MITSUBISHI SEMICONDUCTOR < Application Specific Intelligent Power Module>



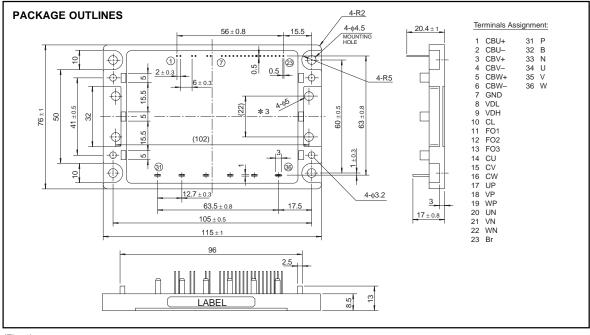
PS12018-A

FLAT-BASE TYPE



APPLICATION

Acoustic noise-less 3.7kW/AC400V Class 3 Phase inverter and other motor control applications.

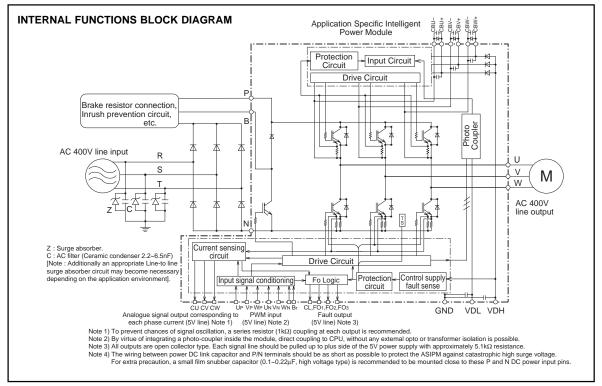


(Fig. 1)





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(Fig. 2)

MAXIMUM RATINGS (Tj = 25°C) INVERTER PART (Including Brake Part)

Symbol	Item	Condition	Ratings	Unit
Vcc	Supply voltage	Applied between P-N	900	V
VCC(surge)	Supply voltage (surge)	Applied between P-N, Surge-value	1000	V
VP or VN	Each output IGBT collector-emitter static voltage	1200	V	
VP(S) or VN(S)	Each output IGBT collector-emitter surge voltage Applied between P-U, V, W, Br or U, V, W, Br-N		1200	V
±lc(±lcp)	Each output IGBT collector current	±25 (±50)	A	
lc(lcp)	Brake IGBT collector current		10 (20)	A
IF(IFP)	Brake diode anode current	Note : "()" means IC peak value	10 (20)	A

CONTROL PART

Symbol	Item	Item Condition		Unit
Vdh, Vdb	Supply voltage	Applied between VDH-GND, CBU+-CBU–, CBV+-CBV–, CBW+-CBW–	20	V
Vdl	Supply voltage	Applied between VDL-GND	7	V
VCIN	Input signal voltage	Applied between UP \cdot VP \cdot WP \cdot UN \cdot VN \cdot WN \cdot Br-GND	-0.5 ~ Vdl+0.5	V
Vfo	Fault output supply voltage	Applied between F01 · F02 · F03-GND	-0.5 ~ 7	V
IFO	Fault output current	Sink current of F01 · F02 · F03	15	mA
VCL	Current-limit warning output voltage	Applied between CL-GND	-0.5 ~ 7	V
ICL	CL output current	Sink current of CL	15	mA
Ico	Analogue-current-signal output current	Sink current of CU · CV · CW	±1	mA





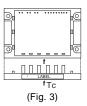
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TOTAL SYSTEM

Symbol	Item	Condition	Ratings	Unit
Tj	Junction temperature	n temperature (Note 2)		°C
Tstg	Storage temperature	—	-40 ~ +125	°C
Тс	Module case operating temperature	ule case operating temperature (Fig. 3)		°C
Viso	Isolation voltage	60 Hz sinusoidal AC for 1 minute, between all terminals and base plate.	2500	Vrms
_	Mounting torque	Mounting screw: M4.0	0.98 ~ 1.47	N∙m

Note 2) : The item defines the maximum junction temperature for the power elements (IGBT/Diode) of the ASIPM to ensure safe operation. However, these power elements can endure instantaneous junction temperature as high as 150°C. To make use of this additional temperature allowance, a detailed study of the exact application conditions is required and, accordingly, necessary information is to be provided before use.

