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1 Electrical data

1.1 Absolute maximum ratings

Table 1. Absolute maximum ratings

Symbol	Parameter	Value	Unit
V_{out}	Output voltage	-3 to $V_{boot} - 18$	V
V_{cc}	Supply voltage	- 0.3 to +18	V
V_{boot}	Floating supply voltage	-1 to 618	V
V_{hvg}	High-side gate output voltage	- 1 to V_{boot}	V
V_{lvg}	Low-side gate output voltage	-0.3 to $V_{cc} + 0.3$	V
V_i	Logic input voltage	-0.3 to $V_{cc} + 0.3$	V
V_{diag}	Open drain forced voltage	-0.3 to $V_{cc} + 0.3$	V
V_{cin}	Comparator input voltage	-0.3 to 10 V	V
dV_{out}/dt	Allowed output slew rate	50	V/ns
P_{tot}	Total power dissipation ($T_J = 85\text{ °C}$)	750	mW
T_J	Junction temperature	150	°C
T_{stg}	Storage temperature	-50 to 150	°C

Note: ESD immunity for pins 12, 13 and 14 is guaranteed up to 900V (Human Body Model)

1.2 Thermal data

Table 2. Thermal data

Symbol	Parameter	SO-14	Unit
$R_{th(JA)}$	Thermal Resistance Junction to ambient	165	°C/W

1.3 Recommended operating conditions

Table 3. Recommended operating conditions

Symbol	Pin	Parameter	Test condition	Min	Typ	Max	Unit
V_{out}	12	Output voltage		(1)		580	V
$V_{BS}^{(2)}$	14	Floating supply voltage		(1)		17	V
f_{sw}		Switching frequency	HVG,LVG load $C_L = 1\text{ nF}$			400	kHz
V_{cc}	4	Supply voltage				17	V
T_J		Junction temperature		-45		125	°C

1. If the condition $V_{boot} - V_{out} < 18\text{ V}$ is guaranteed, V_{out} can range from -3 to 580 V

2. $V_{BS} = V_{boot} - V_{out}$

2 Pin connection

Figure 2. Pin connection (Top view)

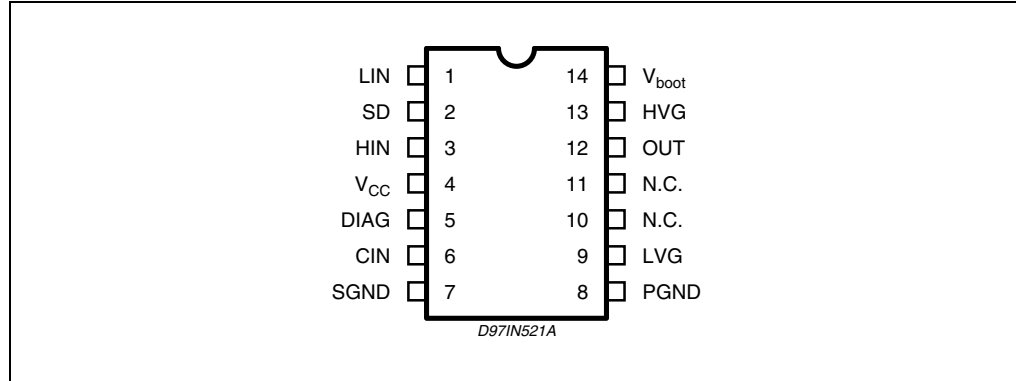


Table 4. Pin description

N°	Pin	Type	Function
1	LIN	I	Low-side driver logic input
2	SD ⁽¹⁾	I	Shut down logic input
3	HIN	I	High-side driver logic input
4	V _{CC}		Low voltage supply
5	DIAG	O	Open drain diagnostic output
6	CIN	I	Comparator input
7	SGND		Ground
8	PGND		Power ground
9	LVG ⁽¹⁾	O	Low-side driver output
10, 11	N.C.		Not connected
12	OUT	O	High-side driver floating driver
13	HVG ⁽¹⁾	O	High-side driver output
14	V _{boot}		Bootstrapped supply voltage

1. The circuit guarantees 0.3V maximum on the pin (@ I_{sink} = 10 mA), with V_{CC} > 3V. This allows to omit the "bleeder" resistor connected between the gate and the source of the external MOSFET normally used to hold the pin low; the gate driver assures low impedance also in SD condition.

