## AS1115

## 64 LEDs, ${ }^{2}$ C Interfaced LED Driver with Keyscan

## 1 General Description

The AS1115 is a compact LED driver for 64 single LEDs or 8 digits of 7 -segments. The devices can be programmed via an $I^{2} \mathrm{C}$ compatible 2-wire interface.
Every segment can be individually addressed and updated separately. Only one external resistor (RSET) is required to set the current. LED brightness can be controlled by analog or digital means.
The devices include an integrated BCD code-B/HEX decoder, multiplex scan circuitry, segment and display drivers, and a 64-bit memory. Internal memory stores the shift register settings, eliminating the need for continuous device reprogramming.

All outputs of the AS1115 can be configured for key readback. Key-switch status is obtained by polling for up to 64 keys while 16 keys can be used to trigger an interrupt.
Additionally the AS1115 offers a diagnostic mode for easy and fast production testing.
The AS1115 features a low shutdown current of typically 200 nA , and an operational current of typically $350 \mu \mathrm{~A}$. The number of digits can be programmed, the devices can be reset by software, and an external clock is also supported.
The device is available in a QSOP-24 and the TQFN(4×4)-24 package.

## 2 Key Features

- $3.4 \mathrm{MHz}{ }^{2} \mathrm{C}$-Compatible Interface
- Individual LED Segment Control
- Readback for 16 Keys plus Interrupt
- Open and Shorted LED Error Detection
- Global or Individual Error Detection
- Hexadecimal- or BCD-Code for 7-Segment Displays
- 200nA Low-Power Shutdown Current (typ; data retained)
- Digital and Analog Brightness Control
- Display Blanked on Power-Up
- Drive Common-Cathode LED Displays
- Supply Voltage Range: 2.7 to 5.5 V
- Software Reset
- Optional External Clock
- Package:
- QSOP-24
- TQFN(4x4)-24


## 3 Applications

The AS1115 is ideal for seven-segment or dot matrix user interface displays of set-top boxes, VCRs, DVDplayers, washing machines, micro wave ovens, refrigerators and other white good or personal electronic applications.

Figure 1. AS1115-Typical Application Diagram


## 4 Pinout

## Pin Assignments

Figure 2. Pin Assignments (Top View)


## Pin Descriptions

Table 1. Pin Descriptions

| Pin Name | QSOP-24 | TQFN(4x4)-24 | Description |
| :---: | :---: | :---: | :--- |
| SDA | 1 | 22 | Serial-Data I/O. Open drain digital I/O IC data pin. |
| DIG0:DIG7 | $2-5,7-10$ | $1,2,4,5,6,7$, <br> 23,24 | Digit Drive Lines. Eight digit drive lines that sink current from the display <br> common cathode. Keyscan detection optional, but must be polled by the <br> $\mu$ Prozessor. |
| GND | 6 | 3 | Ground. |
| KEYA | 11 | 8 | Keyscan Input. Keyscan lines for key readback. Can be used for self- <br> adressing. |
| KEYB | 12 | 9 | Keyscan Input. Keyscan lines for key readback. <br> ISET$\quad 13$ |

## 5 Absolute Maximum Ratings

Stresses beyond those listed in Table 2 may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in Section 6 Electrical Characteristics on page 4 is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Table 2. Absolute Maximum Ratings

| Parameter |  | Min | Max | Units | Notes |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Input Voltage Range | VDD to GND | -0.3 | 7 | V |  |
|  | All other pins to GND | -0.3 | $\begin{gathered} 7 \text { or } \\ \operatorname{VDD}+0.3 \end{gathered}$ | V |  |
| Current | DIG0:DIG7 Sink Current |  | 500 | mA |  |
|  | SEGA:SEGG, SEGDP |  | 100 | mA |  |
| Humidity |  | 5 | 85 | \% | Non-condensing |
| Electrostatic Discharge | Digital outputs |  | 1000 | V | Norm: MIL 833 E method 3015 |
|  | All other pins |  | 1000 | V |  |
| Latch-Up Immunity |  | $\pm 100$ |  | mA | EIA/JESD78 |
| Thermal Resistance ©JA |  |  | 88 | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ | on PCB, QSOP-24 package |
|  |  |  | 30.5 | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ | on PCB, TQFN(4×4)-24 package |
| Ambient Temperature |  | -40 | +85 | ${ }^{\circ} \mathrm{C}$ |  |
| Storage Temperature |  | -55 | 150 | ${ }^{\circ} \mathrm{C}$ |  |
| Package Body Temperature |  |  | +260 | ${ }^{\circ} \mathrm{C}$ | The reflow peak soldering temperature (body temperature) specified is in accordance with IPC/ JEDEC J-STD-020D "Moisture/ Reflow Sensitivity Classification for Non-Hermetic Solid State Surface Mount Devices". The lead finish for Pb -free leaded packages is matte tin $(100 \% \mathrm{Sn})$. |

## 6 Electrical Characteristics

VDD $=2.7 \mathrm{~V}$ to 5.5 V, RSET $=9.53 \mathrm{k} \Omega$, $\mathrm{TAMB}=-40$ to $+85^{\circ} \mathrm{C}$, typ. values @ $\mathrm{TAMB}=+25^{\circ} \mathrm{C}$, VDD $=5.0 \mathrm{~V}$ (unless otherwise specified).
Table 3. Electrical Characteristics

| Symbol | Parameter | Conditions | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| VDD | Operating Supply Voltage |  | 2.7 |  | 5.5 | V |
| IDDSD | Shutdown Supply Current | All digital inputs at VDD or GND, TAMB $=+25^{\circ} \mathrm{C}$ |  | 0.2 | 2 | $\mu \mathrm{A}$ |
|  |  | single digit, TAMB $=+85^{\circ} \mathrm{C}$ |  |  | 4 | $\mu \mathrm{A}$ |
| IDD | Operating Supply Current | RSET = open circuit. |  | 0.35 | 0.6 | mA |
|  |  | All segments and decimal point on; ISEG $=-40 \mathrm{~mA}$. |  | 335 |  |  |
| fosc | Display Scan Rate | 8 digits scanned | 0.48 |  | 0.96 | kHz |
| IDIGIT | Digit Drive Sink Current | Vout $=0.65 \mathrm{~V}$ | 320 |  |  | mA |
| ISEG | Segment Drive Source Current | VDD $=5.0 \mathrm{~V}$, Vout $=(\mathrm{VDD} \mathrm{-1V})$ | -35 | -41 | -47 | mA |
| $\Delta$ ISEG | Segment Drive Current Matching |  |  | 3 |  | \% |
| ISEG | Segment Drive Source Current | Average Current |  |  | 47 | mA |

