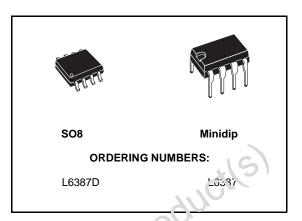


HIGH-VOLTAGE HIGH AND LOW SIDE DRIVER

- HIGH VOLTAGE RAIL UP TO 600 V
- dV/dt IMMUNITY +- 50 V/nsec IN FULL TEM-PERATURE RANGE
- DRIVER CURRENT CAPABILITY: 400 mA SOURCE, 650 mA SINK
- SWITCHING TIMES 50/30 nsec RISE/FALL WITH 1nF LOAD
- CMOS/TTL SCHMITT TRIGGER INPUTS WITH HYSTERESIS AND PULL DOWN
- INTERNAL BOOTSTRAP DIODE
- OUTPUTS IN PHASE WITH INPUTS

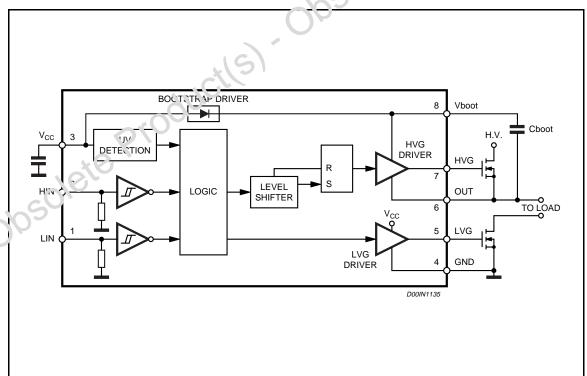
DESCRIPTION

The L6387 is an high-voltage device, manufactured with the BCD"OFF-LINE" technology. It has a Driver structure that enables to drive independent referenced N Channel Power MOS or



IGBT. The Upper (Floating) Section is enabled to work with voltage Kail up to 600V. The Logic Inputs are CMOS/TTL compatible for ease of interfacing with controlling devices.

BLOCK DIAGRAM



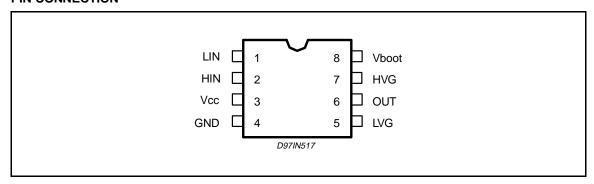
May 2001 1/9

ABSOLUTE MAXIMUM RATINGS

Symbol	Parameter	Value	Unit
Vout	Output Voltage	-3 to Vboot - 18	V
Vcc	Supply Voltage	- 0.3 to +18	V
Vboot	Floating Supply Voltage	- 1 to 618	V
Vhvg	Upper Gate Output Voltage	- 1 to Vboot	V
VIvg	Lower Gate Output Voltage	-0.3 to Vcc +0.3	V
Vi	Logic Input Voltage	-0.3 to Vcc +0.3	V
dVout/dt	Allowed Output Slew Rate	50	V/ns
Ptot	Total Power Dissipation (Tj = 85 °C)	750	mW
Tj	Junction Temperature	150	°C
Ts	Storage Temperature	-50 to 150	°C

Note: ESD immunity for pins 6, 7 and 8 is guaranteed up to 900V (Human Body Model)

PIN CONNECTION



THERMAL DATA

Symbol	Parameter	S08	Minidip	Unit
R _{th j-amb}	Thermal Resistance Junction to Ambient	150	100	°C/W

PIN DESCRIPTION

N.	Name	Type	Function			
1	LIN	I	Lower Driver Logic Input			
2	HIN	I	Upper Driver Logic Input			
3	Vcc	I	Low Voltage Power Supply			
4	GND		Ground			
5	LVG (*)	0	Low Side Driver Output			
6	VOUT	0	Upper Driver Floating Reference			
7	HVG (*)	0	High Side Driver Output			
8	Vboot		Bootstrap Supply Voltage			

^(*) The circuit guarantees 0.3V maximum on the pin (@ I_{sink} = 10mA). This allows to omit the "bleeder" resistor connected between the gate and the source of the external MOSFET normally used to hold the pin low.