3 Electrical characteristics

3.1 AC operation

Table 5. AC operation electrical characteristics ($V_{CC} = 15\text{ V}$; $T_J = 25\text{ }^\circ\text{C}$)

Symbol	Pin	Parameter	Test condition	Min	Typ	Max	Unit
t_{on}	1,3 vs 9,13	High/low-side driver turn-on propagation delay	$V_{out} = 0\text{ V}$		110	150	ns
t_{off}		High/low-side driver turn-off propagation delay			110	150	ns
t_{sd}	2 vs 9,13	Shut down to high/low side propagation delay			105	150	
t_r	9, 13	Rise time	$C_L = 1000\text{ pF}$		50		ns
t_f		Fall time	$C_L = 1000\text{ pF}$		30		ns

3.2 DC operation

Table 6. DC operation electrical characteristics ($V_{CC} = 15\text{ V}$; $T_J = 25\text{ }^\circ\text{C}$)

Symbol	Pin	Parameter	Test condition	Min	Typ	Max	Unit
Low supply voltage section							
V_{ccth1}	4	V_{CC} UV turn on threshold		9.1	9.6	10.1	V
V_{ccth2}		V_{CC} UV turn off threshold		7.9	8.3	8.8	V
V_{cchys}		V_{CC} UV hysteresis			1.3		V
I_{qccu}		Undervoltage quiescent supply current	$V_{CC} \leq 9\text{ V}$		200		μA
I_{qcc}		Quiescent current	$V_{CC} = 15\text{ V}$		250	320	μA
Bootstrapped supply section							
V_{bth1}	14	V_{boot} UV turn on threshold		8.5	9.5	10.5	V
V_{bth2}		V_{boot} UV turn off threshold		7.2	8.2	9.2	V
V_{bhys}		V_{boot} UV hysteresis			1.3		V
I_{qboot}		V_{boot} quiescent current	HVG ON			200	μA
I_{lk}		High voltage leakage current	$V_{hvg} = V_{out} = V_{boot} = 600\text{ V}$			10	μA
$R_{DS(on)}$		Bootstrap driver on resistance ⁽¹⁾	$V_{CC} \geq 12.5\text{ V}$; $V_{in} = 0\text{ V}$			125	Ω

Table 6. DC operation electrical characteristics (continued) ($V_{CC} = 15\text{ V}$; $T_J = 25\text{ °C}$)

Symbol	Pin	Parameter	Test condition	Min	Typ	Max	Unit	
Driving buffers section								
I_{so}	9, 13	High/low side source short circuit current	$V_{IN} = V_{ih}$ ($t_p < 10\ \mu\text{s}$)	300	400		mA	
I_{si}	9, 13	High/low side sink short circuit current	$V_{IN} = V_{il}$ ($t_p < 10\ \mu\text{s}$)	500	650		mA	
Logic inputs								
V_{il}	1, 2, 3	Low level logic voltage				1.5	V	
V_{ih}		High level logic voltage		3.6			V	
I_{ih}		High level logic input current	$V_{IN} = 15\text{ V}$		50	70		μA
I_{il}		Low level logic input current	$V_{IN} = 0\text{ V}$			1		μA
Sense comparator								
V_{io}		Input offset voltage		-10		10	mV	
I_{io}	6	Input bias current	$V_{cin} \geq 0.5$		0.2		μA	
V_{ol}	2	Open drain low level output voltage	$I_{od} = -2.5\text{ mA}$			0.8	V	
V_{ref}		Comparator reference voltage		0.46	0.50	0.54	V	

