# EVALUATION KIT AVAILABLE

## **1.2A White LED Regulating Charge Pump for Camera Flashes and Movie Lights**

### **General Description**

The MAX1577Y/MAX1577Z charge pumps drive white LEDs, including camera strobes, with regulated current up to 1.2A (guaranteed to 800mA). The very low openloop output resistance allows high flash brightness, even from a low battery input voltage. The adaptive 1x/2x regulating charge-pump operation provides high efficiency (up to 92%) while in movie mode or backlighting. Fast (1MHz) switching allows the use of tiny external components.

The MAX1577Y/MAX1577Z use an external resistor to set the full-scale LED current. Two enable inputs (EN1 and EN2) provide simple on/off control and LED current of 20%, 33%, or 100% full scale. If there is less than 24mV across the external resistor, the output voltage is regulated. The MAX1577Z has fixed 5.1V output-voltage regulation, while the MAX1577Y has logic-controlled output voltage of 3.4V, 4.6V, or 5.1V for backlighting, keyboard, and RGB applications.

The MAX1577Y/MAX1577Z are available in an 8-pin, 3mm x 3mm TDFN package (0.8mm max height).

### Applications

White LED Flashes, Strobes, and Movie Lights LCD Backlighting

Camera Phones, Cell Phones, and Smart Phones PDAs, Digital Cameras, and Camcorders

### Features

Up to 1.2A Output Drive Capability (800mA) Guaranteed)

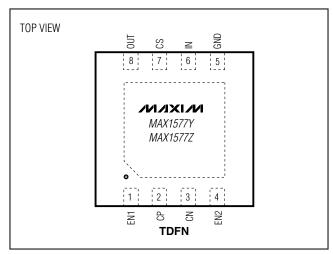
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- Up to 92% PLED / PBATT Efficiency for Movie Mode
- Flexible Brightness Control Two-Bit Logic for 100%, 33%, 20%, and Off PWM for 20% to 100%
- 3% Current Regulation Over Line and Temperature
- Current or Voltage Regulation for Flash and Backlighting (MAX1577Y)
- Current Regulation for Flash (MAX1577Z)
- Low Ripple and EMI
- 2.7V to 5.5V Supply Voltage Range
- Soft-Start Limits Inrush Current
- Output-Voltage Regulation Mode
- Thermal-Shutdown Protection
- 8-Pin 3mm x 3mm TDFN Package

### **Ordering Information**

PART	TEMP RANGE	PIN- PACKAGE	TOP MARK	
MAX1577YETA	-40°C to +85°C	8 TDFN (T833-2)	AMG	
MAX1577ZETA	-40°C to +85°C	8 TDFN (T833-2)	AMQ	

## **Pin Configuration**



<sup>1</sup>µF OR 2.2µF CP CN UP TO 1.2A 2.7V TO 5.5V 5.3V OVP IN OUT  $^{/}$ 10µF GND /VI/IXI/VI MAX1577Z V<sub>OUT</sub> - (300mV OR 100mV OR 60mV) 100%, CS EN1 33%, 20%, AND SHUTDOWN FN2

## Typical Operating Circuit

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For pricing, delivery, and ordering information, please contact Maxim/Dallas Direct! at 1-888-629-4642, or visit Maxim's website at www.maxim-ic.com.

### **ABSOLUTE MAXIMUM RATINGS**

IN, OUT, EN1, EN2 to GND	0.3V to +6.0V
OUT to GND During Shutdown	0.3V to (VIN + 0.3V)
OUT to CS	0.6V to +0.6V
CS to GND	0.3V to (Vout + 0.3V)
CN to GND	
CP to GND0.3V to the Greate	r of ( $V_{OUT}$ + 1V) or ( $V_{IN}$ + 1V)
OUT Short Circuit to GND	Continuous

Continuous Power Dissipation (T<sub>A</sub> = +70°C) 8-Pin TDFN (derate 18.2mW/°C above +70°C) ......1455mW Operating Temperature Range .....40°C to +85°C Junction Temperature ......+150°C Storage Temperature Range ......65°C to +150°C Lead Temperature (soldering, 10s) .....+300°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## **ELECTRICAL CHARACTERISTICS**