Table 4. Logic Inputs/Outputs Characteristics

| Symbol | Parameter | Conditions | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| IIH, IIL | Input Current SDA, SCL | VIN $=0 \mathrm{~V}$ or VDD | -1 |  | 1 | $\mu \mathrm{A}$ |
| VIH | Logic High Input Voltage SDA, SCL |  | 0.7xVDD |  |  | V |
| VIL | Logic Low Input Voltage SDA, SCL |  |  |  | 0.3xVDD | V |
| Vol(SDA) | SDA Output Low Voltage | ISINK $=3 \mathrm{~mA}$ |  |  | 0.4 | V |
| VkEYopen | Keyscan Open Input Voltage |  | 0.8xVDD |  |  | V |
| VkeYshort | Keyscan Short Input Voltage |  |  |  | $0.7 \times \mathrm{VdD}$ | V |
| Vol(IRQ) | Interrupt Output Low Voltage | Isink $=3 \mathrm{~mA}$ |  |  | 0.4 | V |
| $\Delta \mathrm{V}$ I | Hysteresis Voltage | DIN, CLK, LD/CS |  | 1 |  | V |
| $\mathrm{C}_{\mathrm{B}}$ | Capacitive Load for each Bus Line |  |  |  | 400 | pF |
|  | Open Detection Level Threshold |  | $\begin{aligned} & 0.7 x \\ & \text { VDD } \end{aligned}$ | $\begin{gathered} 0.75 x \\ \text { VDD } \end{gathered}$ | $\begin{aligned} & 0.8 x \\ & \text { VDD } \end{aligned}$ | V |
|  | Short Detection Level Threshold |  | $\begin{aligned} & 0.05 x \\ & \text { VDD } \end{aligned}$ | $\begin{aligned} & 0.1 x \\ & \text { VDD } \end{aligned}$ | $\begin{aligned} & 0.15 x \\ & \text { VDD } \end{aligned}$ | V |

Table 5. Timing Characteristics ( $C_{B}=100 \mathrm{pF}$ (max) on each Bus Line)

| Symbol | Parameter | Conditions | Min | Typ | Max | Unit |
| :---: | :--- | :---: | :---: | :---: | :---: | :---: |
| fscl | SCL Frequency |  | 0.1 |  | 3.4 | MHz |
| tBuF | Bus Free Time Between STOP and <br> START Conditions |  | 1.3 |  |  | $\mu \mathrm{~s}$ |
| tholdstart | Hold Time for Repeated <br> START Condition |  | 160 |  |  | ns |
| tLow | SCL Low Period |  | 160 |  |  | ns |
| tHIGH | SCL High Period |  | 60 |  |  | ns |
| tSETUPSTART | Setup Time for Repeated <br> START Condition | 160 |  |  | ns |  |
| tSETUPDATA | Data Setup Time |  | 10 |  |  | ns |
| tholdDATA | Data Hold Time |  | 10 |  | 70 | ns |
| tRISE(SCL) | SCL Rise Time |  |  | ns |  |  |

Table 5. Timing Characteristics ( $C_{B}=100 \mathrm{pF}$ (max) on each Bus Line)

| Symbol | Parameter | Conditions | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| tRISE(SCL1) | SCL Rise Time after Repeated START Condition and After an ACK Bit |  | 10 |  | 80 | ns |
| tFALL(SCL) | SCL Fall Time |  | 10 |  | 40 | ns |
| tRISE(SDA) | SDA Rise Time |  | 10 |  | 80 | ns |
| tFALL(SDA) | SDA Fall Time |  | 10 |  | 80 | ns |
| tSETUPSTOP | STOP Condition Setup Time |  | 160 |  |  | ns |
| tSPIKESUP | Pulse Width of Spike Suppressed |  |  | 50 |  | ns |
| Key Readback |  |  |  |  |  |  |
|  | Debounce Time |  |  | 20 |  | ms |

## Notes:

1. The Min / Max values of the Timing Characteristics are guaranteed by design.
2. All limits are guaranteed. The parameters with min and max values are guaranteed with production tests or SQC (Statistical Quality Control) methods.

Figure 3. Timing Diagram


## 7 Typical Operating Characteristics

Rset $=9.53 \mathrm{k} \Omega$, VRset $=$ Vdd;

Figure 4. Display Scan Rate vs. Supply Voltage;


Figure 6. Segment Current vs. Temperature;


Figure 8. Segment Current vs. Supply Voltage;


Figure 5. Display Scan Rate vs. Temperature;


Figure 7. Segment Current vs. Rset;


Figure 9. Segment Current vs. VDd; VRset $=2.8 \mathrm{~V}$


Figure 10. VDIGIT vs. IDIGIT


Figure 12. ISEG vs. VSEG; VDD $=5 \mathrm{~V}$


Figure 14. IsEg vs. VsEG; VDD $=3.3 \mathrm{~V}$


Figure 11. Input High Level vs. Supply Voltage


Figure 13. Iseg vs. Vseg; Vdd $=4 \mathrm{~V}$


Figure 15. Iseg vs. Vseg; VdD $=2.7 \mathrm{~V}$


## 8 Detailed Description

## Block Diagram

Figure 16. AS1115-Block Diagram (QSOP-24 Package)


Figure 17. ESD Structure

|  | valid for the pins: <br> - IRQ <br> - SCL <br> - SDA <br> - ISET <br> - SEGA-G, SEGDP <br> - KEYA, KEYB |  | valid for the pins: <br> - DIG0 to DIG7 |
| :---: | :---: | :---: | :---: |

## $I^{2}$ C Interface

The AS1115 supports the $I^{2} \mathrm{C}$ serial bus and data transmission protocol in high-speed mode at 3.4 MHz . The AS1115 operates as a slave on the $I^{2} \mathrm{C}$ bus. The bus must be controlled by a master device that generates the serial clock (SCL), controls the bus access, and generates the START and STOP conditions. Connections to the bus are made via the open-drain I/O pins SCL and SDA.

Figure 18. $1^{12} \mathrm{C}$ Interface Initialisation


Default values at power up: $\mathrm{A} 1=\mathrm{A} 0=0$

Figure 19. Bus Protocol


The bus protocol (as shown in Figure 19) is defined as:

- Data transfer may be initiated only when the bus is not busy.
- During data transfer, the data line must remain stable whenever the clock line is HIGH. Changes in the data line while the clock line is HIGH will be interpreted as control signals.