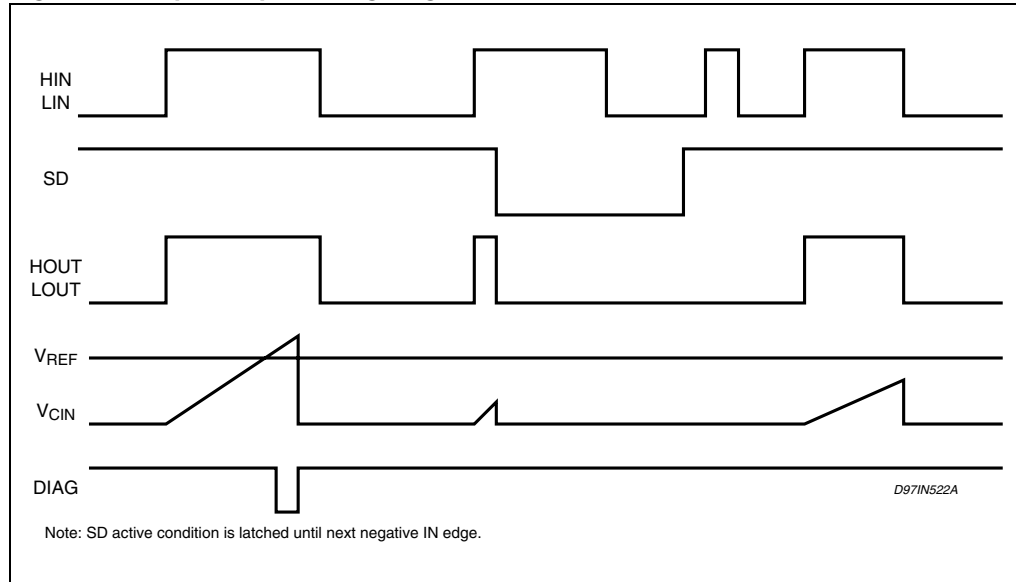
1. $R_{DS(on)}$ is tested in the following way:

$$R_{DS(on)} = \frac{(V_{CC} - V_{CBOOT1}) - (V_{CC} - V_{CBOOT2})}{I_1(V_{CC}, V_{CBOOT1}) - I_2(V_{CC}, V_{CBOOT2})}$$

where I_1 is pin 14 current when $V_{CBOOT} = V_{CBOOT1}$, I_2 when $V_{CBOOT} = V_{CBOOT2}$

3.3 Timing diagram

Figure 3. Input/output timing diagram



4 Bootstrap driver

A bootstrap circuitry is needed to supply the high voltage section. This function is normally accomplished by a high voltage fast recovery diode (*Figure 4 a*). In the L6386AD a patented integrated structure replaces the external diode. It is realized by a high voltage DMOS, driven synchronously with the low side driver (LVG), with in series a diode, as shown in *Figure 4 b*. An internal charge pump (*Figure 4 b*) provides the DMOS driving voltage. The diode connected in series to the DMOS has been added to avoid undesirable turn on of it.

4.1 C_{BOOT} selection and charging

To choose the proper C_{BOOT} value the external MOS can be seen as an equivalent capacitor. This capacitor C_{EXT} is related to the MOS total gate charge:

$$C_{EXT} = \frac{Q_{gate}}{V_{gate}}$$

The ratio between the capacitors C_{EXT} and C_{BOOT} is proportional to the cyclical voltage loss. It has to be:

$$C_{BOOT} \gg C_{EXT}$$

e.g.: if Q_{gate} is 30 nC and V_{gate} is 10 V, C_{EXT} is 3 nF. With C_{BOOT} = 100 nF the drop would be 300 mV.

If HVG has to be supplied for a long time, the C_{BOOT} selection has to take into account also the leakage losses.

e.g.: HVG steady state consumption is lower than 200 μA, so if HVG T_{ON} is 5ms, C_{BOOT} has to supply 1 μC to C_{EXT}. This charge on a 1 μF capacitor means a voltage drop of 1 V.

The internal bootstrap driver gives great advantages: the external fast recovery diode can be avoided (it usually has great leakage current).

This structure can work only if V_{OUT} is close to GND (or lower) and in the meanwhile the LVG is on. The charging time (T_{charge}) of the C_{BOOT} is the time in which both conditions are fulfilled and it has to be long enough to charge the capacitor.

The bootstrap driver introduces a voltage drop due to the DMOS R_{DS(on)} (typical value: 125 Ω). At low frequency this drop can be neglected. Anyway increasing the frequency it must be taken in to account.

