TOSHIBA Bi-CMOS Integrated Circuit Silicon Monolithic

## TB6556F/FG

3-Phase Full-Wave Sine-Wave PWM Brushless Motor Controller

## Features

- Sine-wave PWM control
- Built-in triangular-wave generator (carrier cycle $=\mathrm{f}_{\text {osc }} / 252(\mathrm{~Hz})$ )
- Built-in lead angle control function ( $0^{\circ}$ to $58^{\circ}$ in 32 steps) External setting/automatic internal setting
- Built-in dead time function (setting $2.6 \mu \mathrm{~s}$ or $3.8 \mu \mathrm{~s}$ )
- Overcurrent protection signal input pin
- Built-in regulator $\left(\mathrm{V}_{\text {refout }}=5 \mathrm{~V}(\right.$ typ. $\left.), 30 \mathrm{~mA}(\max )\right)$
- Operating supply voltage range: $\mathrm{VCC}=6 \mathrm{~V}$ to 10 V


Weight: 0.33 g (typ.)

TB6556FG:
TB6556FG is a Pb -free product.
The following conditions apply to solderability:
*Solderability

1. Use of $\mathrm{Sn}-37 \mathrm{~Pb}$ solder bath
${ }^{*}$ solder bath temperature $=230^{\circ} \mathrm{C}$
*dipping time $=5$ seconds
*number of times $=$ once
*use of R-type flux
2. Use of $\mathrm{Sn}-3.0 \mathrm{Ag}-0.5 \mathrm{Cu}$ solder bath
${ }^{*}$ solder bath temperature $=245^{\circ} \mathrm{C}$
*dipping time $=5$ seconds
*the number of times = once
*use of R-type flux

Block Diagram


## Pin Description

| Pin No. | Symbol | Description | Remarks |
| :---: | :---: | :---: | :---: |
| 21 | HU | Positional signal input pin U | When positional signal is HHH or LLL, gate block protection operates. With built-in pull-up resistor, built-in digital filter ( $\simeq 500 \mathrm{~ns}$ ) |
| 20 | HV | Positional signal input pin V |  |
| 19 | HW | Positional signal input pin W |  |
| 18 | CW/CCW | Rotation direction signal input pin | L: Forward H: Reverse |
| 11 | RES | Reset-signal-input pin | L : Reset (output is non-active) operation/halt operation, also used for gate protection, built-in pull-up resistor |
| 2 | $\mathrm{V}_{\mathrm{e}}$ | Voltage command signal | With built-in pull-down resistor |
| 24 | $\mathrm{G}_{\text {in }}$ | Gain setting | $\mathrm{I}_{\mathrm{dc}}$ signal level at a gain that optimizes the LA |
| 25 | Gout |  |  |
| 26 | PH | Peak hold | Connect the peak-hold capacitor and discharge resistor to GND, parallel to each other |
| 27 | LPF | RC low-pass filter | Connect the low-pass filter capacitor (built-in $100 \mathrm{k} \Omega$ resistor) |
| 28 | LA | Lead angle setting signal input pin | Sets $0^{\circ}$ to $58^{\circ}$ in 32 steps |
| 29 | LL | Lower limit for LA | Set lower limit for LA ( $\mathrm{LL}=0 \mathrm{~V}$ to 5.0 V) |
| 30 | UL | Upper limit for LA | Set upper limit for LA ( $\mathrm{UL}=0 \mathrm{~V}$ to 5.0 V ) |
| 12 | OS | Inputs output logic select signal | L: Active LOW <br> H: Active HIGH |
| 3 | $I_{\text {dc }}$ | Inputs overcurrent protection signal | Inputs DC link current. <br> Reference voltage: 0.5 V <br> With built-in filter $(\simeq 1 \mu \mathrm{~s})$, built-in digital filter $(\simeq 1 \mu \mathrm{~s})$ |
| 14 | $\mathrm{X}_{\text {in }}$ | Inputs clock signal | With built-in feedback resistor |
| 15 | $\mathrm{X}_{\text {out }}$ | Outputs clock signal |  |
| 23 | $V_{\text {refout }}$ | Outputs reference voltage signal | 5 V (typ.), 30 mA (max) |
| 17 | FG | FG signal output pin | Outputs 3 PPR of positional signal |
| 16 | REV | Reverse rotation detection signal | Detects reverse rotation. |
| 9 | U | Outputs turn-on signal | Select active HIGH or active LOW using the output logic select pin. |
| 8 | V | Outputs turn-on signal |  |
| 7 | W | Outputs turn-on signal |  |
| 6 | X | Outputs turn-on signal |  |
| 5 | Y | Outputs turn-on signal |  |
| 4 | Z | Outputs turn-on signal |  |
| 1 | $\mathrm{V}_{\mathrm{CC}}$ | Power supply voltage pin | $\mathrm{V}_{\mathrm{CC}}=6$ to 10 V |
| 10 | Td | Inputs setting dead time | L: $3.8 \mu \mathrm{~s}, \mathrm{H}$ or OPEN: $1.9 \mu \mathrm{~s}$ |
| 22 | SS | $120^{\circ} 180^{\circ}$ select signal | $\mathrm{L}: 120^{\circ}$ turn-on mode, H or OPEN: $180^{\circ}$ turn-on mode |
| 13 | GND | Ground pin | - |