 $(V_{IN} = 3.6V, \text{ circuit of Figure 2, EN1} = EN2 = IN, R_{CS} = 60\Omega, C_{IN} = 4.7\mu\text{F}, C1 = 2.2\mu\text{F}, C_{OUT} = 10\mu\text{F}, T_A = -40^{\circ}\text{C}$  to +85°C, unless otherwise noted. Typical values are at T<sub>A</sub> = +25°C.) (Note 1)

PARAMETER	CONDITIONS			ТҮР	MAX	UNITS
IN Operating Voltage					5.5	V
Undervoltage-Lockout Threshold	V <sub>IN</sub> rising			2.45	2.6	V
Undervoltage-Lockout Hysteresis				35		mV
Supply Current	1MHz switching, no load			5		mA
Supply Current	1x LDO mode, no switching			0.225	0.3	ШA
Shutdown Supply Current	EN1 = EN2 = GND	$T_A = +25^{\circ}C$		0.01	1	
Shutdown Supply Current	EINT = EINZ = GIND	$T_A = +85^{\circ}C$		0.2		μA
Soft-Start Time				0.5		ms
OUT Overvoltage Protection				5.3		V
	MAX1577Z, $T_A = +25^{\circ}C$ , I <sub>OUT</sub> = 100mA	EN1 or EN2 = IN	4.95	5.10	5.25	V
OUT Voltage Regulation Mode	MAX1577Y, T <sub>A</sub> = +25°C, I <sub>OUT</sub> = 100mA	EN1 = GND, EN2 = IN	3.17	3.40	3.60	
		EN1 = IN, EN2 = GND	4.40	4.60	4.80	
		EN1 = EN2 = IN	4.95	5.10	5.25	
CS Voltage-Mode Threshold (Vout - Vcs)				24		mV
	EN1 = GND, EN2 = IN			60		
CS Regulation Voltage	EN1 = IN, EN2 = GND			100		mV
(Vout - Vcs)	EN1 = EN2 = IN			300		
	$T_A = +25^{\circ}C$		-2	±0.5	+2	0/
CS Threshold Accuracy	$V_{IN} = 2.7V$ to 5.5V, TA = -40°C to +85°C		-3		+3	%
Maximum OUT Current		$V_{IN} = 3.1V$	600	900		mA
	$V_{\text{LED}} = 4.1\text{V}, \text{ EN1} = \text{EN2} = \text{IN} \qquad \qquad$		800	1200		mA
	1x LDO mode			0.6		Ω
Open-Loop OUT Resistance	2x mode, V <sub>IN</sub> = 3.4V			2	3	52
OUT Internal Pulldown in Shutdown	EN1 = EN2 = GND			5		kΩ
Switching Frequency				1.0	1.25	MHz

## **ELECTRICAL CHARACTERISTICS (continued)**

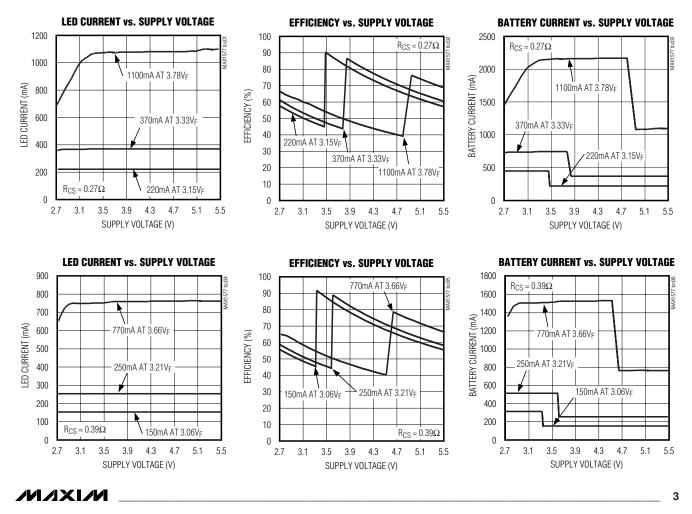
 $(V_{IN} = 3.6V, \text{ circuit of Figure 2, EN1} = EN2 = IN, R_{CS} = 60\Omega, C_{IN} = 4.7\mu\text{F}, C1 = 2.2\mu\text{F}, C_{OUT} = 10\mu\text{F}, T_A = -40^{\circ}\text{C}$  to +85°C, unless otherwise noted. Typical values are at  $T_A = +25^{\circ}\text{C}$ .) (Note 1)