The following equation is useful to compute the drop on the bootstrap DMOS:

$$V_{drop} = I_{charge} R_{dson} \rightarrow V_{drop} = \frac{Q_{gate}}{T_{charge}} R_{dson}$$

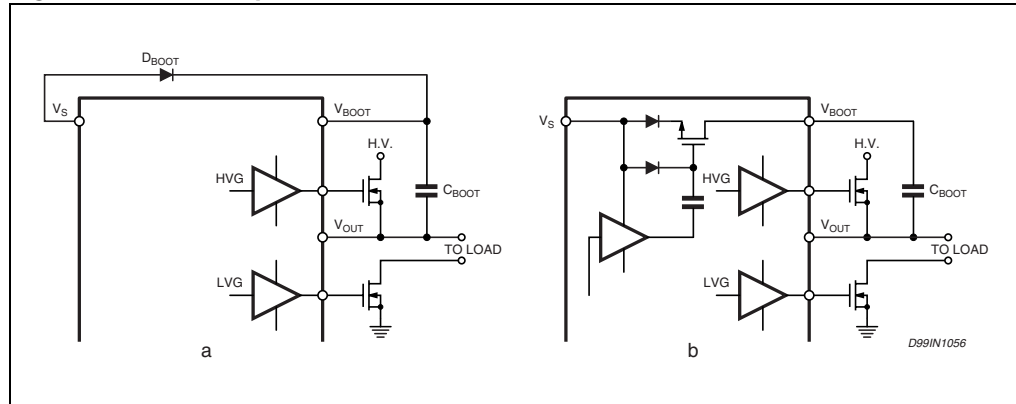
where Q_{gate} is the gate charge of the external power MOS, R_{dson} is the on resistance of the bootstrap DMOS, and T_{charge} is the charging time of the bootstrap capacitor.

For example: using a power MOS with a total gate charge of 30 nC the drop on the bootstrap DMOS is about 1 V, if the T_{charge} is 5 μs . In fact:

$$V_{\text{drop}} = \frac{30\text{nC}}{5\mu\text{s}} \cdot 125\Omega = 0.8\text{V}$$

V_{drop} has to be taken into account when the voltage drop on C_{BOOT} is calculated: if this drop is too high, or the circuit topology doesn't allow a sufficient charging time, an external diode can be used.