## Input/Output Equivalent Circuits

| Pin Description | Symbol | Input/Output Signal | Input/Output Internal Circuit |
| :---: | :---: | :---: | :---: |
| Positional signal input pin $U$ Positional signal input pin V Positional signal input pin W | HU <br> HV <br> HW | Digital <br> With Schmitt trigger <br> Hysteresis 300 mV (typ.) <br> Digital filter: 500 ns (typ.) <br> L: 0.8 V (max) <br> H: $\mathrm{V}_{\text {refout }}-1 \mathrm{~V}(\mathrm{~min})$ |  |
| Forward/reverse switching input pin <br> L: Forward (CW) <br> H: Reverse (CCW) | CW/CCW | Digital $\begin{aligned} & \text { L: } 0.8 \mathrm{~V}(\max ) \\ & \text { H: } \mathrm{V}_{\text {refout }}-1 \mathrm{~V} \text { (min) } \end{aligned}$ |  |
| Reset input <br> L: Stops operation (reset) <br> H: Operates | RES | Digital <br> L: 0.8 V (max) <br> H : $\mathrm{V}_{\text {refout }}-1 \mathrm{~V}$ (min) |  |
| $120^{\circ} / 180^{\circ}$ select signal <br> L: $120^{\circ}$ turn-on mode H: $180^{\circ}$ turn-on mode (OPEN) | SS | Digital <br> With Schmitt trigger Hysteresis: 300 mV (typ.) <br> L: 0.8 V (max) <br> H : $\mathrm{V}_{\text {refout }}-1 \mathrm{~V}$ (min) |  |
| Voltage command signal <br> $1.0 \mathrm{~V}<\mathrm{Ve} \leq 2.1 \mathrm{~V}$ <br> Refresh operation (X, Y, Z pins: ON duty of 8\%) | $\mathrm{V}_{\mathrm{e}}$ | Analog <br> Input voltage range 0 to 5.4 V Input voltage of 5.4 V or higher is clipped to 5.4 V . |  |


| Pin Description | Symbol | Input/Output Signal | Input/Output Internal Circuit |
| :---: | :---: | :---: | :---: |
| Lead angle setting signal input pin <br> 0 V : $0^{\circ}$ <br> 5 V : $58^{\circ}$ <br> (5-bit AD) | LA | When LA is fixed externally, connect LL to GND and UL to $V_{\text {refout, }}$ and then input the setting voltage to the LA pin. <br> Input voltage range: 0 V to 5.0 V ( $\mathrm{V}_{\text {refout }}$ ) <br> Input voltage of $\mathrm{V}_{\text {refout }}$ or higher is clipped to $\mathrm{V}_{\text {refout }}$. <br> When LA is fixed automatically, open the LA pin. In this state, the LA pin is used only for confirmation of LA width. |  |
| Gain setting signal input (LA setting) | $\mathrm{G}_{\text {in }}$ $G_{\text {out }}$ | Non-inverted amplifier 25 dB (max) <br> Gout output voltage <br> LOW: GND <br> HIGH: VCC -1.7 V |  |
| Peak hold (LA setting) | PH | Connect the peak-hold capacitor and discharge resistor to GND, parallel to each other. <br> $100 \mathrm{k} \Omega / 0.1 \mu \mathrm{~F}$ recommended |  |
| Low-pass filter (LA setting) | LPF | Connect the low-pass filter capacitor (built-in $100 \mathrm{k} \Omega$ resistor) <br> $0.1 \mu \mathrm{~F}$ recommended |  |
| Lower limit for LA | LL | Clip lower limit for LA <br> $\mathrm{LL}=0 \mathrm{~V}$ to 5.0 V <br> When $\mathrm{LL}>\mathrm{UL}$, LA is fixed at LL value. |  |