4

RECOMMENDED OPERATING CONDITIONS

Symbol	Pin	Parameter	Test Condition	Min.	Тур.	Max.	Unit
Vout	6	Output Voltage		Note 1		580	V
Vboot- Vout	8	Floating Supply Voltage		Note 1		17	V
fsw		Switching Frequency	HVG,LVG load CL = 1nF			400	kHz
Vcc	2	Supply Voltage				17	V
Tj		Junction Temperature		-45		125	°C

Note 1: If the condition Vboot - Vout < 18V is guaranteed, Vout can range from -3 to 580V.

ELECTRICAL CHARACTERISTICS AC Operation (Vcc = 15V; Tj = 25°C)

Symbol	Pin	Parameter	Test Condition	Min.	Тур.	Max.	Unit
ton	1 vs 7	High/Low Side Driver Turn-On Propagation Delay	Vout = 0V		110		ns
toff	2 vs 5	High/Low Side Driver Turn-Off Propagation Delay	Vout = 600V		105		ns
tr	7,5	Rise Time	CL = 1000pF		50		ns
tf	7,5	Fall Time	CL = 1000pF		30		ns

DC OPERATION (Vcc = 15V; Tj = 25°C)

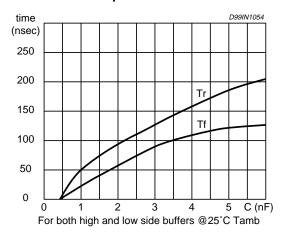
Symbol	Pin	Parameter	Test Condition	Min.	Тур.	Max.	Unit
Low Sup	oply Vo	ltage Section					
Vcc	3	Supply Voltage				17	V
Vccth1		Vcc UV Turn On Threshold		5.5	6	6.5	V
Vccth2		Vcc UV Turn Off Threshold		5	5.5	6	V
Vcchys		Vcc UV Hysteresis			0.5		V
Iqccu		Undervoltage Quiescent Supply Current	Vcc ≤ 9V		150	220	μΑ
Iqcc		Quiescent Current	Vcc = 15V		250	320	μΑ
R _{dson}		Bootstrap Driver on Resistance (*)	Vcc ≥ 12.5V		125		Ω
Bootstra	apped s	supply Voltage Section		•			
VBS	8	Bootstrap Supply Voltage				17	V
IQBS		VBS Quiescent Current	HVG ON			200	μΑ
ILK		High Voltage Leakage Current	VS = VB = 600V			10	μΑ
High/Lo	w Side	Driver					
Iso	5,7	Source Short Circuit Current	$VIN = Vih (tp < 10\mu s)$	300	400		mA
Isi		Sink Short Circuit Current	VIN = Vil (tp < 10μs)	450	650		mA
Logic In	puts						
Vil	2,3	Low Level Logic Threshold Voltage				1.5	V
Vih		High Level Logic Threshold Voltage		3.6			V
lih		High Level Logic Input Current	VIN = 15V		50	70	μΑ
lil		Low Level Logic Input Current	VIN = 0V			1	μΑ

^(*) R_{DSON} is tested in the following way: $R_{DSON} = \frac{(V_{CC} - V_{CBOOT1}) - (V_{CC} - V_{CBOOT2})}{I_1(V_{CC}, V_{CBOOT1}) - I_2(V_{CC}, V_{CBOOT2})}$

where I_1 is pin 8 current when V_{CBOOT} = $V_{\text{CBOOT1}},\,I_2$ when V_{CBOOT} = $V_{\text{CBOOT2}}.$



Figure 1. Typical Rise and Fall Times vs. Load Capacitance



Input Logic

L6387 Input Logic is V_{CC} (17V) compatible. An interlocking features is offered (see truth table below) to avoid undesired simultaneous turn ON of both Power Switches driven.

Table 1.

Input	HIN	0	0	1	1
	LIN	0	1	0	1
Output	HVG	0	0	1	0
	LVG	0	1	0	0

BOOTSTRAP DRIVER

A bootstrap circuitry is needed to supply the high voltage section. This function is normally accomplished by a high voltage fast recovery diode (fig. 3a). In the L6387 a patented integrated structure replaces the external diode. It is realized by a high voltage DMOS, driven synchronously with the low side driver (LVG), with in series a diode, as shown in fig. 3b

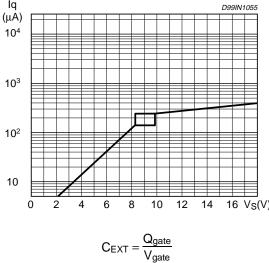
An internal charge pump (fig. 3b) provides the DMOS driving voltage.

The diode connected in series to the DMOS has been added to avoid undesirable turn on of it.

