VN31SP

HIGH SIDE SMART POWER SOLID STATE RELAY

| TYPE	V_{DSS}	$R_{DS(on)}$	$I_{n(*)}$	V_{CC}
VN31SP	60 V	0.03 Ω	11.5 A	26 V

■ MAXIMUM CONTINUOUS OUTPUT CURRENT (#):31 A @ T_c=85°C
■ 5 V LOGIC LEVEL COMPATIBLE INPUT
■ THERMAL SHUT-DOWN
■ UNDER VOLTAGE PROTECTION
■ OPEN DRAIN DIAGNOSTIC OUTPUT
■ INDUCTIVE LOAD FAST DEMAGNETIZATION
■ VERY LOW STAND-BY POWER DISSIPATION

PowerSO-10™

DESCRIPTION

The VN31SP is a monolithic device made using STMicroelectronics VIPower Technology, intended for driving resistive or inductive loads with one side grounded.
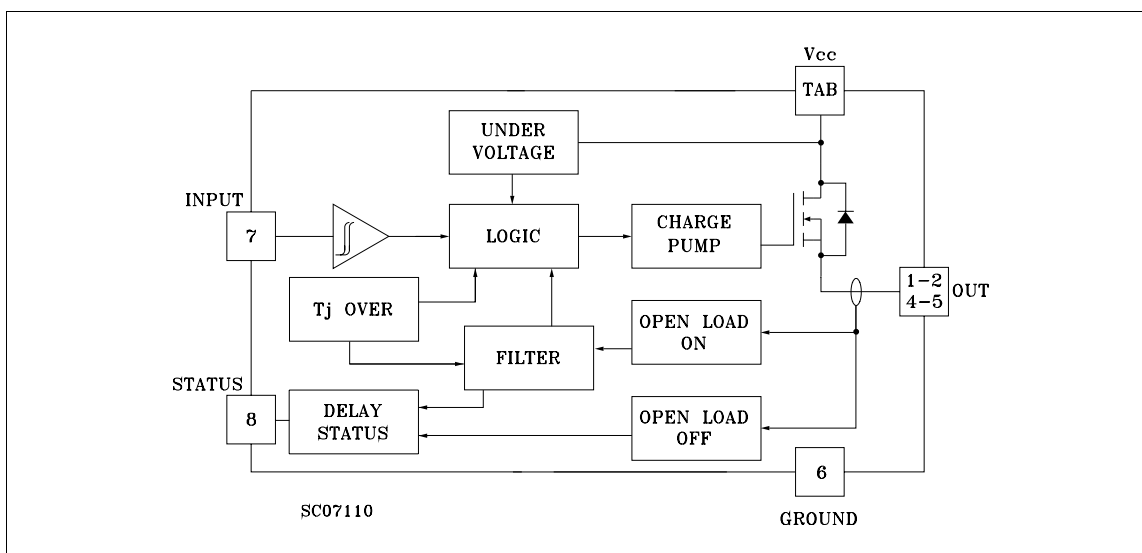
Built-in thermal shut-down protects the chip from over temperature and short circuit.

The open drain diagnostic output indicates: open load in off state, and in on state, output shorted to
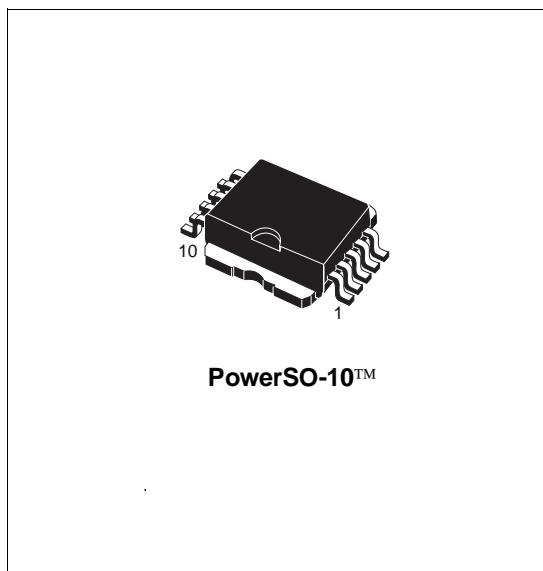
V_{CC} and overtemperature. Fast demagnetization of inductive loads is archivied by negative (-18V) load voltage at turn-off.