| Pin Description | Symbol | Input/Output Signal | Input/Output Internal Circuit |
| :---: | :---: | :---: | :---: |
| Upper limit for LA | UL | Clip upper limit for LA <br> $\mathrm{UL}=0 \mathrm{~V}$ to 5.0 V <br> When LL > UL, LA is fixed at LL value. |  |
| Setting dead time input pin $\begin{aligned} & \mathrm{L}: 3.8 \mu \mathrm{~s} \\ & \mathrm{H} \text { or OPEN: } 1.9 \mu \mathrm{~s} \end{aligned}$ | Td | Digital $\begin{aligned} & \text { L: } 0.8 \mathrm{~V} \text { (max) } \\ & \mathrm{H}: \mathrm{V}_{\text {refout }}-1 \mathrm{~V}(\min ) \end{aligned}$ |  |
| Output logic select signal input pin <br> L: Active LOW <br> H: Active HIGH | OS | Digital $\begin{aligned} & \text { L: } 0.8 \mathrm{~V} \text { (max) } \\ & \mathrm{H}: \mathrm{V}_{\text {refout }}-1 \mathrm{~V}(\min ) \end{aligned}$ |  |
| Overcurrent protection signal input pin | $l_{\text {dc }}$ | Analog <br> Digital filter: $1 \mu \mathrm{~s}$ (typ.) <br> Gate protected at 0.5 V or higher (released at carrier cycle) |  |
| Clock signal input pin | $\mathrm{x}_{\text {in }}$ | Operating range <br> 2 MHz to 8 MHz (crystal oscillation) |  |
| Clock signal output pin | $\mathrm{X}_{\text {out }}$ |  |  |


| Pin Description | Symbol | Input/Output Signal | Input/Output Internal Circuit |
| :---: | :---: | :---: | :---: |
| Reference voltage signal output pin | $V_{\text {refout }}$ | $5 \pm 0.5 \mathrm{~V}(\max 30 \mathrm{~mA})$ |  |
| Reverse-rotation-detection signal output pin | REV | Digital <br> Push-pull output: $\pm 1$ mA (max) |  |
| FG signal output pin | FG | Digital <br> Push-pull output: $\pm 1 \mathrm{~mA}$ (max) |  |
| Turn-on signal output pin $U$ Turn-on signal output pin V Turn-on signal output pin W Turn-on signal output pin X Turn-on signal output pin $Y$ Turn-on signal output pin Z | $\begin{gathered} U \\ V \\ W \\ W \\ X \\ Y \\ Z \end{gathered}$ | Analog <br> Push-pull output: $\pm 2$ mA (max) <br> L: 0.78 V (max) <br> $\mathrm{H}: \mathrm{V}_{\text {refout }}-0.78 \mathrm{~V}(\mathrm{~min})$ |  |

Absolute Maximum Ratings ( $\mathbf{T a}=\mathbf{2 5}{ }^{\circ} \mathrm{C}$ )

| Characteristics | Symbol | Rating | Unit |
| :--- | :---: | :---: | :---: |
| Supply voltage | $\mathrm{V}_{\mathrm{CC}}$ | 12 | V |
| Input voltage | $\mathrm{V}_{\text {in (1) }}$ | $-0.3 \sim \mathrm{~V}_{\mathrm{CC}} \quad$ (Note 1) | V |
|  | $\mathrm{V}_{\text {in (2) }}$ | $-0.3 \sim \mathrm{~V}_{\text {refout }}+0.3$ (Note 2) |  |
| Turn-on signal output current | $\mathrm{I}_{\mathrm{OUT}}$ | 2 | mA |
| Power dissipation | $\mathrm{P}_{\mathrm{D}}$ | $1.50 \quad$ (Note 3) | W |
| Operating temperature | $\mathrm{T}_{\mathrm{opr}}$ | $-30 \sim 115 \quad$ (Note 4) | ${ }^{\circ} \mathrm{C}$ |
| Storage temperature | $\mathrm{T}_{\text {stg }}$ | $-50 \sim 150$ | ${ }^{\circ} \mathrm{C}$ |