The bus conditions are defined as:

- Bus Not Busy. Data and clock lines remain HIGH.
- Start Data Transfer. A change in the state of the data line, from HIGH to LOW, while the clock is HIGH, defines a START condition.
- Stop Data Transfer. A change in the state of the data line, from LOW to HIGH, while the clock line is HIGH, defines the STOP condition.
- Data Valid. The state of the data line represents valid data, when, after a START condition, the data line is stable for the duration of the HIGH period of the clock signal. There is one clock pulse per bit of data.
Each data transfer is initiated with a START condition and terminated with a STOP condition. The number of data bytes transferred between START and STOP conditions is not limited and is determined by the master device. The information is transferred byte-wise and each receiver acknowledges with a ninth-bit. Within the $\mathrm{I}^{2} \mathrm{C}$ bus specifications a high-speed mode ( 3.4 MHz clock rate) is defined.
- Acknowledge: Each receiving device, when addressed, is obliged to generate an acknowledge after the reception of each byte. The master device must generate an extra clock pulse that is associated with this acknowledge bit. A device that acknowledges must pull down the SDA line during the acknowledge clock pulse in such a way that the SDA line is stable LOW during the HIGH period of the acknowledge clock pulse. Of course, setup and hold times must be taken into account. A master must signal an end of data to the slave by not generating an
acknowledge bit on the last byte that has been clocked out of the slave. In this case, the slave must leave the data line HIGH to enable the master to generate the STOP condition.
- Figure 19 on page 9 details how data transfer is accomplished on the $I^{2} \mathrm{C}$ bus. Depending upon the state of the R/ $\bar{W}$ bit, two types of data transfer are possible:
- Master Transmitter to Slave Receiver. The first byte transmitted by the master is the slave address, followed by a number of data bytes. The slave returns an acknowledge bit after the slave address and each received byte.
- Slave Transmitter to Master Receiver. The first byte, the slave address, is transmitted by the master. The slave then returns an acknowledge bit. Next, a number of data bytes are transmitted by the slave to the master. The master returns an acknowledge bit after all received bytes other than the last byte. At the end of the last received byte, a not-acknowledge is returned. The master device generates all of the serial clock pulses and the START and STOP conditions. A transfer is ended with a STOP condition or a repeated START condition. Since a repeated START condition is also the beginning of the next serial transfer, the bus will not be released.

The AS1115 can operate in the following slave modes:

- Slave Receiver Mode. Serial data and clock are received through SDA and SCL. After each byte is received, an acknowledge bit is transmitted. START and STOP conditions are recognized as the beginning and end of a serial transfer. Address recognition is performed by hardware after reception of the slave address and direction bit.
- Slave Transmitter Mode. The first byte (the slave address) is received and handled as in the slave receiver mode. However, in this mode the direction bit will indicate that the transfer direction is reversed. Serial data is transmitted on SDA by the AS1115 while the serial clock is input on SCL. START and STOP conditions are recognized as the beginning and end of a serial transfer.


## $1^{2}$ C Device Self Addressing

If this feature is used, 2 of the 16 key readback nodes can be left open or shorted for self-addressing. This is done with KEYA together with SEGG and SEGF. This two nodes cannot be used for key-readback in this case. After startup all devices have the predefined adress 0000000 . A single command for self-addressing will update all connected AS1115. This command has to be done after startup or everytime the AS1115 gets disconnected from the supply. The $I^{2} \mathrm{C}$ address definition must be done with fixed connection, since $1^{2} C$ detection is excluded from debounce time of key registers.

## I²C Device Address Byte

The address byte (see Figure 20) is the first byte received following the START condition from the master device.
Figure 20. $1^{2} \mathrm{C}$ Device Address Byte

| predefined address: | MSB | 6 | 5 | 4 | 3 | 2 | 1 | LSB |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | 0 | 0 | 0 | 0 | 0 | 0 | 0 | $\mathrm{R} / \overline{\mathrm{W}}$ |
|  | MSB | 6 | 5 | 4 | 3 | 2 | 1 | LSB |
| updated address: | 0 | 0 | 0 | 0 | 0 | A1 | AO | $\mathrm{R} / \overline{\mathrm{W}}$ |

- The default slave address is factory-set to 0000000.
- The two LSB bits of the address byte are the device select bits, A0 to A1, which can be set by the self-adress command after startup. A maximum of four devices with the same pre-set code can therefore be connected on the same bus at one time.
- The last bit of the address byte $(R / \bar{W})$ define the operation to be performed. When set to a 1 a read operation is selected; when set to a 0 a write operation is selected.
Following the START condition, the AS1115 monitors the $I^{2} \mathrm{C}$ bus, checking the device type identifier being transmitted. Upon receiving the address code, and the $R / \bar{W}$ bit, the slave device outputs an acknowledge signal on the SDA line.


## Command Byte

The AS1115 operation, (see Table 6) is determined by a command byte (see Figure 21 on page 11).
Figure 21. Command Byte

| MSB | 6 | 5 | 4 | 3 | 2 | 1 | LSB |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| D15 | D14 | D13 | D12 | D11 | D10 | D09 | D08 |

Figure 22. Command and Single Data Byte Received


Figure 23. Setting the Pointer to a Address Register to select a Data Register for a Read Operation


Figure 24. Reading nBytes from AS1115


## Initial Power-Up

On initial power-up, the AS1115 registers are reset to their default values, the display is blanked, and the device goes into shutdown mode. At this time, all registers should be programmed for normal operation.

Note: The default settings enable only scanning of one digit; the internal decoder is disabled and the Intensity Control Register (see page 17) is set to the minimum values.

## Shutdown Mode

The AS1115 devices feature a shutdown mode, where they consume only 200nA (typ) current. Shutdown mode is entered via a write to the Shutdown Register (see Table 7). During shutdown mode the Digit-Registers maintain their data.
Shutdown mode can either be used as a means to reduce power consumption or for generating a flashing display (repeatedly entering and leaving shutdown mode). For minimum supply current in shutdown mode, logic input should be at GND or VDD (CMOS logic level).
When entering or leaving shutdown mode, the Feature Register is reset to its default values (all 0s) when Shutdown Register bit D7 $($ page 13) $=0$.

Note: When Shutdown Register bit D7 = 1, the Feature Register is left unchanged when entering or leaving shutdown mode. If the AS1115 is used with an external clock, Shutdown Register bit D7 should be set to 1 when writing to the Shutdown Register.

## Digit- and Control-Registers

The AS1115 devices contain 8 Digit-Registers, 11 control-registers and 10 diagnostic-registers, which are listed in Table 6 . All registers are selected using a 8 -bit address word, and communication is done via the $I^{2} \mathrm{C}$ interface.

- Digit Registers - These registers are realized with an on-chip 64-bit memory. Each digit can be controlled directly without rewriting the whole register contents.
- Control Registers - These registers consist of decode mode, display intensity, number of scanned digits, shutdown, display test and features selection registers.