BLOCK DIAGRAM

SC07110

(*) In = Nominal current according to ISO definition for high side automotive switch (see note 1)
(#) The maximum continuous output current is the the current at T_c = 85 °C for a battery voltage of 13V which does not activate self protection.

ABSOLUTE MAXIMUM RATING

Symbol	Parameter	Value	Unit
$V_{(BR)DSS}$	Drain-Source Breakdown Voltage	60	V
I_{OUT}	Output Current (cont.) at $T_c = 85\ ^oC$	31	A
I_R	Reverse Output Current at $T_c = 85\ ^oC$	-31	A
I_{IN}	Input Current	± 10	mA
$-V_{CC}$	Reverse Supply Voltage	-4	V
I_{STAT}	Status Current	± 10	mA
V_{ESD}	Electrostatic Discharge (1.5 kΩ, 100 pF)	2000	V
P_{tot}	Power Dissipation at $T_c = 85\ ^oC$	54	W
T_j	Junction Operating Temperature	-40 to 150	^oC
T_{stg}	Storage Temperature	-55 to 150	^oC

CONNECTION DIAGRAMS
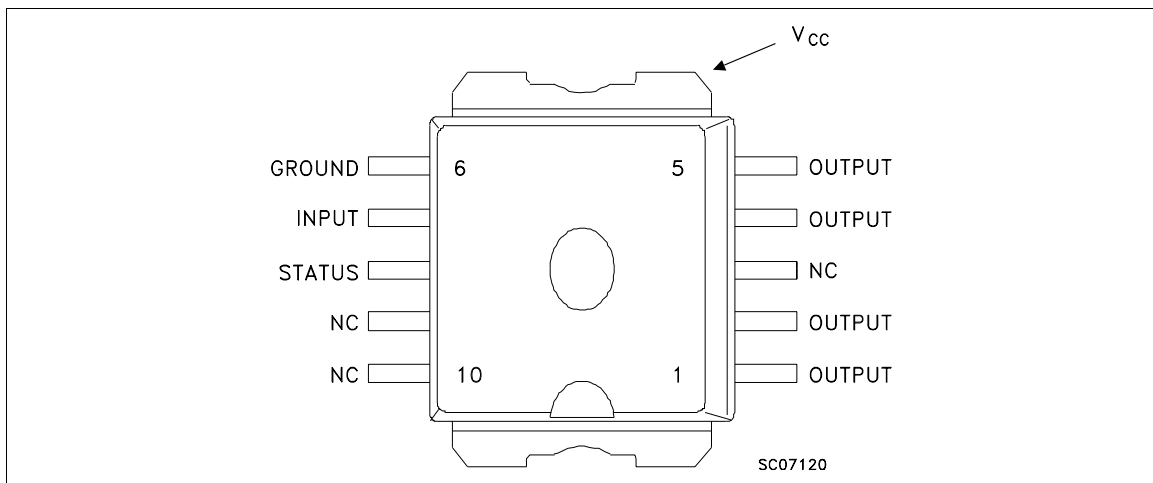
CURRENT AND VOLTAGE CONVENTIONS
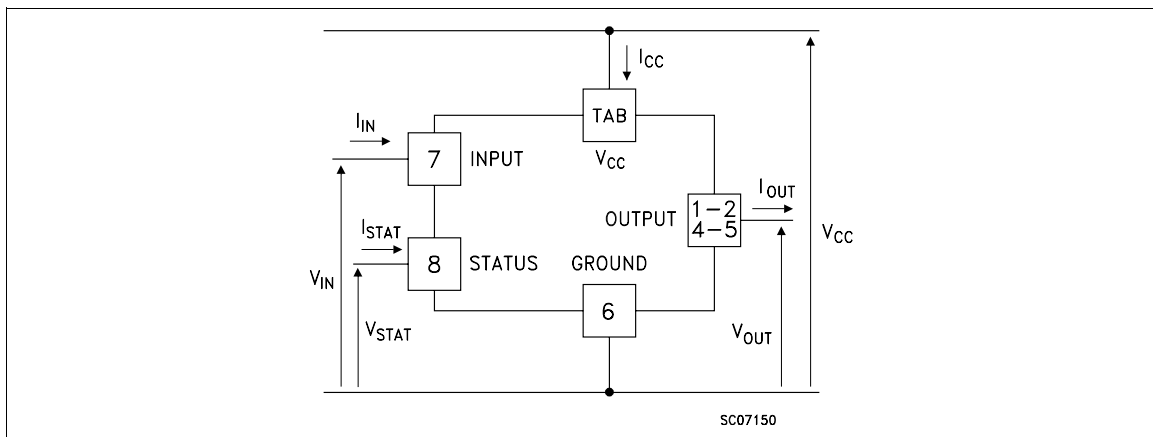
THERMAL DATA