Note 1: $V_{\text {in (1) }}$ pin: $V_{e}, L A, G_{i n}, G_{o u t}, P H, L P F, L L, ~ U L$
Note 2: $\mathrm{V}_{\text {in (2) }} \mathrm{pin}: \mathrm{HU}, \mathrm{HV}, \mathrm{HW}, \mathrm{CW} / C C W, R E S, O S, I_{\mathrm{dc}}$, Td, SS
Note 3: When mounted on PCB (universal $50 \times 50 \times 1.6 \mathrm{~mm}, \mathrm{Cu} 30 \%$ )
Note 4: Operating temperature range is determined by the $P_{D}-T a$ characteristic.
Operating Conditions ( $\mathbf{T a}=25^{\circ} \mathrm{C}$ )

| Characteristics | Symbol | Min | Typ. | Max | Unit |
| :--- | :---: | :---: | :---: | :---: | :---: |
| Supply voltage | $\mathrm{V}_{\mathrm{CC}}$ | 6 | 7 | 10 | V |
| Crystal oscillation frequency | $\mathrm{X}_{\text {in }}$ | 2 | 4 | 8 | MHz |



Electrical Characteristics ( $\mathbf{T a}=\mathbf{2 5}{ }^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{cc}}=\mathbf{7 V}$ )