PARAMETER	CONDITIONS			ТҮР	MAX	UNITS
	$V_{IN} = 4.2V$ to 5.5V		1.6			V
EN_ Logic-High Voltage	$V_{IN} = 2.7V$ to 4.2V		1.4			
EN_ Logic-Low Voltage	$V_{IN} = 2.7V$ to 5.5V			0.4	V	
Logic Input Current		$T_A = +25^{\circ}C$		0.02	1	
	$V_{EN} = 0V \text{ or } 5.5V$ $T_A = +85^{\circ}C$			0.02		μΑ
Thermal-Shutdown Threshold	T <sub>J</sub> rising			+160		°C
Thermal-Shutdown Hysteresis				20		°C

**Note 1:** Limits are 100% production tested at  $T_A = +25^{\circ}C$ . Limits over the operating temperature range are guaranteed by design.

## **Typical Operating Characteristics**

 $(V_{IN} = 3.6V, \text{ circuit of Figure 2}, R_{CS} = 0.27\Omega, Luxeon LXCL-PWF1 LED, T_A = +25^{\circ}C, unless otherwise noted.)$ 



 $(V_{IN} = 3.6V, \text{ circuit of Figure 2}, R_{CS} = 0.27\Omega, Luxeon LXCL-PWF1 LED, T_A = +25^{\circ}C, unless otherwise noted.)$ 

**Typical Operating Characteristics (continued)** 

/N/IXI/N



#### **TYPICAL OPERATING WAVEFORMS** LED CURRENT vs. R<sub>CS</sub> **LED CURRENT vs. TEMPERATURE** IN 1x MODE MAX1577 toc09 225 1.2 1.0 220 50mV/div EN1 AND EN2 HIGH $V_{IN}$ (AC-COUPLED) LED CURRENT (mA) LED CURRENT (A) 0.8 215 50mV/div 0.6 VOUT (AC-COUPLED) 210 EN1 HIGH, EN2 LOW 0.4 EN1 LOW, EN2 HIGH 50mV/div $V_{\text{CS}}$ (AC-COUPLED) 205 <sup>-</sup> R<sub>CS</sub> = 0.27Ω EN1 LOW EN2 HIGH 0.2 200 0 -15 85 -40 10 35 60 0.2 0.4 0.6 0.8 1.0 400ns/div TEMPERATURE (°C) $R_{CS}(\Omega)$ TYPICAL OPERATING WAVEFORMS IN 2x MODE STARTUP AND SHUTDOWN RESPONSE MAX1577 toc10 V<sub>EN1</sub>, 2V/div 50mV/div $V_{\text{IN}}$ V<sub>EN2</sub> (AC-COUPLED) 50mV/div VOUT 2V/div VLED (AC-COUPLED) 0V 50mV/div Vcs (AC-COUPLED) LED 500mA/div 0mA 400ns/div 400µs/div **EFFICIENCY vs. SUPPLY VOLTAGE** LED CURRENT vs. PWM DUTY CYCLE WITH PWM DIMMING **DIMMING RESPONSE** MAX1577 toc14 1200 80 70 2V/div 1000 V<sub>EN1</sub> 25% 60 LED CURRENT (mA) 800 EFFICIENCY (%) 50 2V/div $V_{LED}$ 40 600 50% 30 0V 75% 400 20 $R_{CS} = 0.27 \Omega$ $R_{CS} = 0.27\Omega$ 200 ILED EN1 PWM EN1 PWM 500mA/div 10 EN2 HIGH EN2 HIGH 0 0 ΩmA 3.9 4.3 60 100 3.1 3.5 4.7 5.1 5.5 0 20 40 80 2.7 100µs/div DUTY CYCLE (%) SUPPLY VOLTAGE (V)