CASE TEMPERATURE MEASUREMENT POINT (3mm from the base surface)



THERMAL RESISTANCE

Symbol	ltom	Condition	Ratings			Linit
	Item	Condition	Min.	Тур.	Max.	Unit
Rth(jc)Q	Junction to case Thermal Resistance	Inverter IGBT (1/6)	—	—	1.6	°C/W
Rth(jc)F		Inverter FWDi (1/6)	—		3.0	°C/W
Rth(jc)QB		Brake IGBT	—	—	2.9	°C/W
Rth(jc)FB		Brake FWDi	—	_	5.5	°C/W
Rth(c-f)	Contact Thermal Resistance	Case to fin, thermal grease applied (1 Module)	—	—	0.031	°C/W

ELECTRICAL CHARACTERISTICS (Tj = 25°C, VDH = 15V, VDB = 15V, VDL = 5V unless otherwise noted)

Cumbal	lite an	Condition		Ratings		
Symbol	Item	Condition	Min.	Тур.	Max.	Unit
VCE(sat)	Collector-emitter saturation voltage	VDL = 5V, VDH = VDB = 15V Input = ON, Tj = 25°C, Ic = 25A		_	3.6	V
VEC	FWDi forward voltage	Tj = 25° C, Ic = -25 A, Input = OFF	—	—	3.5	V
VCE(sat)Br	Brake IGBT Collector-emitter saturation voltage	VDL = 5V, VDH = 15V Input = ON, Tj = 25°C, Ic = 10A		_	3.6	V
VFBr	Brake diode forward voltage $Tj = 25^{\circ}C$, IF = 10A, Input = OFF		_	—	3.5	V
ton		1/2 Bridge inductive, Input = ON	0.40	1.4	2.5	μs
tc(on)	Switching times	Vcc = 600V, lc = 25A, Tj = 125°C	—	0.60	1.5	μs
toff		Vdl = 5V, Vdh = 15V, Vdb = 15V	_	2.2	4.0	μs
tc(off)		Note : ton, toff include delay time of the internal control	—	0.9	1.6	μs
trr	FWD reverse recovery time	circuit.	—	0.2	—	μs
	Short circuit endurance (Output, Arm, and Load, Short Circuit Modes)	Vcc \leq 800V, Input = ON (One-Shot) Tj = 125°C start 13.5V \leq VDH = VDB = \leq 16.5V, VDL = 5V	No destruction Fo output by protection operation No destruction No protecting operation No Fo output			
	Switching SOA	Vcc \leq 800V, Tj \leq 125°C, lc $<$ loL(CL) operation level, Input = ON,				
		$13.5V \le VDH = VDB = \le 16.5V, VDL = 5V$				
IDH	VDH Circuit Current	VDL = 5V, VDH = 15V, VCIN = 5V	-	· —	150	mA
Idl	VDL Circuit Current	VDL = 5V, VDH = 15V, VCIN = 5V	—	—	50	mA
Vth(on)	Input on threshold voltage		0.8	1.4	2.0	V
Vth(off)	Input off threshold voltage		2.5	3.0	4.0	V
Ri	Input pull-up resistor	Integrated between input terminal-VDH	-	150	-	kΩ