| Characteristics | Symbol | Test Circuit | Test Condition | Min | Typ. | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Supply current | ICC | - | $V_{\text {refout }}=$ open | - | 5 | 8 | mA |
| Input current | $\mathrm{lin}_{\text {( }} \mathrm{l}^{-1}$ |  | $\mathrm{V}_{\text {in }}=5 \mathrm{~V} \quad \mathrm{LA}$ | - | 25 | 50 | $\mu \mathrm{A}$ |
|  | $\mathrm{I}_{\text {in }(1)-2}$ |  | $\mathrm{V}_{\text {in }}=5 \mathrm{~V} \quad \mathrm{~V}_{\mathrm{e}}$ | - | 35 | 70 |  |
|  | $\mathrm{l}_{\text {in }(2)^{-1}}$ |  | $\mathrm{V}_{\text {in }}=0 \mathrm{~V}$ HU, HV, HW, SS | -50 | -25 | - |  |
|  | $\mathrm{I}_{\text {in (2) }}{ }^{-2}$ |  | $\mathrm{V}_{\text {in }}=0 \mathrm{~V}$ CW/CCW, OS, Td, RES | -100 | -50 | - |  |
| Input voltage | $V_{\text {in }}{ }^{\text {H }}$ HIGH | - | HU, HV, HW, CW/CCW, RES, OS, Td, SS | $\left\lvert\, \begin{gathered} V_{\text {refout }} \\ -1 \end{gathered}\right.$ | - | $V_{\text {refout }}$ | V |
|  | LOW |  |  | - | - | 0.8 |  |
|  | H | - | Modulation factor maximum | 5.1 | 5.4 | 5.7 | V |
|  | $\mathrm{V}_{\mathrm{e}} \quad \mathrm{M}$ |  | Refresh $\rightarrow$ Start motor operation | 1.8 | 2.1 | 2.4 |  |
|  | L |  | Turned-off $\rightarrow$ Refresh | 0.7 | 1.0 | 1.3 |  |
| Input hysteresis voltage | $\mathrm{V}_{\mathrm{H}}$ | - | HU, HV, HW, SS (Note 5) | - | 0.3 | - | V |
| Input delay time | $\mathrm{V}_{\mathrm{DT}}$ | - | $\mathrm{HU}, \mathrm{HV}, \mathrm{HW} \quad \mathrm{X}_{\text {in }}=4.19 \mathrm{MHz}$ | - | 0.5 | - | $\mu \mathrm{S}$ |
|  | $V_{D C}$ |  | Idc $\quad \mathrm{X}_{\text {in }}=4.19 \mathrm{MHz}$ | - | 1.0 | - |  |
| Output voltage | Vout (H)-1 | - | IOUT = 2 mA U, V, W, X, Y, Z | $\begin{gathered} V_{\text {refout }} \\ -0.78 \end{gathered}$ | $\begin{array}{\|c\|} V_{\text {refout }} \\ -0.3 \end{array}$ | - | V |
|  | $\mathrm{V}_{\text {OUT (L)-1 }}$ |  | IOUT $=-2 \mathrm{~mA} \quad \mathrm{U}, \mathrm{V}, \mathrm{W}, \mathrm{X}, \mathrm{Y}, \mathrm{Z}$ | - | 0.3 | 0.78 |  |
|  | $V_{\text {REV }}(\mathrm{H})$ |  | IOUT $=1 \mathrm{~mA}$ REV | $\begin{gathered} V_{\text {refout }} \\ -1.0 \end{gathered}$ | $\begin{gathered} V_{\text {refout }} \\ -0.2 \end{gathered}$ | - |  |
|  | $V_{\text {REV (L) }}$ |  | IOUT $=-1 \mathrm{~mA}$ REV | - | 0.2 | 1.0 |  |
|  | $\mathrm{V}_{\mathrm{FG}}(\mathrm{H})$ |  | IOUT $=1 \mathrm{~mA} \quad$ FG | $V_{\text {refout }}$ $-1.0$ | $\begin{gathered} V_{\text {refout }} \\ -0.2 \end{gathered}$ | - |  |
|  | $\mathrm{V}_{\mathrm{FG}}(\mathrm{L})$ |  | IOUT $=-1 \mathrm{~mA} \quad$ FG | - | 0.2 | 1.0 |  |
|  | $V_{\text {refout }}$ |  | IOUT $=30 \mathrm{~mA} \quad \mathrm{~V}_{\text {refout }}$ | 4.5 | 5.0 | 5.5 |  |
| Output leakage current | l ( H ) | - | $\mathrm{V}_{\text {OUT }}=0 \mathrm{~V}$ | - | 0 | 10 | $\mu \mathrm{A}$ |
|  | L (L) |  | $\mathrm{V}_{\text {OUT }}=3.5 \mathrm{~V} \quad \mathrm{U}, \mathrm{V}, \mathrm{W}, \mathrm{X}, \mathrm{Y}, \mathrm{Z}$ | - | 0 | 10 |  |
| Output off-time by upper/lower transistor (Note 6) | Toff (H) | - | $\mathrm{Td}=\mathrm{HIGH}$ or OPEN, $\mathrm{X}_{\mathrm{in}}=4.19 \mathrm{MHz}$, lout $= \pm 2 \mathrm{~mA}, \mathrm{OS}=$ HIGH/LOW | 1.5 | 1.9 | - | $\mu \mathrm{S}$ |
|  | Toff (L) |  | $\begin{aligned} & \text { Td = LOW, } \mathrm{X}_{\mathrm{in}}=4.19 \mathrm{MHz}, \\ & \mathrm{l} \text { OUT }= \pm 2 \mathrm{~mA}, \mathrm{OS}=\mathrm{HIGH} / \mathrm{LOW} \end{aligned}$ | 3.0 | 3.8 | - |  |
| Overcurrent detection | $\mathrm{V}_{\mathrm{dc}}$ | - | $\mathrm{I}_{\mathrm{dc}}$ | 0.46 | 0.5 | 0.54 | V |
| LA gain setting amp | AMPOUT | - | Gout output current | 5 | - | - | mA |
|  | AMPoFs |  | $\mathrm{GIN}_{\mathrm{N}}, \mathrm{G}_{\text {OUt }} 11 \mathrm{k} \Omega / 1 \mathrm{k} \Omega$ | - | -40 | - | mV |
| LA limit setting difference | $\Delta \mathrm{L}$ | - | LL $=0.7 \mathrm{~V}$ | -20 | - | 20 | mV |
|  | $\Delta \mathrm{U}$ |  | $\mathrm{UL}=2.0 \mathrm{~V}$ | -20 | - | 20 |  |
| LA peak hold output current | PHout | - | PH output current | - | - | 5 | mA |
| Lead angle correction | TLA (0) | - | LA $=0 \mathrm{~V}$ or OPEN, Hall $\mathrm{IN}=100 \mathrm{~Hz}$ | - | 0 | - | - |
|  | TLA (2.5) | - | LA $=2.5 \mathrm{~V}$, Hall $\mathrm{IN}=100 \mathrm{~Hz}$ | 27.5 | 32 | 34.5 |  |
|  | TLA (5) | - | LA $=5 \mathrm{~V}$, Hall $\mathrm{IN}=100 \mathrm{~Hz}$ | 53.5 | 59 | 62.5 |  |
| $\mathrm{V}_{\mathrm{CC}}$ monitor | $\mathrm{V}_{\text {CC }}(\mathrm{H})$ | - | Output start operation point | 4.2 | 4.5 | 4.8 | V |
|  | $\mathrm{V}_{\mathrm{CC}}(\mathrm{L})$ | - | No output operation point | 3.7 | 4.0 | 4.3 |  |
|  | $\mathrm{V}_{\mathrm{H}}$ | - | Input hysteresis width | - | 0.5 | - |  |