Figure 4. Bootstrap driver



5 Typical characteristic

Figure 5. Typical rise and fall times vs load capacitance

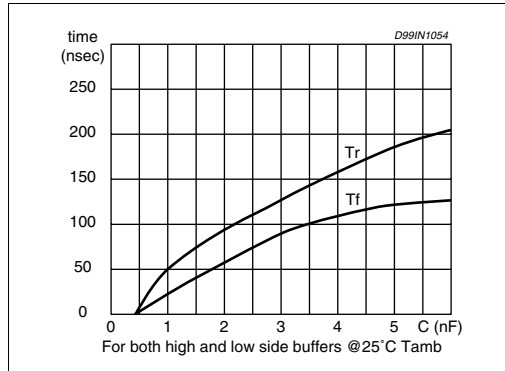


Figure 6. Quiescent current vs supply voltage

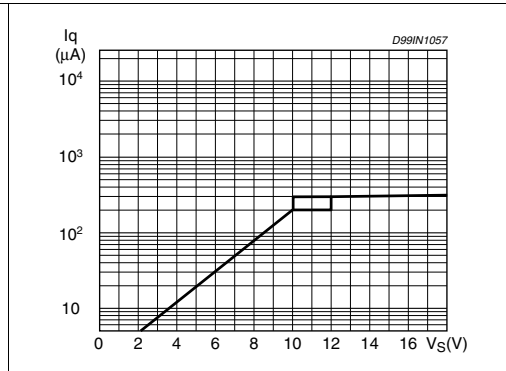


Figure 7. Turn on time vs temperature

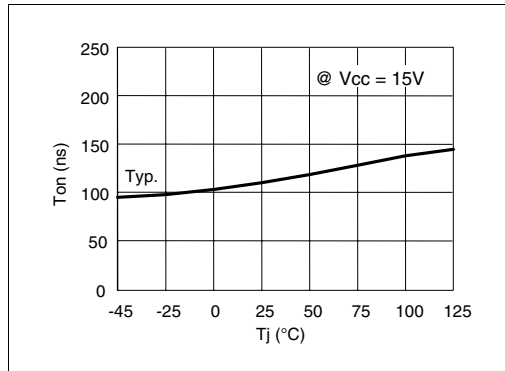


Figure 8. VBOOT UV turn on threshold vs temperature

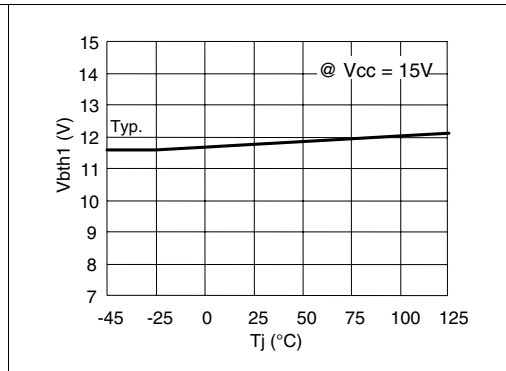


Figure 9. Turn Off time vs temperature

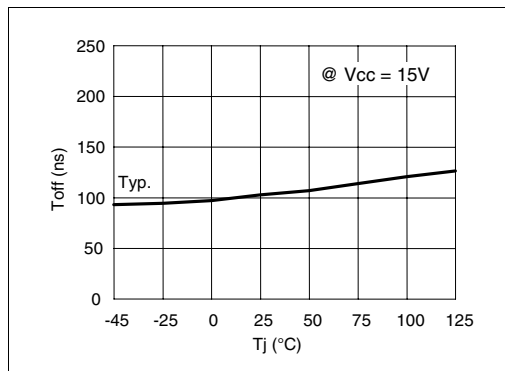


Figure 10. VBOOT UV turn off threshold vs temperature

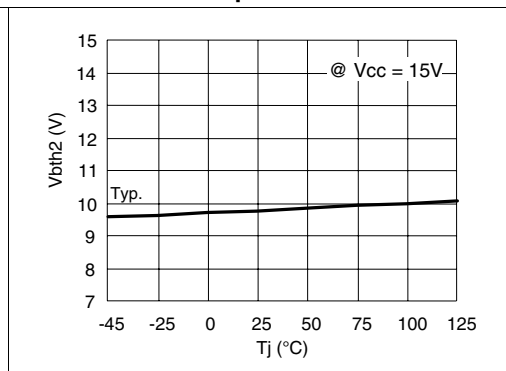