$R_{thj-case}$	Thermal Resistance Junction-case	Max	1.2	$^{o}C/W$
$R_{thj-amb}$	Thermal Resistance Junction-ambient ($)	Max	50	$^{o}C/W$

($) When mounted using minimum recommended pad size on FR-4 board

ELECTRICAL CHARACTERISTICS (V_{CC} = 13 V; -40 $\leq$ T_j $\leq$ 125 ^{o}C unless otherwise specified)

POWER

Symbol	Parameter	Test Conditions	Min.	Typ.	Max.	Unit
V_{CC}	Supply Voltage		5.5	13	26	V
In(*)	Nominal Current	T_c = 85 ^{o}C $V_{DS(on)}$ $\leq$ 0.5 (note 1)	11.5			A
R_{on}	On State Resistance	I_{OUT} = 11.5 A I_{OUT} = 11.5 A T_j = 25 ^{o}C			0.06 0.03	Ω Ω
I_S	Supply Current	Off State T_j $\geq$ 25 ^{o}C On State			50 15	µA mA
$V_{DS(MAX)}$	Maximum Voltage Drop	I_{OUT} = 25 A T_c = 85 ^{o}C			1.5	V

SWITCHING

Symbol	Parameter	Test Conditions	Min.	Typ.	Max.	Unit
$t_{d(on)}$(^)	Turn-on Delay Time Of Output Current	I_{OUT} = 11.5 A Resistive Load Input Rise Time < 0.1 µs		90		µs
t_r(^)	Rise Time Of Output Current	I_{OUT} = 11.5A Resistive Load Input Rise Time < 0.1 µs		100		µs
$t_{d(off)}$(^)	Turn-off Delay Time Of Output Current	I_{OUT} = 11.5 A Resistive Load Input Rise Time < 0.1 µs		140		µs
t_f(^)	Fall Time Of Output Current	I_{OUT} = 11.5 A Resistive Load Input Rise Time < 0.1 µs		50		µs
$(di/dt)_{on}$	Turn-on Current Slope	I_{OUT} = 11.5 A I_{OUT} = I_{OV}		0.08	0.5 1	A/µs A/µs
$(di/dt)_{off}$	Turn-off Current Slope	I_{OUT} = 11.5 A I_{OUT} = I_{OV}		0.2	3 3	A/µs A/µs
V_{demag}	Inductive Load Clamp Voltage	I_{OUT} = 11.5 A L = 1 mH	-24	-18	-14	V

LOGIC INPUT

Symbol	Parameter	Test Conditions	Min.	Typ.	Max.	Unit
V_{IL}	Input Low Level Voltage				0.8	V
V_{IH}	Input High Level Voltage		2		(•)	V
$V_{I(hyst.)}$	Input Hysteresis Voltage			0.5		V
I_{IN}	Input Current	V_{IN} = 5 V V_{IN} = 2 V V_{IN} = 0.8 V	 25	250	500 250	µA µA µA
V_{ICL}	Input Clamp Voltage	I_{IN} = 10 mA I_{IN} = -10 mA	5.5	6 -0.7	 -0.3	V V

ELECTRICAL CHARACTERISTICS (continued)