Note 5: Toshiba does not implement testing before shipping.

Note 6: TOFF
$\mathrm{OS}=\mathrm{HIGH}$


## Functional Description

## 1. Basic operation

The motor is driven by the square-wave turn-on signal based on a positional signal. When the positional signal reaches number of rotations $f=5 \mathrm{~Hz}$ or higher, the rotor position is estimated according to the positional signal and a modulation wave is generated. The modulation wave and the triangular wave are compared; then the sine-wave PWM signal is generated and the motor is driven.

From start to 5 Hz : When driven by square wave ( $120^{\circ}$ turn-on) $\mathrm{f}=\mathrm{f}_{\mathrm{osc}} /\left(2^{12} \times 32 \times 6\right.$ )
$5 \mathrm{~Hz} \sim$ : When driven by sine-wave PWM ( $180^{\circ}$ turn־on); when $\mathrm{f}_{\text {osc }}=4 \mathrm{MHz}$, approx. 5 Hz

## 2. Select drive function

This function can select drive mode.
SS pin
HIGH or OPEN $=$ Sine-wave PWM drive ( $180^{\circ}$ turn ${ }^{-}$on mode)
LOW $=$ Square-wave drive ( $120^{\circ}$ turn ${ }^{-}$on mode)
Note: If the position sensing signal is $\mathrm{f}=5 \mathrm{~Hz}$ or lower, the driver is $120^{\circ}$ turn-on mode even when $\mathrm{SS}=$ HIGH.

## 3. $V_{e}$ voltage command signal function and function to stabilize bootstrap voltage

(1) When the voltage command signal is input at $\mathrm{V}_{\mathrm{e}} \leqq 1.0 \mathrm{~V}$ :

Turns off output (gate protection)
(2) When the voltage command signal is input at $1.0 \mathrm{~V}<\mathrm{Ve} \leqq 2.1 \mathrm{~V}$ :

Turns on the lower transistor at the regular (carrier) cycle. (ON duty is approx. 8\%.)
(3) When the voltage command signal is input at $\mathrm{Ve}>2.1 \mathrm{~V}$ :

During sin-wave drive, outputs drive signal as it is. During square-drive, forcibly turns on the lower transistor at regular (carrier) cycle. (ON duty is approx. 8\%)

Note: At startup, turn the lower transistor on for a fixed time with $1.0 \mathrm{~V}<\mathrm{Ve} \leqq 2.1 \mathrm{~V}$ to charge the upper transistor gate power supply.


## 4. Dead time function: upperllower transistor output off-time

When the motor is driven by sine-wave PWM, dead time is digitally generated in the IC to prevent a short circuit caused by the simultaneous turning on of upper and lower external power devices. When a square wave is generated in full-duty cycle mode, the dead time function is turned on to prevent a short circuit.

| Td Pin | Internal Counter | TOFF |
| :---: | :---: | :---: |
| HIGH or OPEN | $8 / \mathrm{f}_{\text {Osc }}$ | $1.9 \mu \mathrm{~s}$ |
| LOW | $16 / \mathrm{f}_{\text {osc }}$ | $3.8 \mu \mathrm{~s}$ |

Toff values above are obtained when fosc $=4.19 \mathrm{MHz}$. $\mathrm{f}_{\text {osc }}=$ reference clock (crystal oscillation)

## 5. Correcting the lead angle

The lead angle can be corrected in the turn-on signal range from 0 to $58^{\circ}$ in relation to the induced voltage.