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Symbol	Item		Condition		Ratings			Unit		
Symbol	ILE	2111			Condition		Min.	Тур.	Max.	
fpwm	PWM input free	quency		$TC \le 100^{\circ}C, Tj \le 125^{\circ}C$				—	15	kHz
txx	Allowable input	on-pul	se width	Vdh = 15V, Vdl =	$VDH = 15V, VDL = 5V, TC = -20^{\circ}C \sim +100^{\circ}C$ Note 3)		1	—	500	μs
tdead	Allowable input signal dead time for blocking arm shoot-through		Relates to corres Tc = -20°C ~ +1	ponding inputs (Except brake 00°C	e part)	4.0	_	_	μs	
tint	Input inter-lock sensing		Relates to corres	ponding inputs (Except brake	e part)		65	100	ns	
Vco				Ic = 0A	VDH = 15V		1.87	2.27	2.57	V
VC+(200%)	Analogue signa	al linea	rity with	Ic = IOP(200%)	VDL = 5V		0.77	1.17	1.47	V
VC-(200%)	output current		Ic = -IOP(200%)	Tc = −20 ~ 100°C	(Fig.4)	2.97	3.37	3.67	V	
∆Vco	Offset change are	ea vs ten	nperature	VDH = 15V, VDL =	5V, Tc = -20 ~ 100°C		_	15	_	mV
VC+	A	h		Ic > IOP(200%), VE	он = 15V,		_	_	0.7	V
Vc-	Analogue signal output voltage limit		VDL = 5V		(Fig. 4)	4.0	_	_	V	
∆Vc(200%)	Analogue signal overall linear		all linear	VCO-VC±(200%)			_	1.1	_	V
rCH	Analogue signal data hold accuracy		Correspond to m Ic = IOP(200%)	ax. 500µs data hold period or	nly, (Fig. 5)	-5	_	5	%	
td(read)	Analogue signal reading time		ng time	After input signal	trigger point	(Fig. 8)	_	3	_	μs
ICL(H)	Signal output	Signal output our Idle				-	_	1	μΑ	
ICL(L)	rent of CL ope	ration	Active	Open collector onput		_	1	_	mA	
±IOL	CL warning op	eration	level	Vdl = 5V, Vdh =	15V, Tc = -20 ~ 100°C	(Note 4)	26.0	32.5	39.0	A
SC	Short circuit cu	irrent tr	ip level	Tj = 25°C	(Fig. 7)	, (Note 5)	43.0	55.0	_	Α
ОТ	Over tenperature	Trip	level				100	110	120	°C
OTr	protection	Res	et level	VDL = 5V, VDH = 1	15V		_	90	_	°C
UVdb		Trip	level				10.0	11.0	12.0	V
UVDBr		Res	et level				10.5	11.5	12.5	V
UVdн	Supply circuit	Trip	level		0000		11.05	12.00	12.75	V
UVDHr	under and over voltage	Res	et level	$T_{C} = -20^{\circ}C \sim +10^{\circ}$	UU~C		11.55	12.50	13.25	V
OVDH	protection Trip	level	Tj ≤ 125°C			18.00	19.20	20.15	V	
OVDHr		Reset level		1			16.50	17.50	18.65	V
tdv		Filte	r time				_	10	—	μs
IFO(H)			Idle	0			_	—	1	μΑ
IFO(L)	Fault output cu	urent	Active	Open collector of	pen collector output		_	1	—	mA

ELECTRICAL CHARACTERISTICS (Tj = 25° C, VDH = 15V, VDB = 15V, VDL = 5V unless otherwise noted)

(Note 3) : (a) Allowable minimum input on-pulse width : This item applies to P-side circuit only.

(b) Allowable maximum input on-pulse width : This item applies to both P-side and N-side circuits excluding the brake circuit.

(Note4): CL output : The "current limit warning (CL) operation circuit outputs warning signal whenever the arm current exceeds this limit. The circuit is reset automatically by the next input signal and thus, it operates on a pulse-by-pulse scheme.

(Note5): The short circuit protection works instantaneously when a high short circuit current flows through an internal IGBT rising up momentarily. The protection function is, thus meant primarily to protect the ASIPM against short circuit distraction. Therefore, this function is not recommended to be used for any system load current regulation or any over load control as this might, cause a failure due to excessive temperature rise. Instead, the analogue current output feature or the over load warning feature (CL) should be appropriately used for such current regulation or over load control operation. In other words, the PWM signals to the ASIPM should be shut down, in principle, and not to be restarted before the junction temperature would recover to normal, as soon as a fault is feed back from its Fo1 pin of the ASIPM indicating a short circuit situation.

RECOMMENDED CONDITIONS

Symbol	No	Condition		Ratings		
	Item Condition		Min.	Тур.	Max.	Unit
Vcc	Supply voltage	Applied between P-N	—	600	800	V
Vdh, Vdb	Control supply voltage	Applied between VDH-GND, CBU+-CBU-, CBV+-CBV-, CBW+-CBW-		15.0	16.5	v
Vdl	Control supply voltage	Applied between VDL-GND		5.0	5.2	V
$\Delta VDH, \Delta VDB, \Delta VDL$	^{ΔVDB,} Supply voltage ripple		-1	_	+1	V/µs
VCIN(on)	Input ON voltage		-	—	0.3	V
VCIN(off)	Input OFF voltage		4.8	—	_	V
fpwm	PWM Input frequency	Using application circuit	2	10	15	kHz
tdead	Arm shoot-through blocking time Using application circuit		4.0	_	_	μs