CBOOT selection and charging:

To choose the proper CBOOT value the external MOS can be seen as an equivalent capacitor. This capacitor CEXT is related to the MOS total gate charge:

Figure 2. Quiescent Current vs. Supply Voltage



The ratio between the capacitors CEXT and CBOOT is proportional to the cyclical voltage loss.

It has to be:

e.g.: if Q_{gate} is 30nC and V_{gate} is 10V, C_{EXT} is 3nF. With $C_{BOOT} = 100nF$ the drop would be 300mV.

If HVG has to be supplied for a long time, the CBOOT selection has to take into account also the leakage losses.

e.g.: HVG steady state consumption is lower than 200μA, so if HVG ToN is 5ms, CBOOT has to supply 1μC to C_{EXT}. This charge on a 1μF capacitor means a voltage drop of 1V.

The internal bootstrap driver gives great advantages: the external fast recovery diode can be avoided (it usually has great leakage current). This structure can work only if Vout is close to GND (or lower) and in the meanwhile the LVG is on. The charging time (T_{charge}) of the C_{BOOT} is the time in which both conditions are fulfilled and it has to be long enough to charge the capacitor.

The bootstrap driver introduces a voltage drop due to the DMOS RDSON (typical value: 125 Ohm). At low frequency this drop can be neglected. Anyway increasing the frequency it must be taken in to account.

The following equation is useful to compute the

Downloaded from Elcodis.com electronic components distributor

drop on the bootstrap DMOS:

$$V_{drop} = I_{charge} R_{dson} \rightarrow V_{drop} = \frac{Q_{gate}}{T_{charge}} R_{dson}$$

where Q_{gate} is the gate charge of the external power MOS, R_{dson} is the on resistance of the bootstrap DMOS, and T_{charge} is the charging time of the bootstrap capacitor.

For example: using a power MOS with a total gate charge of 30nC the drop on the bootstrap

DMOS is about 1V, if the T_{charge} is $5\mu s$. In fact:

$$V_{drop} = \frac{30nC}{5\mu s} \cdot 125\Omega \sim 0.8V$$

 V_{drop} has to be taken into account when the voltage drop on C_{BOOT} is calculated: if this drop is too high, or the circuit topology doesn't allow a sufficient charging time, an external diode can be used.

Figure 3. Bootstrap Driver.

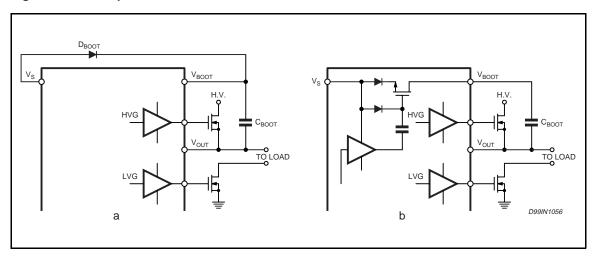


Figure 4. Turn On Time vs. Temperature

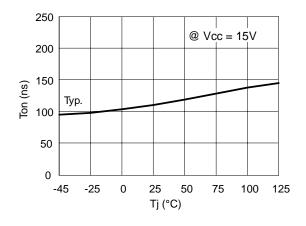
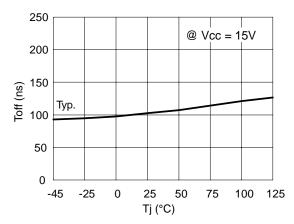


Figure 5. Turn Off Time vs. Temperature



57

Figure 6. Output Source Current vs. Temperature

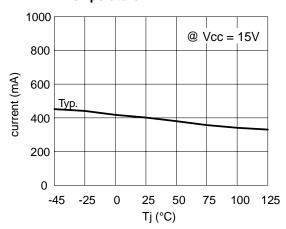
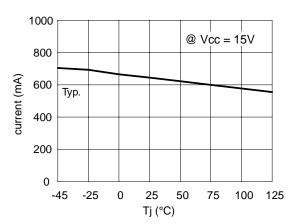