Table 6. Register Address Map

| $\stackrel{0}{2}$ | Register | Address |  |  |  |  |  |  | Page |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | D15:D13 | D12 | D11 | D10 | D9 | D8 | D7:D0 |  |
|  | Digit 0 | 000 | 0 | 0 | 0 | 0 | 1 | (see Table 9 on page 14, Table 10 on page 14 and Table 11 on page 15) | N/A |
|  | Digit 1 | 000 | 0 | 0 | 0 | 1 | 0 |  | N/A |
|  | Digit 2 | 000 | 0 | 0 | 0 | 1 | 1 |  | N/A |
|  | Digit 3 | 000 | 0 | 0 | 1 | 0 | 0 |  | N/A |
|  | Digit 4 | 000 | 0 | 0 | 1 | 0 | 1 |  | N/A |
|  | Digit 5 | 000 | 0 | 0 | 1 | 1 | 0 |  | N/A |
|  | Digit 6 | 000 | 0 | 0 | 1 | 1 | 1 |  | N/A |
|  | Digit 7 | 000 | 0 | 1 | 0 | 0 | 0 |  | N/A |
|  | Decode-Mode | 000 | 0 | 1 | 0 | 0 | 1 | (see Table 8 on page 13) | 13 |
|  | Global Intensity | 000 | 0 | 1 | 0 | 1 | 0 | (see Table 17 on page 17) | 17 |
|  | Scan Limit | 000 | 0 | 1 | 0 | 1 | 1 | (see Table 19 on page 17) | 17 |
|  | Shutdown | 000 | 0 | 1 | 1 | 0 | 0 | (see Table 7 on page 13) | 12 |
|  | Self-Adressing | 001 | 0 | 1 | 1 | 0 | 1 |  | N/A |
|  | Feature | 000 | 0 | 1 | 1 | 1 | 0 | (see Table 20 on page 18) | 18 |
|  | Display Test Mode | 000 | 0 | 1 | 1 | 1 | 1 | (see Table 14 on page 16) | 13 |
|  | DIG0:DIG1 Intensity | 000 | 1 | 0 | 0 | 0 | 0 | (see Table 18 on page 17) |  |
|  | DIG2:DIG3 Intensity | 000 | 1 | 0 | 0 | 0 | 1 | (see Table 18 on page 17) |  |
|  | DIG4:DIG5 Intensity | 000 | 1 | 0 | 0 | 1 | 0 | (see Table 18 on page 17) |  |
|  | DIG6:DIG7 Intensity | 000 | 1 | 0 | 0 | 1 | 1 | (see Table 18 on page 17) |  |

Table 6. Register Address Map

| $\stackrel{\otimes}{\mathrm{D}}$ | Register | Address |  |  |  |  |  |  | Page |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | D15:D13 | D12 | D11 | D10 | D9 | D8 | D7:D0 |  |
|  | Diagnostic Digit 0 | 000 | 1 | 0 | 1 | 0 | 0 |  | N/A |
|  | Diagnostic Digit 1 | 000 | 1 | 0 | 1 | 0 | 1 |  | N/A |
|  | Diagnostic Digit 2 | 000 | 1 | 0 | 1 | 1 | 0 |  | N/A |
| $\begin{aligned} & \frac{1}{0} \\ & \stackrel{0}{\omega} \\ & 0 \end{aligned}$ | Diagnostic Digit 3 | 000 | 1 | 0 | 1 | 1 | 1 |  | N/A |
|  | Diagnostic Digit 4 | 000 | 1 | 1 | 0 | 0 | 0 |  | N/A |
| $\begin{aligned} & \text { 두 } \\ & \hline . \pi \end{aligned}$ | Diagnostic Digit 5 | 000 | 1 | 1 | 0 | 0 | 1 |  | N/A |
|  | Diagnostic Digit 6 | 000 | 1 | 1 | 0 | 1 | 0 |  | N/A |
|  | Diagnostic Digit 7 | 000 | 1 | 1 | 0 | 1 | 1 |  | N/A |
|  | KEYA | 000 | 1 | 1 | 1 | 0 | 0 |  |  |
|  | KEYB | 000 | 1 | 1 | 1 | 0 | 1 |  |  |

The Shutdown Register controls AS1115 shutdown mode.
Table 7. Shutdown Register Format (Address $(H E X)=0 \times 0 C)$ )

| Mode | HEX Code | Register Data |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
| Shutdown Mode, Reset Feature Register to Default Settings | $0 \times 00$ | 0 | X | X | X | X | X | X | 0 |
| Shutdown Mode, Feature Register Unchanged | 0x80 | 1 | X | X | X | X | X | X | 0 |
| Normal Operation, Reset Feature Register to Default Settings | $0 \times 01$ | 0 | X | X | X | X | X | X | 1 |
| Normal Operation, Feature Register Unchanged | 0x81 | 1 | X | X | X | X | X | X | 1 |

## Decode Enable Register (0x09)

The Decode Enable Register sets the decode mode. BCD/HEX decoding (either BCD code - characters 0:9, E, H, L, P, and -, or HEX code - characters 0:9 and A:F) is selected by bit D2 (page 18) of the Feature Register. The Decode Enable Register is used to select the decode mode or no-decode for each digit. Each bit in the Decode Enable Register corresponds to its respective display digit (i.e., bit D0 corresponds to digit 0, bit D1 corresponds to digit 1 and so on). Table 9 lists some examples of the possible settings for the Decode Enable Register bits.

Note: A logic high enables decoding and a logic low bypasses the decoder altogether.
When decode mode is used, the decoder looks only at the lower-nibble (bits D3:D0) of the data in the Digit-Registers, disregarding bits D6:D4. Bit D7 sets the decimal point (SEG DP) independent of the decoder and is positive logic (bit D7 = 1 turns the decimal point on). Table 9 lists the code-B font; Table 10 lists the HEX font.
When no-decode mode is selected, data bits D7:D0 of the Digit-Registers correspond to the segment lines of the AS1115. Table 11 shows the 1:1 pairing of each data bit to the appropriate segment line.

Table 8. Decode Enable Register Format Examples

| Decode Mode | HEX Code | Register Data |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
| No decode for digits 7:0 | $0 \times 00$ | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Code-B/HEX decode for digit 0. No decode for digits 7:1 | $0 \times 01$ | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 1 |
| Code-B/HEX decode for digit 0:2. No decode for digits 7:3 | $0 \times 07$ | 0 | 0 | 0 | 0 | 0 | 1 | 1 | 1 |
| Code-B/HEX decode for digits 0:5. No decode for digits 7:6 | 0x3F | 0 | 0 | 1 | 1 | 1 | 1 | 1 | 1 |
| Code-B/HEX decode for digits $0,2,5$. No decode for digits 1, 3, 4, 6, 7 | 0x25 | 0 | 0 | 1 | 0 | 0 | 1 | 0 | 1 |