PROTECTION AND DIAGNOSTICS

Symbol	Parameter	Test Conditions	Min.	Typ.	Max.	Unit
V_{STAT}	Status Voltage Output Low	I_{STAT} = 1.6 mA			0.4	V
V_{USD}	Under Voltage Shut Down			5		V
V_{SCL}	Status Clamp Voltage	I_{STAT} = 10 mA I_{STAT} = -10 mA		6 -0.7		V V
I_{OV}	Over Current	R_{LOAD} < 10 mΩ -40 T_c 125 ^oC		140		A
I_{AV}	Average Current in Short Circuit	R_{LOAD} < 10 mΩ T_c = 85 ^oC		2.5		A
I_{OL}	Open Load Current Level		5	600	1250	mA
T_{TSD}	Thermal Shut-down Temperature		140			^oC
T_R	Reset Temperature		125			^oC
V_{OL}	Open Load Voltage Level	Off-State (note 2)	2.5	3.75	5	V
$t_{1(on)}$	Open Load Filtering Time	(note 3)	1	5	10	ms
$t_{1(off)}$	Open Load Filtering Time	(note 3)	1	5	10	ms
$t_{2(off)}$	Open Load Filtering Time	(note 3)	1	5	10	ms
t_{povl}	Status Delay	(note 3)		5	10	µs
t_{pol}	Status Delay	(note 3)	50	700		µs

(^) See Switchig Time Waveforms

() The V_{IH} is internally clamped at 6V about. It is possible to connect this pin to an higher voltage via an external resistor calculated to not exceed 10 mA at the input pin.

note 1: The Nominal Current is the current at T_c = 85 ^oC for battery voltage of 13V which produces a voltage drop of 0.5 V

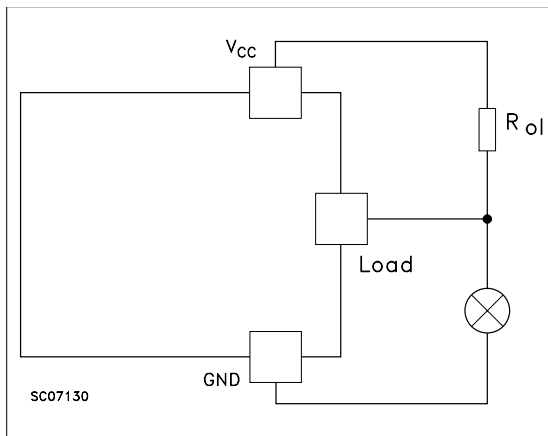note 2: $I_{OL(off)}$ = (V_{CC} -V_{OL})/R_{OL} (see figure)

note 3: $t_{1(on)}$: minimum open load duration which acctivates the status output

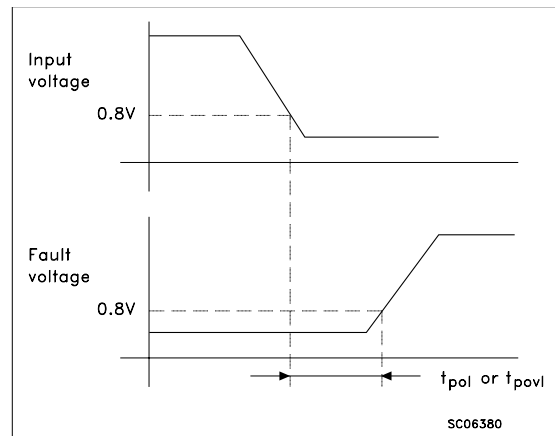$t_{1(off)}$: minimum load recovery time which desactivates the status output

$t_{2(off)}$: minimum on time after thermal shut down which desactivates status output

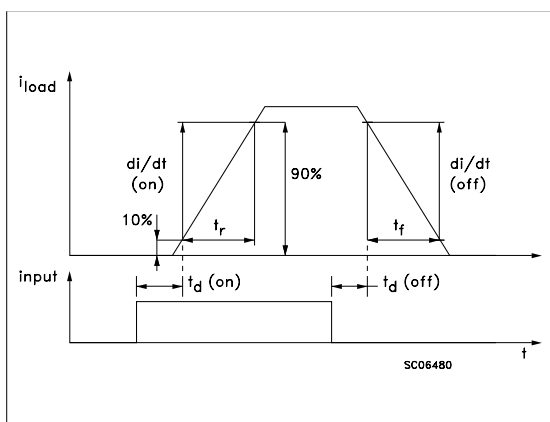t_{povl} t_{pol}: ISO definition (see figure)