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Fig. 4 OUTPUT CURRENT ANALOGUE SIGNALING LINEARITY

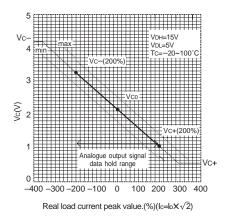
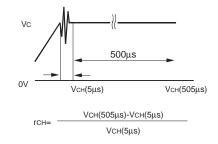
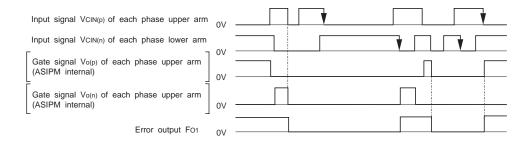


Fig. 5 OUTPUT CURRENT ANALOGUE SIGNALING "DATA HOLD" DEFINITION



Note ; Ringing happens around the point where the signal output voltage changes state from "analogue" to "data hold" due to test circuit arrangement and instrumentational trouble. Therefore, the rate of change is measured at a 5 μs delayed point.

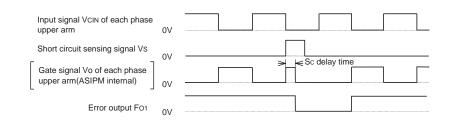
Fig. 6 INPUT INTERLOCK OPERATION TIMING CHART



Note : Input interlock protection circuit ; It is operated when the input signals for any upper-arm / lower-arm pair of a phase are simultaneously in "LOW" level.

By this interlocking, both upper and lower IGBTs of this mal-triggered phase are cut off, and "Fo" signal is outputted. After an "input interlock" operation the circuit is latched. The "Fo" is reset by the high-to-low going edge of either an upper-leg, or a lower-leg input, whichever comes in later.

Fig. 7 TIMING CHART AND SHORT CIRCUIT PROTECTION OPERATION



Note : Short circuit protection operation. The protection operates with "Fo" flag and reset on a pulse-by-pulse scheme. The protection by gate shutdown is given only to the IGBT that senses an overload (excluding the IGBT for the "Brake").





PS12018-A FLAT-BASE TYPE

INSULATED TYPE

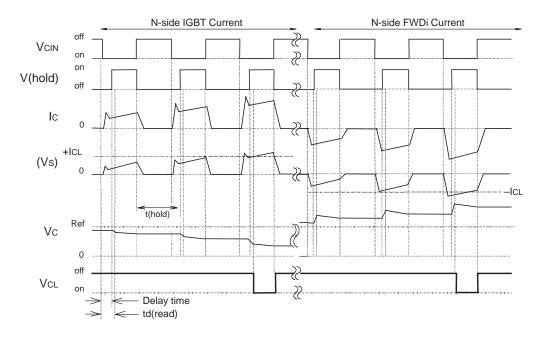


Fig. 8 INVERTER OUTPUT ANALOGUE CURRENT SENSING AND SIGNALING TIMING CHART.

Fig. 9 START-UP SEQUENCE

Normally at start-up, Fo and CL output signals will be pulled-up High to VDL voltage (OFF level); however, Fo1 output may fall to Low (ON) level at the instant of the first ON input pulse to an N-Side IGBT. This can happen particularly when the boot-strap capacitor is of large size. Fo1 resetting sequence (together with the boot-strap charging sequence) is explained in the following graph

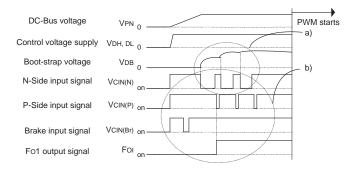
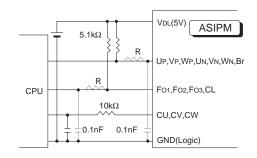


Fig. 10 RECOMMENDED I/O INTERFACE CIRCUIT



a) Boot-strap charging scheme :

Apply a train of short ON pulses at all N-IGBT input pins for adequate charging (pulse width = approx. 20µs number of pulses =10 ~ 500 depending on the boot-strap capacitor size)

b) Fo1 resetting sequence:

Apply ON signals to the following input pins : Br \rightarrow Un/Vn/Wn \rightarrow Up/Vp/Wp in that order.