Figure 25. Standard 7-Segment LED Intensity Control and Inter-Digit Blanking


Table 9. Code-B Font

|  | Register Data |  |  |  |  |  | Character | Register Data |  |  |  |  |  | Character | Register Data |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| acter | D7 | D6:D4 | D3 | D2 | D1 | D0 |  | D7 | D6: D4 | D3 | D2 | D1 | D0 |  | D7 | D6:D4 | D3 | D2 | D1 | D0 |
| 5 |  | X | 0 | 0 | 0 | 0 |  |  | X | 0 | 1 | 1 | 0 | 18 |  | X | 1 | 1 | 0 | 0 |
| $1$ |  | X | 0 | 0 | 0 | 1 | 7 |  | X | 0 | 1 | 1 | 1 | 1 |  | X | 1 | 1 | 0 | 1 |
|  |  | X | 0 | 0 | 1 | 0 | $\square$ |  | X | 1 | 0 | 0 | 0 |  |  | X | 1 | 1 | 1 | 0 |
| I |  | X | 0 | 0 | 1 | 1 |  |  | X | 1 | 0 | 0 | 1 |  |  | X | 1 | 1 | 1 | 1 |
| $11$ |  | X | 0 | 1 | 0 | 0 | - |  | X | 1 | 0 | 1 | 0 |  | $1^{*}$ | X | X | X | X | X |
| $E$ |  | X | 0 | 1 | 0 | 1 | $E$ |  | X | 1 | 0 | 1 | 1 |  |  |  |  |  |  |  |

The decimal point can be enabled with every character by setting bit D7 $=1$.
Table 10. HEX Font

| Character | Register Data |  |  |  |  |  | Character | Register Data |  |  |  |  |  | Character | Register Data |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | D7 | D6:D4 | D3 | D2 | D1 | D0 |  | D7 | D6: D4 | D3 | D2 | D1 | D0 |  | D7 | D6:D4 | D3 | D2 | D1 | D0 |
| $1!$ |  | X | 0 | 0 | 0 | 0 |  |  | X | 0 | 1 | 1 | 0 |  |  | X | 1 | 1 | 0 | 0 |
| $1$ |  | X | 0 | 0 | 0 | 1 | $7$ |  | X | 0 | 1 | 1 | 1 | E1 |  | X | 1 | 1 | 0 | 1 |
|  |  | X | 0 | 0 | 1 | 0 | $\square$ |  | X | 1 | 0 | 0 | 0 | $E$ |  | X | 1 | 1 | 1 | 0 |
| 7 |  | X | 0 | 0 | 1 | 1 |  |  | X | 1 | 0 | 0 | 1 | E |  | X | 1 | 1 | 1 | 1 |
| $11$ |  | X | 0 | 1 | 0 | 0 | 5 |  | X | 1 | 0 | 1 | 0 | . | $1{ }^{*}$ | X | X | X | X | X |
| $E$ |  | X | 0 | 1 | 0 | 1 | 1 |  | X | 1 | 0 | 1 | 1 |  |  |  |  |  |  |  |

[^0]Table 11. No-Decode Mode Data Bits and Corresponding Segment Lines

|  | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Corresponding Segment Line | DP | A | B | C | D | E | F | G |

## $I^{2} \mathrm{C}$ Self Addressing

If this feature is used, 2 of the 16 key readback nodes can be left open or shorted for self-addressing. This is done with KEYA together with SEGG (A0) and SEGF (A1). This two nodes cannot be used for key-readback in this case. After startup all devices have the predefined adress 0000000. A single command for selfaddressing will update all connected AS1115. This command has to be done after startup or everytime the AS1115 gets disconnected from the supply. The $I^{2} \mathrm{C}$ address definition must be done with fixed connection, since $I^{2} \mathrm{C}$ detection is excluded from debounce time of key registers.

Note: A short writes a logical "0" whereas an open writes a logical " 1 " as address bit.
Table 12. Self Addressing Register (Address (HEX) $=0 \times 2 D)$ )

|  | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Factory-set IC address | X | X | X | X | X | X | X | 0 |
| User-set IC address | X | X | X | X | X | X | X | 1 |

## Keyscan Register

These two registers contain the result of the keyscan input of the 16 keys. To ensure proper results the data in these registers are updated only if the logic data scanned is stable for 20 ms (debounce time). A change of the data stored within these two registers is indicated by a logic low on the IRQ pin. The IRQ is high-impedance if a read operation on the key scan registers is started.

Table 13. LED Diagnostic Register Address

| Register HEX Address |  | Segment |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Key | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |  |  |
| $0 \times 1 C$ | KEYA | DP | A | B | C | D | E | F | G |  |  |
| $0 \times 1 D$ | KEYB |  |  |  |  |  |  |  |  |  |  |

Note: If ${ }^{2} \mathrm{C}$ self addressing is used segment G\&F of KEYA is used for the two LSB of the ${ }^{2} \mathrm{C}$ address. In this case these two nodes cannot be used as a key. Additionally the debounce time is disabled for these two bits. The data within the keyscan register is updated continuously during every cycle (1/10 of refresh rate). Therefore, to get a valid readback of keys it is recommended to read out the keyscan registers immediately after the IRQ is triggered. A short writes a logical " 0 " whereas an open writes a logical " 1 " as keyscan register bit.

Note: If the blink_en bit (bit D4 in the Feature Register $0 \times 0 \mathrm{E}$ ) is set to ' 1 ', the keyscan is not returning a valid value.

## Display-Test Mode

The AS1115 can detect open or shorted LEDs. Readout of either open LEDs or short LEDs is possible, as well as a OR relation of open and short.
Note: All settings of the digit- and control-registers are maintained.
Table 14. Testmode Register Summary

| D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| X | RSET_short | RSET_open | LED_global | LED_test | LED_open | LED_short | DISP_test |

Table 15. Testmode Register Bit Description (Address (HEX) = 0x0F))

| Addr: 0x0F |  | Address |  |  |
| :---: | :---: | :---: | :---: | :--- |
| Bit | Bit Name | Default | Access | D7:D0 |
| D0 | DISP_test | 0 | W | Optical display test. (Testmode for external visual test.) <br> 0: Normal operation; 1: Run display test (All digits are tested <br> independently from scan limit \& shutdown register.) |
| D1 | LED_short | 0 | W | Starts a test for shorted LEDs. (Can be set together with D2) <br> 0: Normal operation; 1: Activate testmode |
| D2 | LED_open | 0 | W | Starts a test for open LEDs. (Can be set together with D1) <br> 0: Normal operation; 1: Activate testmode |
| D3 | LED_test | 0 | R | Indicates an ongoing open/short LED test <br> 0: No ongoing LED test; 1: LED test in progress |
| D4 | LED_global | 0 | R | Indicates that the last open/short LED test has detected an error <br> 0: No error detected; 1: Error detected |
| D5 | RSET_open | 0 | R | Checks if external resistor RsET is open <br> 0: RSET correct; 1: RsET is open |
| D6 | RSET_short | 0 | R | Checks if external resistor RsET is shorted <br> 0: RSET correct; 1: RsET is shorted |
| D7 |  | 0 | - | Not used |

