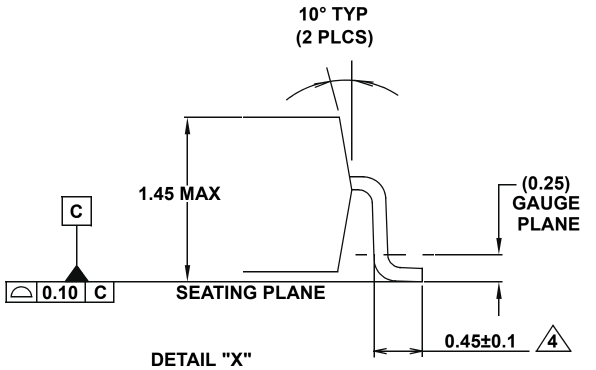


SEE DETAIL X

END VIEW



SIDE VIEW



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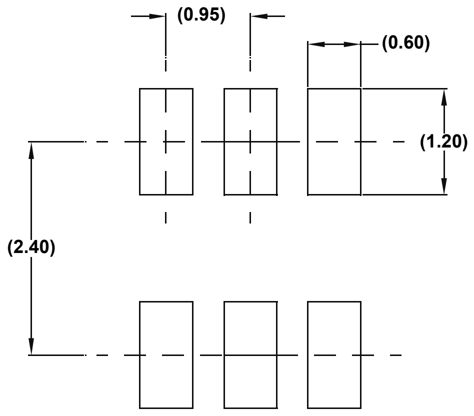

TYPICAL RECOMMENDED LAND PATTERN

NOTES:

1. Dimensions are in millimeters.
   Dimensions in ( ) for Reference Only.
2. Dimensioning and tolerancing conform to ASME Y14.5M-1994.
3. Dimension is exclusive of mold flash, protrusions or gate burrs.
4. Foot length is measured at reference to guage plane.
5. Package conforms to JEDEC MO-178AB.