Note 2 Relevant Figure

Note 3 Relevant Figure

Switching Time Waveforms

SC06480

FUNCTIONAL DESCRIPTION

The device has a diagnostic output which indicates open load conditions in off state as well as in on state, output shorted to V_{CC} and overtemperature. The truth table shows input, diagnostic and output voltage level in normal operation and in fault conditions. The output signals are processed by internal logic. The open load diagnostic output has a 5 ms filtering. The filter gives a continuous signal for the fault condition after an initial delay of about 5 ms. This means that a disconnection during normal operation, with a duration of less than 5 ms does not affect the status output. Equally, any re-connection of less than 5 ms during a disconnection duration does not affect the status output. No delay occur for the status to go low in case of overtemperature conditions. From the falling edge of the input signal the status output initially low in fault condition (over temperature or open load) will go back with a delay (t_{povl})in case of overtemperature condition and a delay (t_{pol}) in case of open load. These feature fully comply with International Standard Office (I.S.O.) requirement for automotive High Side Driver.

To protect the device against short circuit and over current conditions, the thermal protection turns the integrated Power MOS off at a minimum junction temperature of 140 ^{o}C. When the temperature returns to 125 ^{o}C the switch is automatically turned on again. In short circuit the protection reacts with virtually no delay, the sensor being located in the region of the die where the heat is generated. Driving inductive loads, an internal function of the device ensures the fast demagnetization with a

typical voltage (V_{demag}) of -18V.
This function allows to greatly reduce the power dissipation according to the formula:
$P_{dem} = 0.5 \bullet L_{load} \bullet (I_{load})^2 \bullet [(V_{CC}+V_{demag})/V_{demag}] \bullet f$

where f = switching frequency and
V_{demag} = demagnetization voltage
Based on this formula it is possible to know the value of inductance and/or current to avoid a thermal shut-down. The maximum inductance which causes the chip temperature to reach the shut down temperature in a specific thermal environment, is infact a function of the load current for a fixed V_{CC}, V_{demag} and f.

PROTECTING THE DEVICE AGAIST LOAD DUMP - TEST PULSE 5

The device is able to withstand the test pulse No. 5 at level II (V_s = 46.5V) according to the ISO T/R 7637/1 without any external component. This means that all functions of the device are performed as designed after exposure to disturbance at level II. The VN06SP is able to withstand the test pulse No.5 at level III adding an external resistor of 150 ohm between GND pin and ground plus a filter capacitor of 1000 µF between V_{CC} pin and ground (if $R_{LOAD} \le 20 \Omega$).

PROTECTING THE DEVICE AGAINST REVERSE BATTERY

The simplest way to protect the device against a continuous reverse battery voltage (-26V) is to insert a Schottky diode between GND pin and ground, as shown in the typical application circuit (fig.3).
The consequences of the voltage drop across this diode are as follows:
If the input is pulled to power GND, a negative voltage of -V_f is seen by the device. (Vil, Vih thresholds and Vstat are increased by Vf with respect to power GND).
The undervoltage shutdown level is increa- sed by Vf.
If there is no need for the control unit to handle external analog signals referred to the power GND, the best approach is to connect the reference potential of the control unit to node [6] (see application circuit in fig. 4), which becomes the common signal GND for the whole control board avoiding shift of V_{ih}, V_{il} and V_{stat}. This solution allows the use of a standard diode.