## LED Diagnostic Registers

These eight registers contain the result of the LED open/short test for the individual LED of each digit.
Table 16. LED Diagnostic Register Address

| Register |  |  |  |  | Seg | ment |  |  |  | Register |  |  |  |  | Seg | ment |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Address | Digit | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 | Address | Digit | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
| 0x14 | DIG0 | DP | A | B | C | D | E | F | G | 0x18 | DIG4 | DP | A | B | C | D | E | F | G |
| $0 \times 15$ | DIG1 |  |  |  |  |  |  |  |  | 0x19 | DIG5 |  |  |  |  |  |  |  |  |
| $0 \times 16$ | DIG2 |  |  |  |  |  |  |  |  | 0x1A | DIG6 |  |  |  |  |  |  |  |  |
| $0 \times 17$ | DIG3 |  |  |  |  |  |  |  |  | 0x1B | DIG7 |  |  |  |  |  |  |  |  |

Note: If one or more short occures in the LED array, detection of individual LED fault could become ambiguous.

## Intensity Control Register (0x0A)

The brightness of the display can be controlled by digital means using the Intensity Control Registers and by analog means using Rset (see Selecting RSET Resistor Value and Using External Drivers on page 19). The intensity can be controlled globally for all digits, or for each digit individually. The global intensity command will write intensity data to all four individual brightness registers, while the individual intesity command will only write to the associated individual intensity register.

Display brightness is controlled by an integrated pulse-width modulator which is controlled by the lower-nibble of the Intensity Control Register. The modulator scales the average segment-current in 16 steps from a maximum of 15/16 down to $1 / 16$ of the peak current set by RsET.

Table 17. Intensity Register Format

| Duty Cycle | HEX Code | Register Data |  |  |  | Duty Cycle | HEX Code | Register Data |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | MSB | D2 | D1 | LSB |  |  | MSB | D2 | D1 | LSB |
| 1/16 (min on) | 0xX0 | 0 | 0 | 0 | 0 | 9/16 | 0xX8 | 1 | 0 | 0 | 0 |
| 2/16 | 0xX1 | 0 | 0 | 0 | 1 | 10/16 | 0xX9 | 1 | 0 | 0 | 1 |
| 3/16 | 0xX2 | 0 | 0 | 1 | 0 | 11/16 | 0xXA | 1 | 0 | 1 | 0 |
| 4/16 | 0xX3 | 0 | 0 | 1 | 1 | 12/16 | 0xXB | 1 | 0 | 1 | 1 |
| 5/16 | 0xX4 | 0 | 1 | 0 | 0 | 13/16 | 0xXC | 1 | 1 | 0 | 0 |
| 6/16 | 0xX5 | 0 | 1 | 0 | 1 | 14/16 | 0xXD | 1 | 1 | 0 | 1 |
| 7/16 | 0xX6 | 0 | 1 | 1 | 0 | 15/16 | 0xXE | 1 | 1 | 1 | 0 |
| 8/16 | 0xX7 | 0 | 1 | 1 | 1 | 15/16 (max on) | 0xXF | 1 | 1 | 1 | 1 |

Table 18. Intensity Register Address

| Register HEX Address | Type | D7:D4 | Register Data |
| :---: | :---: | :---: | :---: |
|  | Global | X | Global Intensity |
| $0 \times 0$ A | Digit | Digit 1 Intensity | Digit 0 Intensity |
| $0 \times 10$ | Digit | Digit 3 Intensity | Digit 2 Intensity |
| $0 \times 11$ | Digit | Digit 5 Intensity | Digit 4 Intensity |
| $0 \times 12$ | Digit | Digit 7 Intensity | Digit 6 Intensity |
| $0 \times 13$ |  |  |  |

## Scan-Limit Register (0x0B)

The Scan-Limit Register controls which of the digits are to be displayed. When all 8 digits are to be displayed, the update frequency is typically 700 Hz . If the number of digits displayed is reduced, the update frequency is increased. The frequency can be calculated using $10 \times \mathrm{fOSC} /(\mathrm{N}+2)$, where N is the number of digits.

Note: To avoid differences in brightness this register should not be used to blank parts of the display (leading zeros).
Table 19. Scan-Limit Register Format (Address (HEX) $=0 \times 0 B)$ )

| Scan Limit | HEX <br> Code | Register Data |  |  |  | Scan Limit | HEX <br> Code | Register Data |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | D7:D3 | D2 | D1 | D0 |  |  | D7:D3 | D2 | D1 | D0 |
| Display digit 0 only | 0xX0 | X | 0 | 0 | 0 | Display digits 0:4 | 0xX4 | X | 1 | 0 | 0 |
| Display digits 0:1 | 0xX1 | X | 0 | 0 | 1 | Display digits 0:5 | 0xX5 | X | 1 | 0 | 1 |
| Display digits 0:2 | 0xX2 | X | 0 | 1 | 0 | Display digits 0:6 | 0xX6 | X | 1 | 1 | 0 |
| Display digits 0:3 | 0xX3 | X | 0 | 1 | 1 | Display digits 0:7 | 0xX7 | X | 1 | 1 | 1 |

## Feature Register (0x0E)

The Feature Register is used for enabling various features including switching the device into external clock mode, applying an external reset, selecting code-B or HEX decoding, enabling or disabling blinking, setting the blinking rate, and resetting the blink timing.

Note: At power-up the Feature Register is initialized to 0.
Table 20. Feature Register Summary

| D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| blink_ <br> start | sync | blink__ <br> freq_sel | blink_en | NU | decode_sel | reg_res | clk_en |

Table 21. Feature Register Bit Descriptions (Address $(H E X)=0 x X E)$

| Addr: 0xXE |  | Feature Register |  |  |  |
| :---: | :---: | :---: | :---: | :--- | :--- |
| Enables and disables various device features. |  |  |  |  |  |
| Bit | Bit Name | Default | Access | Bit Description |  |
| D0 | clk_en | 0 | R/W | External clock active. <br> $0=$ Internal oscillator is used for system clock. <br> $1=$ Pin CLK of the serial interface operates as system clock input. |  |
| D1 | reg_res | 0 | R/W | Resets all control registers except the Feature Register. <br> $0=$ Reset Disabled. Normal operation. <br> $1=$ All control registers are reset to default state (except the Feature <br> Register) identically after power-up. <br> Note: The Digit Registers maintain their data. |  |
| D2 | decode_sel | 0 | R/W | Selects display decoding for the selected digits (Table 8 on page 13). <br> $0=$ Enable Code-B decoding (see Table 9 on page 14). <br> $1=$ Enable HEX decoding (see Table 10 on page 14). |  |
| D3 | NU |  |  | R/W | Not used <br> Enables blinking. <br> $0=$ Disable blinking. 1 = Enable blinking. |
| D4 | blink_en | 0 | R/W | Sets blink with low frequency (with the internal oscillator enabled): <br> 0 = Blink period typically is 1 second (0.5s on, 0.5s off). <br> $1=$ Blink period is 2 seconds (1s on, 1s off). |  |
| blink_freq_sel | 0 | R/W | Synchronizes blinking on the rising edge of pin LD/CS. The multiplex <br> and blink timing counter is cleared on the rising edge of pin LD/CS. By <br> setting this bit in multiple devices, the blink timing can be synchronized <br> across all the devices. |  |  |
| D6 | sync | 0 | Start Blinking with display enabled phase. When bit D4 (blink_en) is set, <br> bit D7 determines how blinking starts. <br> 0 = Blinking starts with the display turned off. <br> $1=$ Blinking starts with the display turned on. |  |  |
| blink_start | 0 | R/W |  | R |  |