Figure 11. Shutdown time vs temperature

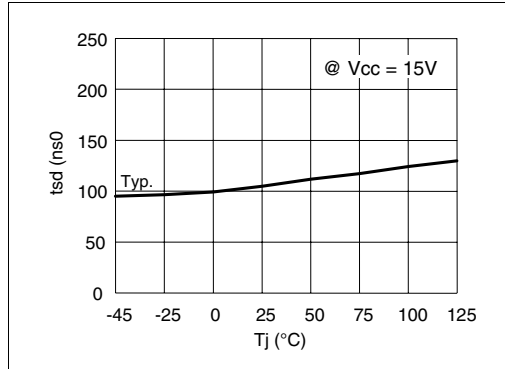


Figure 12. V_{BOOT} UV hysteresis

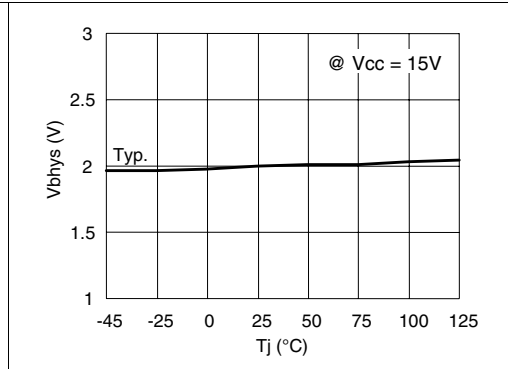


Figure 13. V_{CC} UV turn on threshold vs temperature

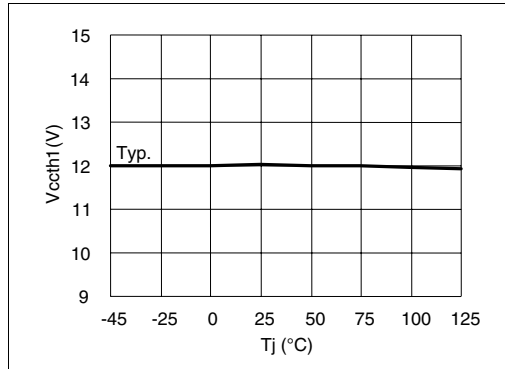


Figure 14. Output source current vs temperature

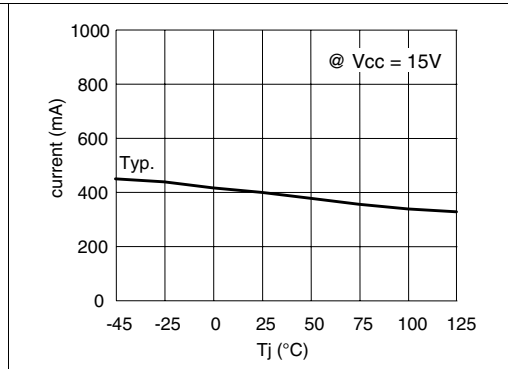


Figure 15. V_{CC} UV turn off threshold vs temperature

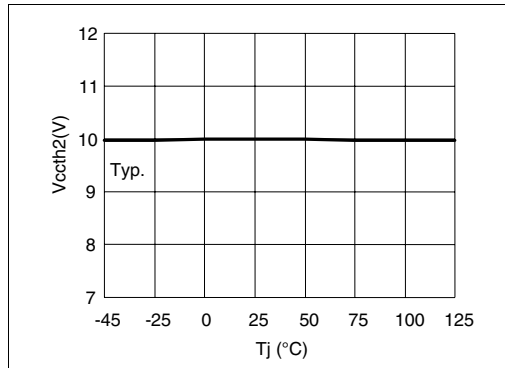


Figure 16. Output sink current vs temperature

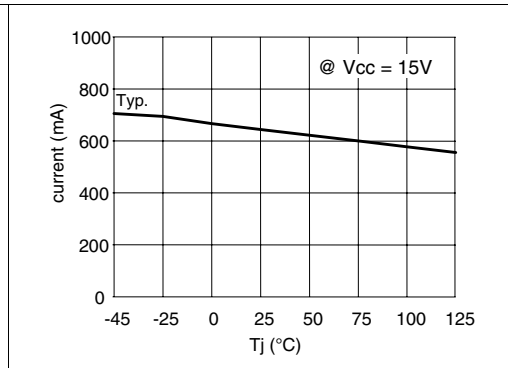
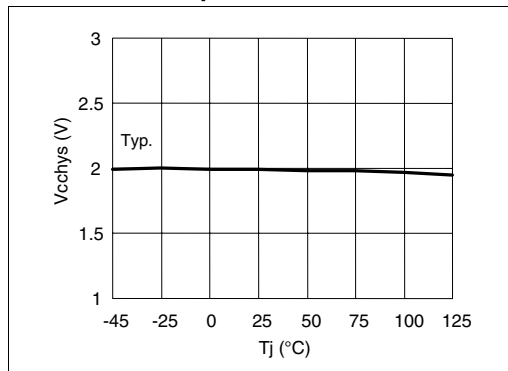