Analog input from LA pin ( 0 V to 5 V divided by 32):
$0 \mathrm{~V}=0^{\circ}$
$5 \mathrm{~V}=58^{\circ}$ (when more than 5 V is input, $58^{\circ}$ )
6. Setting the carrier frequency

This function sets the triangular wave cycle (carrier cycle) necessary for generating the PWM signal. (The triangular wave is used for forcibly turning on the lower transistor when the motor is driven by square wave.)

Carrier cycle $=\mathrm{f}_{\mathrm{osc}} / 252(\mathrm{~Hz}) \quad \mathrm{f}_{\mathrm{osc}}=$ reference clock (crystal oscillation)

## 7. Switching the output of the turn-on signal

This function switches the output of the turn-on signal between HIGH and LOW.
Pin OS:
HIGH = active HIGH
LOW = active LOW

## 8. Outputting the reverse rotation detection signal

This function detects the motor rotation direction every electrical angle of $360^{\circ}$. (The output is HIGH immediately after reset.)

The REV terminal increases with a $180^{\circ}$ turn-on mode during LOW.

| CW/CCW Pin | Actual Motor Rotating Direction | REV Pin |
| :---: | :---: | :---: |
|  | CW (forward) | LOW |
|  | CCW (reverse) | HIGH |
| HIGH (CCW) | CW (forward) | HIGH |
|  | CCW (reverse) | LOW |

## 9. Protecting input pin

1. Overcurrent protection (Pin Idc)

When the DC-link-current exceeds the internal reference voltage, performs gate block protection.
Overcurrent protection is released for each carrier frequency.
Reference voltage $=0.5 \mathrm{~V}$ (typ.)
2. Gate protection (Pin RES)

Output is turned off when the input signal is LOW, restarted when the input signal is HIGH.
The abnormality is detected externally and the signal input to pin RES.

| RES Pin | OS Pin | Output Turn-on Signal <br> $(\mathrm{U}, \mathrm{V}, \mathrm{W}, \mathrm{X}, \mathrm{Y}, \mathrm{Z})$ |
| :---: | :---: | :---: |
| LOW | LOW | HIGH |
|  | HIGH | LOW |

(When RES = LOW, bootstrap capacitor charging stops.)
3. Internal protection

- Positional signal abnormality protection

Output is turned off when the positional signal is HHH or LLL; otherwise, it is restarted.

- Low power supply voltage protection (VCC monitor)

For power supply on/off outside the operating voltage range, the turn-on signal output is kept at high impedance outside the operating voltage range to prevent damage caused by power device short circuits.
However, if the voltage level is supplied from the $V_{e}$ pin, this function is restricted, e.g., when $V_{e}>$ 4.9 V is applied, low power supply voltage protection does not operate.


## Operation Flow



Note: Output ON time is decreased by the dead time (carrier frequency $\times 92 \%-T_{d} \times 2$ )


The modulation waveform is generated using Hall signals. The modulation waveform is then compared with the triangular wave and a sine-wave PWM signal is generated.

The time (electrical degrees: $60^{\circ}$ ) from the rising (or falling) edges of the three Hall signals to the next falling (or rising) edges is counted. The counted time is used as the data for the next $60^{\circ}$ phase of the modulation waveform.
There are 32 items of data for the $60^{\circ}$ phase of the modulation waveform. The time width of one data item is $1 / 32$ of the time width of the $60^{\circ}$ phase of the previous modulation waveform. The modulation waveform moves forward by the width.


In the above diagram, the modulation waveform (1)' data moves forward by the $1 / 32$ time width of the time (1) from HU: $\uparrow$ to HW: $\downarrow$. Similarly, data (2)' moves forward by the $1 / 32$ time width of the time (2) from $\mathrm{HW}: \downarrow$ to $\mathrm{HV}: \uparrow$.

If the next edge does not occur after the 32 data items end, the next 32 data items move forward by the same time width until the next edge occurs.


The modulation wave is brought into phase with every zero-cross point of the Hall signal.
The modulation wave is reset in synchronization with the rising and falling edges of the Hall signal at every $60^{\circ}$ electrical angle. Thus, when the Hall device is not placed at the correct position or during acceleration and deceleration, the modulation waveform is not continuous at every reset.