## 9 Typical Application

## Selecting Rseт Resistor Value and Using External Drivers

Brightness of the display segments is controlled via RSET. The current that flows into ISET defines the current that flows through the LEDs.
Segment current is about 200 times the current in ISET. Typical values for RSET for different segment currents, operating voltages, and LED voltage drop (VLED) are given in Table 22 \& Table 23. The maximum current the AS1115 can drive is 47 mA . If higher currents are needed, external drivers must be used, in which case it is no longer necessary that the devices drive high currents.

Note: The display brightness can also be logically controlled (see Intensity Control Register (0x0A) on page 16).
Table 22. Rset vs. Segment Current and LED Forward Voltage, VDD $=2.7 \mathrm{~V}$ \& 3.3V \& 3.6V

| ISEG (mA) |  | VLed |  |  | VLed |  |  |  | VLed |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | 1.5V | 2.0 V |  | 1.5V | 2.0 V | 2.5V |  | 1.5V | 2.0 V | 2.5 V | 3.0 V |
| 40 | $\begin{array}{l\|} \lambda \\ \lambda \\ \underset{\sim}{\prime \prime} \\ 0 \\ \lambda_{1} \end{array}$ | $5 \mathrm{k} \Omega$ | $4.4 \mathrm{k} \Omega$ | $\begin{aligned} & \hline \begin{array}{l} m \\ m \\ 0 \\ 11 \\ 0 \\ 0 \end{array} \end{aligned}$ | $6.7 \mathrm{k} \Omega$ | $6.4 \mathrm{k} \Omega$ | $5.7 \mathrm{k} \Omega$ | $\begin{array}{\|c} \hline \\ 0 \\ 1 \\ 11 \\ 0 \\ 0 \\ > \end{array}$ | 7.5k $\Omega$ | 7.2k $\Omega$ | 6.6k $\Omega$ | $5.5 \mathrm{k} \Omega$ |
| 30 |  | $6.9 \mathrm{k} \Omega$ | $5.9 \mathrm{k} \Omega$ |  | 9.1k $\Omega$ | 8.8k $\Omega$ | $8.1 \mathrm{k} \Omega$ |  | $10.18 \mathrm{k} \Omega$ | 9.8k $\Omega$ | $9.2 \mathrm{k} \Omega$ | $7.5 \mathrm{k} \Omega$ |
| 20 |  | $10.7 \mathrm{k} \Omega$ | $9.6 \mathrm{k} \Omega$ |  | $13.9 \mathrm{k} \Omega$ | $13.3 \mathrm{k} \Omega$ | $12.6 \mathrm{k} \Omega$ |  | $15.6 \mathrm{k} \Omega$ | $15 \mathrm{k} \Omega$ | $14.3 \mathrm{k} \Omega$ | $13 \mathrm{k} \Omega$ |
| 10 |  | $22.2 \mathrm{k} \Omega$ | $20.7 \mathrm{k} \Omega$ |  | $28.8 \mathrm{k} \Omega$ | $27.7 \mathrm{k} \Omega$ | 26k $\Omega$ |  | $31.9 \mathrm{k} \Omega$ | $31 \mathrm{k} \Omega$ | $29.5 \mathrm{k} \Omega$ | 27.3k $\Omega$ |

Table 23. Rset vs. Segment Current and LED Forward Voltage, VDD $=4.0 \mathrm{~V}$ \& 5.0V

| $\begin{aligned} & \text { ISEG } \\ & (\mathrm{mA}) \end{aligned}$ |  | VLed |  |  |  |  |  | VLed |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | 1.5V | 2.0 V | 2.5V | 3.0V | 3.5V |  | 1.5V | 2.0V | 2.5 V | 3.0 V | 3.5V | 4.0V |
| 40 | $\begin{array}{\|c\|} \hline> \\ 0 \\ \dot{子} \\ 11 \\ 0 \\ 0 \end{array}$ | $8.6 \mathrm{k} \Omega$ | $8.3 \mathrm{k} \Omega$ | $7.9 \mathrm{k} \Omega$ | $7.6 \mathrm{k} \Omega$ | $5.2 \mathrm{k} \Omega$ | 0 | $11.35 \mathrm{k} \Omega$ | $11.12 \mathrm{k} \Omega$ | $10.84 \mathrm{k} \Omega$ | $10.49 \mathrm{k} \Omega$ | $10.2 \mathrm{k} \Omega$ | $9.9 \mathrm{k} \Omega$ |
| 30 |  | $11.6 \mathrm{k} \Omega$ | $11.2 \mathrm{k} \Omega$ | $10.8 \mathrm{k} \Omega$ | $9.9 \mathrm{k} \Omega$ | $7.8 \mathrm{k} \Omega$ |  | $15.4 \mathrm{k} \Omega$ | $15.1 \mathrm{k} \Omega$ | $14.7 \mathrm{k} \Omega$ | $14.4 \mathrm{k} \Omega$ | $13.6 \mathrm{k} \Omega$ | $13.1 \mathrm{k} \Omega$ |
| 20 |  | $17.7 \mathrm{k} \Omega$ | $17.3 \mathrm{k} \Omega$ | $16.6 \mathrm{k} \Omega$ | $15.6 \mathrm{k} \Omega$ | $13.6 \mathrm{k} \Omega$ |  | $23.6 \mathrm{k} \Omega$ | $23.1 \mathrm{k} \Omega$ | $22.6 \mathrm{k} \Omega$ | $22 \mathrm{k} \Omega$ | $21.1 \mathrm{k} \Omega$ | $20.2 \mathrm{k} \Omega$ |
| 10 |  | $36.89 \mathrm{k} \Omega$ | $35.7 \mathrm{k} \Omega$ | $34.5 \mathrm{k} \Omega$ | $32.5 \mathrm{k} \Omega$ | $29.1 \mathrm{k} \Omega$ |  | $48.9 \mathrm{k} \Omega$ | $47.8 \mathrm{k} \Omega$ | $46.9 \mathrm{k} \Omega$ | $45.4 \mathrm{k} \Omega$ | $43.8 \mathrm{k} \Omega$ | $42 \mathrm{k} \Omega$ |