TRUTH TABLE

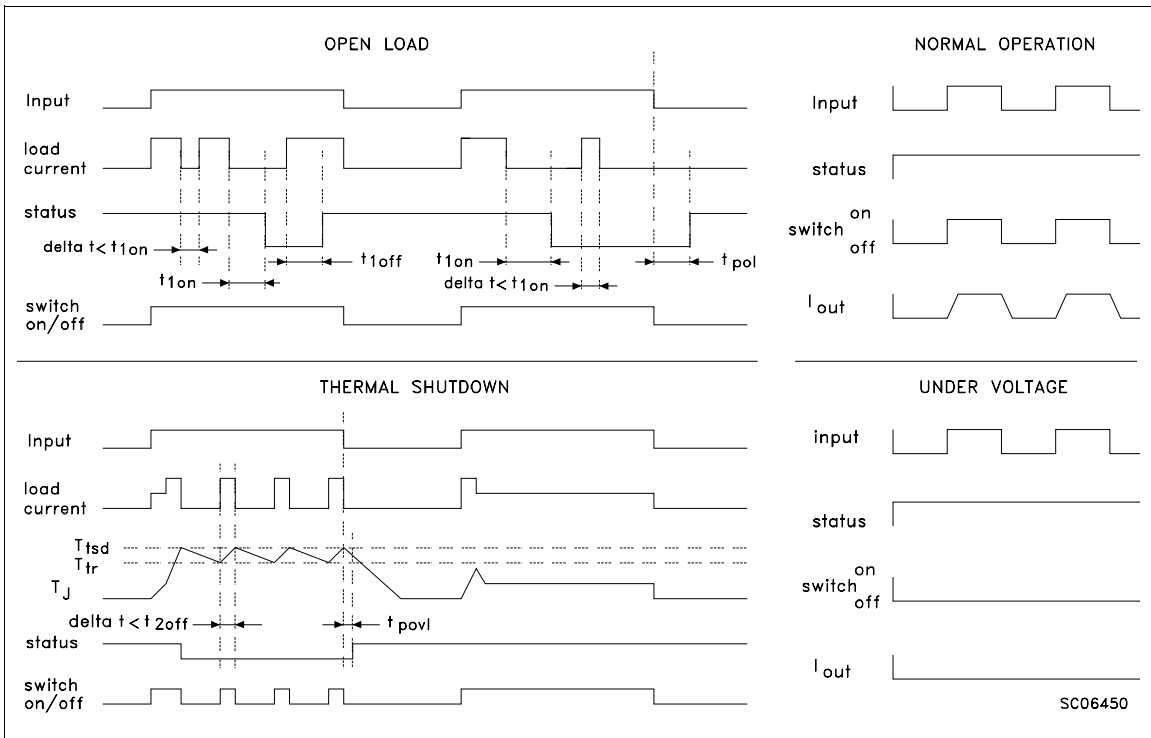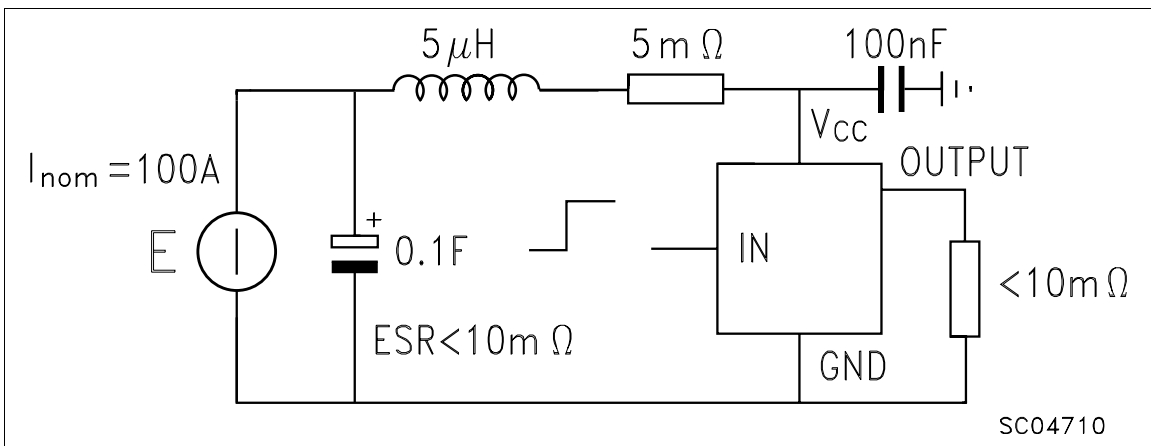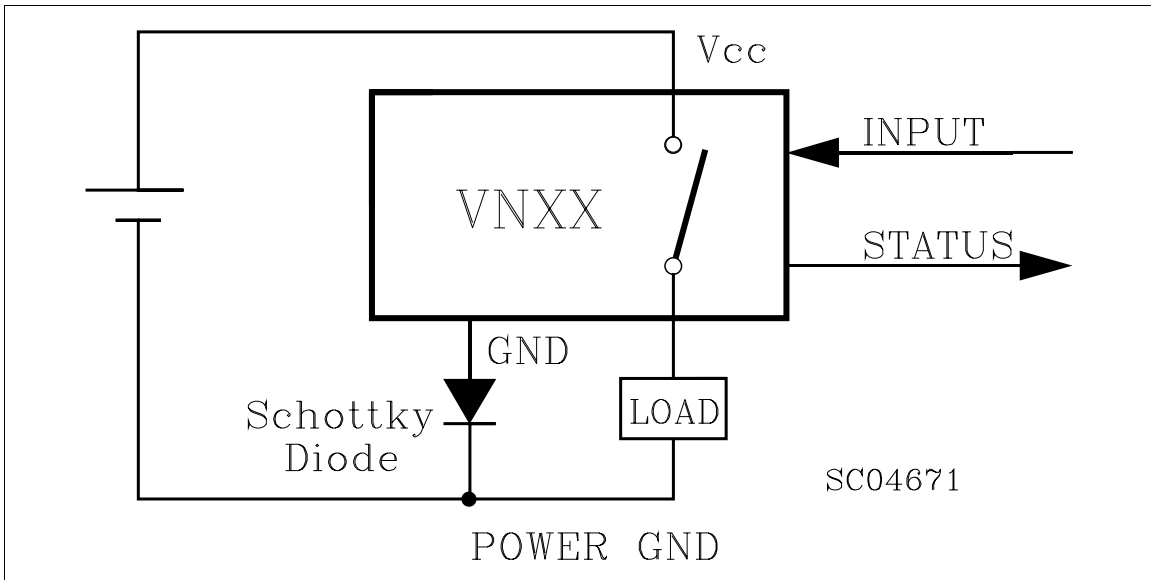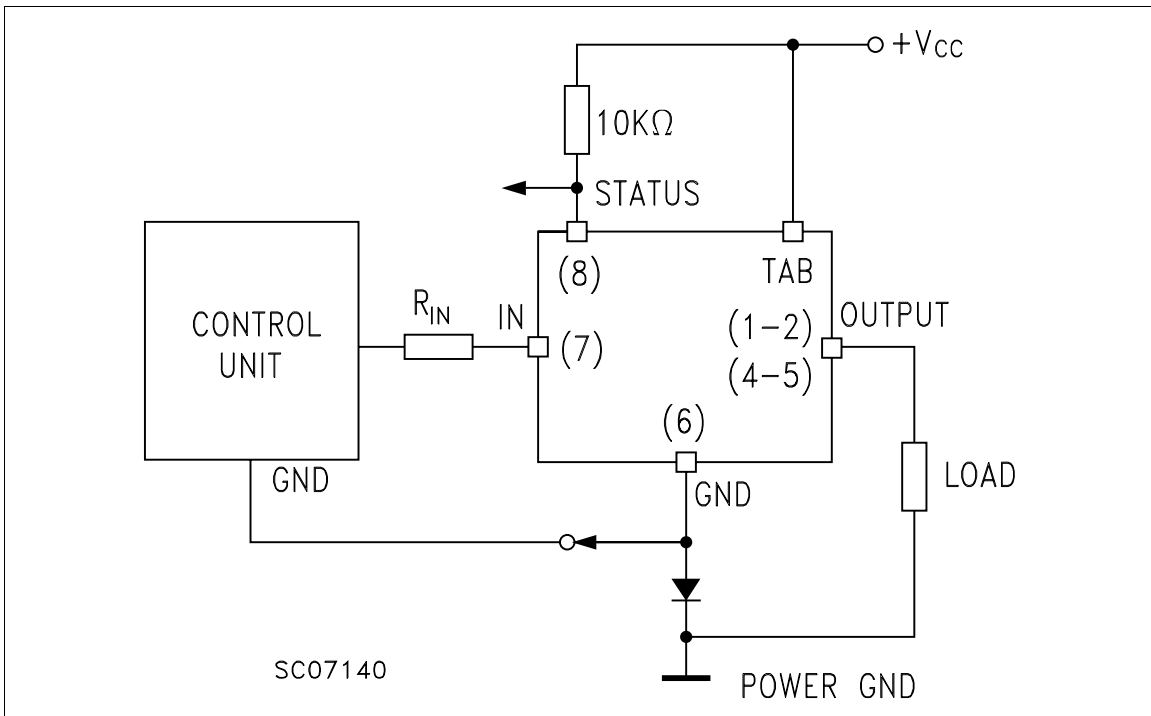|	INPUT	OUTPUT	DIAGNOSTIC
Normal Operation	L H	L H	H H
Open Circuit (No Load)	H	H	L
Over-temperature	H	L	L
Under-voltage	X	L	H
Short load to V_{CC}	L	H	L