## Timing Charts




Reverse

Operating Waveform When Driven by Square Wave (CWICCW = LOW, OS = HIGH)

Hall signal


Output waveform


To stabilize the bootstrap voltage, the lower outputs ( $\mathrm{X}, \mathrm{Y}$, and Z ) are always turned on at the carrier cycle even during off time. At that time, the upper outputs ( $\mathrm{U}, \mathrm{V}$, and W ) are assigned dead time and turned off at the timing when the lower outputs are turned on. ( $\mathrm{T}_{\mathrm{d}}$ varies with input $\mathrm{V}_{\mathrm{e}}$.)

$$
\begin{array}{ll}
\text { Carrier cycle }=\mathrm{f}_{\mathrm{osc}} / 252(\mathrm{~Hz}) & \text { Dead time: } \mathrm{T}_{\mathrm{d}}=16 / \mathrm{f}_{\mathrm{osc}}(\mathrm{~s})\left(\text { In more than } \mathrm{V}_{\mathrm{e}}=5.0 \mathrm{~V}\right) \\
\text { TONL }=\text { carrier cycle } \times 8 \%(\mathrm{~s}) \text { (Uniformity) } &
\end{array}
$$

When the motor is driven by a square wave, acceleration or deceleration is determined by voltage $\mathrm{V}_{\mathrm{e}}$. The motor accelerates or decelerates according to the ON duty of TONU. (See the diagram of output ON duty on page 14.)

Note: At startup, the motor is driven by a square wave when the Hall signals are 5 Hz or lower ( $\mathrm{f}_{\mathrm{osc}}=4 \mathrm{MHz}$ ) and the motor is rotating in the reverse direction to that of the TB6556F/FG controlling it (REV = HIGH).

Generation inside of IC

Phase U


Phase V


Phase W


Output waveform


Inter-line voltage


When driven by a sine wave, the motor is accelerated or decelerated according to the ON duty of TONU as the amplitude of the modulation symbol changes according to voltage Ve . (See the diagram of the output ON duty on page 14.)

Triangular wave frequency $=$ carrier frequency $=\mathrm{f}_{\mathrm{osc}} / 252(\mathrm{~Hz})$
Note: At startup, the motor is driven by a sine wave when the Hall signals are 5 Hz or higher ( $\mathrm{f}_{\mathrm{osc}}=4 \mathrm{MHz}$ ) and the motor is rotating in the same direction as the TB6556F/FG controlling it (REV = LOW).

Example of Application Circuit


Note 1: Connect to ground as necessary to prevent IC malfunction due to noise.
Note 2: Connect GND to signal ground on the application circuit.
 short-circuiting between contiguous pins

Package Dimensions


Unit : mm


Weight: 0.63 g (typ.)

## Notes on Contents

## 1. Block Diagrams

Some of the functional blocks, circuits, or constants in the block diagram may be omitted or simplified for explanatory purposes.

## 2. Equivalent Circuits

The equivalent circuit diagrams may be simplified or some parts of them may be omitted for explanatory purposes.

## 3. Timing Charts

Timing charts may be simplified for explanatory purposes.

## 4. Application Circuits

The application circuits shown in this document are provided for reference purposes only. Thorough evaluation is required, especially at the mass production design stage.
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## 5. Test Circuits

Components in the test circuits are used only to obtain and confirm the device characteristics. These components and circuits are not guaranteed to prevent malfunction or failure from occurring in the application equipment.

## IC Usage Considerations Notes on handling of ICs

[1] The absolute maximum ratings of a semiconductor device are a set of ratings that must not be exceeded, even for a moment. Do not exceed any of these ratings.
Exceeding the rating(s) may cause the device breakdown, damage or deterioration, and may result injury by explosion or combustion.
[2] Do not insert devices in the wrong orientation or incorrectly.
Make sure that the positive and negative terminals of power supplies are connected properly.
Otherwise, the current or power consumption may exceed the absolute maximum rating, and exceeding the rating(s) may cause the device breakdown, damage or deterioration, and may result injury by explosion or combustion.
In addition, do not use any device that is applied the current with inserting in the wrong orientation or incorrectly even just one time.

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