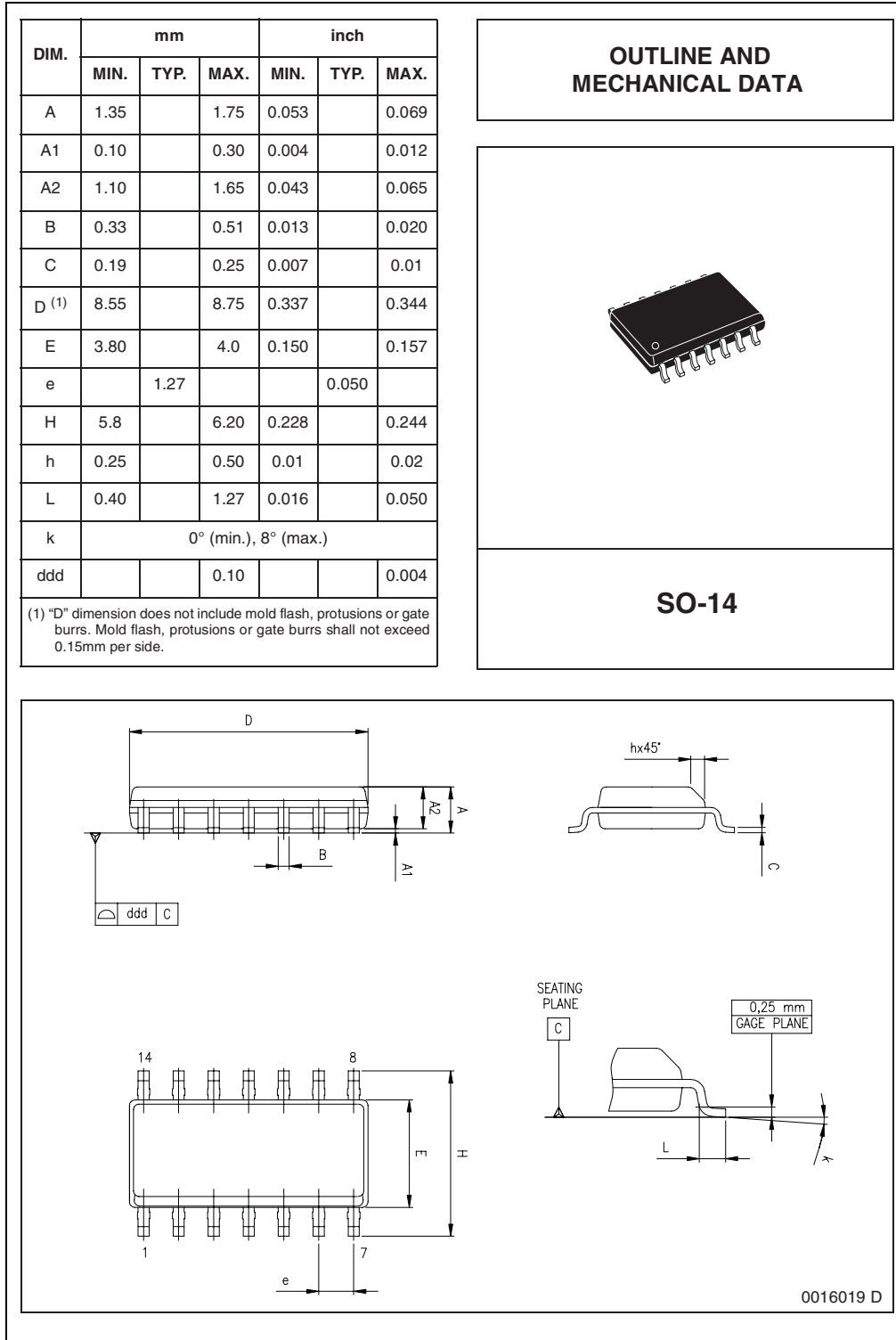
Figure 17. V_{CC} UV hysteresis vs temperature



6 Package mechanical data

In order to meet environmental requirements, ST offers these devices in ECOPACK® packages. These packages have a lead-free second level interconnect . The category of second level interconnect is marked on the package and on the inner box label, in compliance with JEDEC Standard JESD97. The maximum ratings related to soldering conditions are also marked on the inner box label. ECOPACK is an ST trademark. ECOPACK specifications are available at: www.st.com

Figure 18. SO-14 mechanical data and package dimensions



7 Order codes

Table 7. Order codes

Order codes	Package	Packaging
L6386AD	SO-14	Tube
L6386AD013TR	SO-14	Tape and reel

8 Revision history

Table 8. Document revision history

Date	Revision	Changes
14-Jul-2008	1	First release

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