Figure 1: Waveforms

Figure 2: Over Current Test Circuit

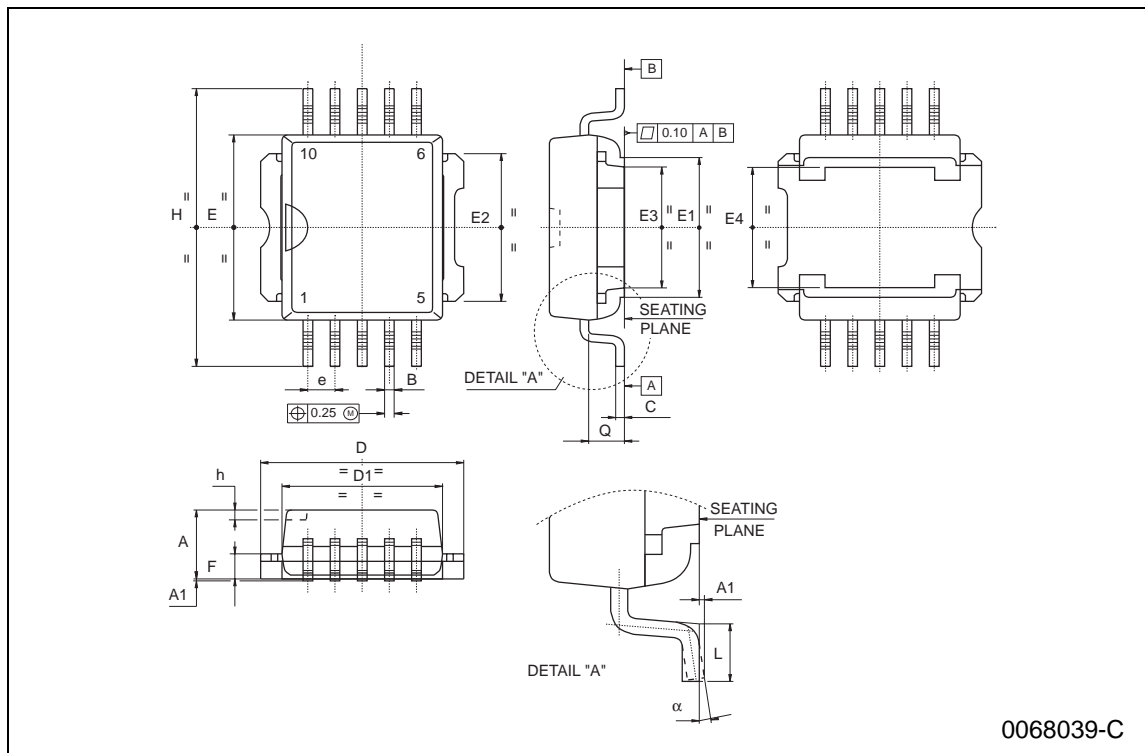**Figure 3:** Typical Application Circuit With A Schottky Diode For Reverse Supply Protection

Figure 4: Typical Application Circuit With Separate Signal Ground

PowerSO-10 MECHANICAL DATA

DIM.	mm			inch		
	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.
A	3.35		3.65	0.132		0.144
A1	0.00		0.10	0.000		0.004
B	0.40		0.60	0.016		0.024
c	0.35		0.55	0.013		0.022
D	9.40		9.60	0.370		0.378
D1	7.40		7.60	0.291		0.300
E	9.30		9.50	0.366		0.374
E1	7.20		7.40	0.283		0.291
E2	7.20		7.60	0.283		0.300
E3	6.10		6.35	0.240		0.250
E4	5.90		6.10	0.232		0.240
e		1.27			0.050	
F	1.25		1.35	0.049		0.053
H	13.80		14.40	0.543		0.567
h		0.50			0.002	
L	1.20		1.80	0.047		0.071
q		1.70			0.067	
α	0°		8°			

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