Figure 7. Output Sink Current vs. Temperature

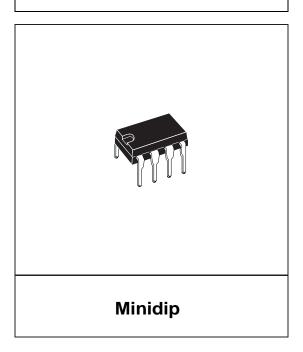


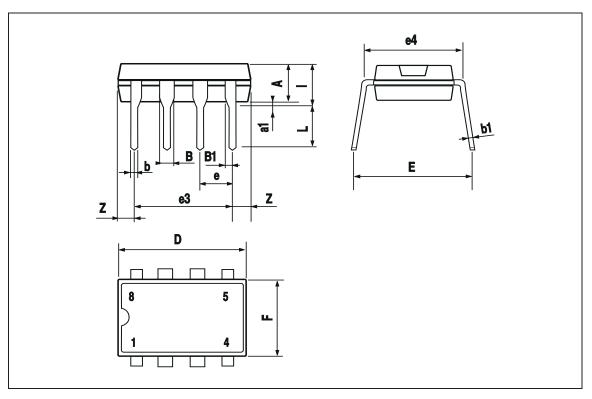
6/9

7/9

DIM.		mm		inch			
Diwi.	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	
Α		3.32			0.131		
a1	0.51			0.020			
В	1.15		1.65	0.045		0.065	
b	0.356		0.55	0.014		0.022	
b1	0.204		0.304	0.008		0.012	
D			10.92			0.430	
Е	7.95		9.75	0.313		0.384	
е		2.54			0.100		
е3		7.62			0.300		
e4		7.62			0.300		
F			6.6			0.260	
I			5.08			0.200	
L	3.18		3.81	0.125		0.150	
Z			1.52			0.060	

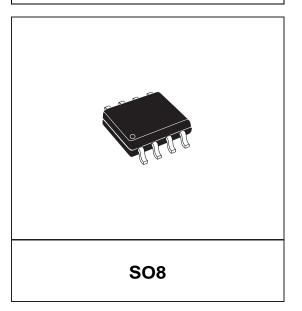
OUTLINE AND MECHANICAL DATA



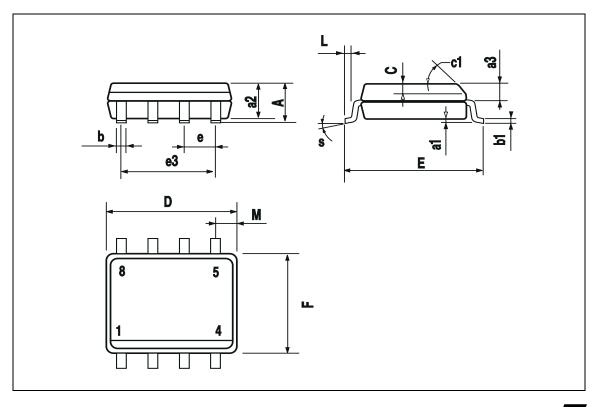


DIM.		mm			inch	
Dilvi.	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.
Α			1.75			0.069
a1	0.1		0.25	0.004		0.010
a2			1.65			0.065
а3	0.65		0.85	0.026		0.033
b	0.35		0.48	0.014		0.019
b1	0.19		0.25	0.007		0.010
С	0.25		0.5	0.010		0.020
c1			45° ((typ.)		
D (1)	4.8		5.0	0.189		0.197
Е	5.8		6.2	0.228		0.244
е		1.27			0.050	
еЗ		3.81			0.150	
F (1)	3.8		4.0	0.15		0.157
L	0.4		1.27	0.016		0.050
М			0.6			0.024
S			8° (n	nax.)		

OUTLINE AND MECHANICAL DATA



⁽¹⁾ D and F do not include mold flash or protrusions. Mold flash or potrusions shall not exceed 0.15mm (.006inch).



8/9

Information furnished is believed to be accurate and reliable. However, STMicroelectronics assumes no responsibility for the consequences of use of such information nor for any infringement of patents or other rights of third parties which may result from its use. No license is granted by implication or otherwise under any patent or patent rights of STMicroelectronics. Specification mentioned in this publication are subject to change without notice. This publication supersedes and replaces all information previously supplied. STMicroelectronics products are not authorized for use as critical components in life support devices or systems without express written approval of STMicroelectronics.

The ST logo is a registered trademark of STMicroelectronics

© 2001 STMicroelectronics - Printed in Italy - All Rights Reserved

STMicroelectronics GROUP OF COMPANIES

Australia - Brazil - China - Finland - France - Germany - Hong Kong - India - Italy - Japan - Malaysia - Malta - Morocco - Singapore - Spain - Sweden - Switzerland - United Kingdom - U.S.A.

http://www.st.com



9/9