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## \_Pin Description

PIN	NAME	FUNCTION				
1	EN1	Control Logic Input (see Table 1)				
2	СР	Transfer Capacitor Positive Connection. Connect a $1\mu$ F capacitor (or 2.2 $\mu$ F for I <sub>OUT</sub> > 600mA) from CP to CN.				
3	CN	Transfer Capacitor Negative Connection. Connect a 1 $\mu$ F capacitor (or 2.2 $\mu$ F for I <sub>OUT</sub> > 600mA) from CP to CN.				
4	EN2	Control Logic Input (see Table 1)				
5	GND	Ground				
6	IN	Input Supply Voltage. Connect IN to a battery or supply voltage from 2.7V to 5.5V. Connect a $4.7\mu F$ ceramic capacitor from IN to GND.				
7	CS	Current-Sense Input. Connect a current-sense resistor from OUT to CS to set the output current. Connect the anodes of the LEDs to CS and the cathodes to GND.				
8	OUT	Output. Connect a 10 $\mu$ F ceramic capacitor from OUT to GND. In shutdown, OUT is pulled to GND with an internal 5k $\Omega$ resistor.				
—	EP	Exposed Pad. Connect the exposed pad to GND directly under the IC.				

## Table 1. Enable, Dimming, Voltage Mode, and Overvoltage Protection

EN1 EN2		DIMMING LEVEL	CS THRESHOLD	VOLTAGE MODE		OVP		
ENI			(Vout - Vcs)	MAX1577Z	MAX1577Y	MAX1577Z	MAX1577Y	
0	0	Shutdown	_	—	—	—		
0	1	20% Current	60mV	5.10V	3.40V	5.30V	5.15V	
1	0	33% Current	100mV	5.10V	4.60V	5.30V	5.17V	
1	1	100% Current	300mV	5.10V	5.10V	5.30V	5.30V	

Note: A "0" indicates EN\_ is logic low or connected to GND. A "1" indicates EN\_ is logic high or connected to IN.

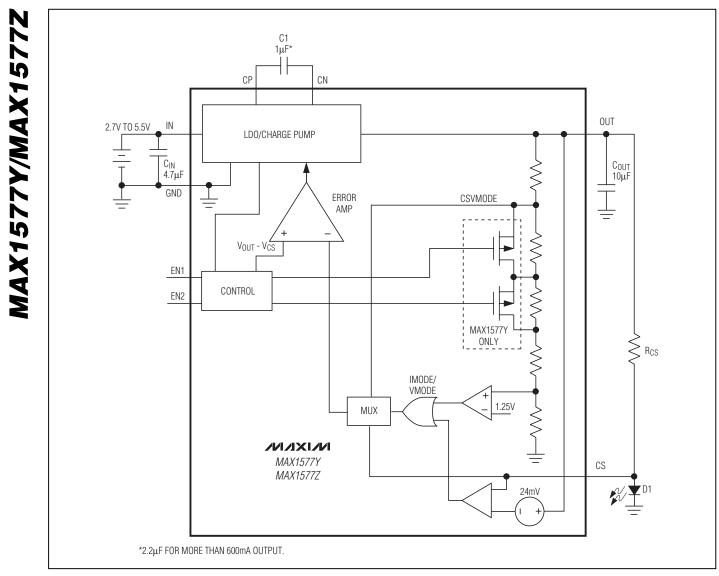


Figure 1. Functional Diagram

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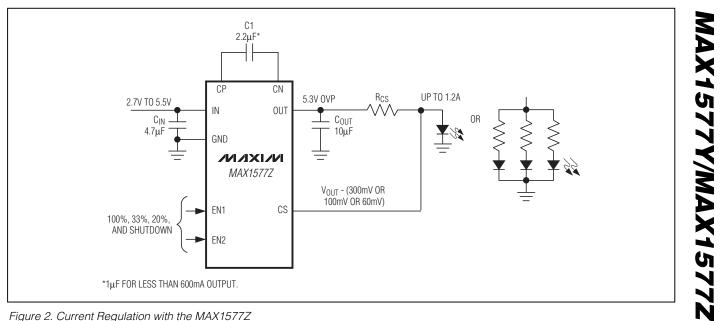