## Calculating Power Dissipation

The upper limit for power dissipation (PD) for the AS1115 is determined from the following equation:

$$
\begin{equation*}
P D=(V D D \times 5 m A)+(V D D-V L E D)(D U T Y \times I \text { SEG } \times N) \tag{EQ1}
\end{equation*}
$$

## Where:

VDD is the supply voltage.
DUTY is the duty cycle set by intensity register (page 17).
$N$ is the number of segments driven (worst case is 8)
VLED is the LED forward voltage
ISEG = segment current set by RSET
Dissipation Example:

$$
\begin{gather*}
\text { ISEG }=40 \mathrm{~mA}, N=8, D U T Y=15 / 16, V \text { LED }=2.2 \mathrm{~V} \text { at } 40 \mathrm{~mA}, \mathrm{VDD}=5 \mathrm{~V}  \tag{EQ2}\\
P D=5 \mathrm{~V}(5 \mathrm{~mA})+(5 \mathrm{~V}-2.2 \mathrm{~V})(15 / 16 \times 40 \mathrm{~mA} \times 8)=0.865 \mathrm{~W} \tag{EQ3}
\end{gather*}
$$

Thus, for a $\operatorname{TQFN}(4 \times 4)-24$ package $\Theta J A=+30.5^{\circ} \mathrm{C} / \mathrm{W}$, the maximum allowed TAMB is given by:

$$
\begin{equation*}
T_{J, M A X}=\text { TAMB }+P D \times \Theta_{J A}=150^{\circ} \mathrm{C}=\text { TAMB }^{2}+0.865 \mathrm{~W} \times 30.5^{\circ} \mathrm{C} / \mathrm{W} \tag{EQ4}
\end{equation*}
$$

In this example the maximum ambient temperature must stay below $123.61^{\circ} \mathrm{C}$.

## 8x8 Dot Matrix Mode

The application example in Figure 26 shows the AS1115 in the $8 \times 8$ LED dot matrix mode.
The LED columns have common cathodes and are connected to the DIG0:7 outputs. The rows are connected to the segment drivers. Each of the 64 LEDs can be addressed separately. The columns are selected via the digits as listed in Table 6 on page 12.

The Decode Enable Register (see page 13) must be set to ‘00000000’ as described in Table 8 on page 13. Single LEDs in a column can be addressed as described in Table 11 on page 15, where bit D0 corresponds to segment G and bit D7 corresponds to segment DP.

Figure 26. Application Example as LED Dot Matrix Driver


## Keyscan

The key readback of the AS1115 can be used either for push buttons as well as switches. If only a single key is pressed (shorted) at a time no additional diodes are required. If a detection of multiple simultaneous keystrokes is required diodes within the keypath, as shown in Figure 27, are required. Pressing multiple keys without the diodes would result in ambiguous results. Since KEYA and KEYB have independent inputs only keys on the same path are affected.

Figure 27. Keyscan Configuration


Diodes are optional and only required if multiple keystrokes must be detected simultaneously.
If $I^{2} C$ Self-Adressing is used these two keys cannot be used for readback and must be either hard wired opened or shorted.
A short writes a logical "0" whereas an open writes a logical " 1 " as address bit.

Note: If the blink_en bit (bit D4 in the Feature Register $0 \times 0 \mathrm{E}$ ) is set to ' 1 ', the keyscan is not returning a valid value.

## Supply Bypassing and Wiring

In order to achieve optimal performance the AS1115 should be placed very close to the LED display to minimize effects of electromagnetic interference and wiring inductance.

Furthermore, it is recommended to connect a $10 \mu \mathrm{~F}$ and a $0.1 \mu \mathrm{~F}$ ceramic capacitor between pins VDD and GND to avoid power supply ripple (see Figure 26).

## 10 Package Drawings and Markings

The AS1115 is available in the QSOP-24 package.
Figure 28. QSOP-24 Package


| Symbol | Min | Max |
| :---: | :---: | :---: |
| A | 1.35 | 1.75 |
| A1 | 0.10 | 0.25 |
| A2 | 1.37 | 1.57 |
| b | 0.20 | 0.30 |
| C | 0.19 | 0.25 |
| D | 8.55 | 8.74 |
| E | 5.79 | 6.20 |
| E1 | 3.81 | 3.99 |
| e | 0.635 |  |
| BSC |  |  |
| h | 0.22 | 0.49 |
| L | 0.40 | 1.27 |
| $\theta$ | $0^{\circ}$ | $8^{\circ}$ |

Figure 29. TQFN(4x4)-24 Package


| Symbol | Min | Typ | Max |
| :---: | :---: | :---: | :---: |
| A | 0.50 | 0.55 | 0.60 |
| A1 | 0.00 |  | 0.05 |
| A3 |  | $0.152 R E F$ |  |
| b | 0.18 | 0.23 | 0.28 |
| D |  | 4.00 BSC |  |
| E |  | 4.00 BSC |  |
| D2 | 2.70 | 2.80 | 2.90 |
| E2 | 2.70 | 2.80 | 2.90 |


| Symbol | Min | Typ | Max |
| :---: | :---: | :---: | :---: |
| e |  | 0.50 BSC |  |
| L | 0.30 | 0.35 | 0.40 |
| L1 | 0.00 |  | 0.10 |
| aaa |  | 0.10 |  |
| bbb |  | 0.10 |  |
| ccc |  | 0.10 |  |
| ddd |  | 0.05 |  |
| eee |  | 0.08 |  |

Notes:Unilateral coplanarity zone applies to the exposed heat sink slug as well as the terminals.

1. All dimensions are in millimeters; angles in degrees.
2. Dimension b applies to metallized terminal and is measured between 0.25 mm and 0.30 mm from terminal tip. Dimension L1 represents terminal full back from package edge up to 0.1 mm is acceptable.
3. Coplanarity applies to the exposed heat slug as well as the terminal.
4. Radius on terminal is optional.

## 11 Ordering Infomation

The devices are available as the standard products shown in Table 24.
Table 24. Ordering Information

| Ordering Code | Marking | Description | Delivery Form | Package |
| :---: | :---: | :---: | :---: | :---: |
| AS1115-BSST | AS1115 | 64 LEDs, ${ }^{2} \mathrm{C}$ Interfaced LED Driver |  |  |
|  | with Keyscan | Tape and Reel | QSOP-24 |  |
|  | AS1115-BQFT | ASSD | Tape and Reel | TQFN $(4 \times 4)-24$ |

Note: All products are RoHS compliant and Pb -free.
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[^0]:    The decimal point can be enabled with every character by setting bit D7 $=1$.