Figure 2. Current Regulation with the MAX1577Z

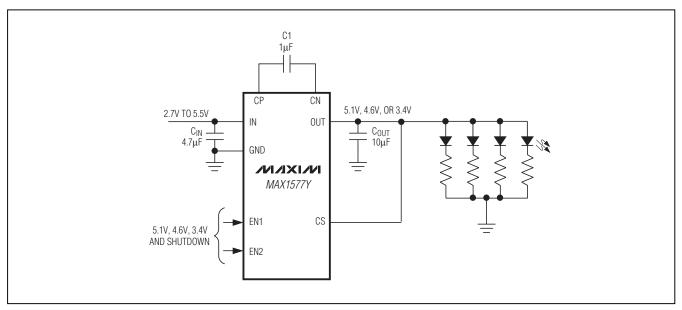


Figure 3. Voltage Regulation with the MAX1577Y

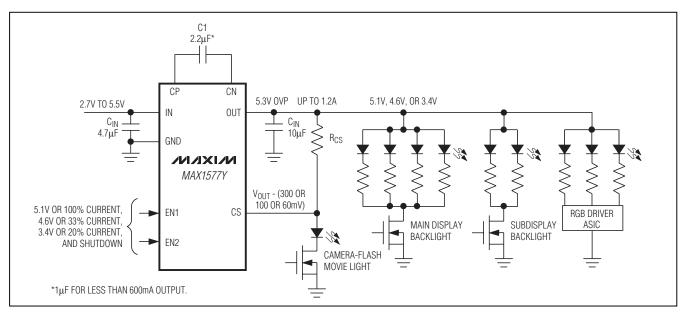


Figure 4. Flash and Backlighting with the MAX1577Y

### \_Detailed Description

#### 1x LDO to 2x Mode Switchover

When V<sub>IN</sub> is greater than V<sub>OUT</sub>, the MAX1577Y/ MAX1577Z are in 1x LDO mode and regulate the current through R<sub>CS</sub>. As V<sub>IN</sub> decreases and the LDO approaches dropout, the 2x charge pump starts boosting the output to maintain the regulation current. When V<sub>IN</sub> rises enough above V<sub>OUT</sub> to maintain LDO regulation, the charge pump stops and the MAX1577Y/MAX1577Z switch back to the more efficient 1x LDO mode.

#### Soft-Start

The MAX1577Y/MAX1577Z include soft-start circuitry to limit inrush current at turn-on. When starting up with an output voltage that is not near the input voltage, the output capacitor is charged directly from the input with a ramped current source (with no charge-pump action) until the output voltage approaches the input voltage. If current-mode or voltage-mode regulation is met, softstart is terminated and normal operation begins in 1x mode. Otherwise, after 500µs, the 2x charge-pump softstart operates until the voltage across the sense resistor reaches the current regulation threshold, the output voltage is in regulation, or 500µs has elapsed, whichever occurs first. If the output is shorted to ground, soft-start is repeated until the output exceeds 2.3V.

#### **Dimming and Shutdown**

The MAX1577Y/MAX1577Z use two inputs (EN1 and EN2) to select between three brightness levels and lowpower shutdown mode (see Table 1). By regulating the voltage across the current-sense resistor (V<sub>OUT</sub> - V<sub>CS</sub>) to one of three levels, the MAX1577Y/MAX1577Z control the output current, thus providing the three-level dimming. In shutdown mode, the IC is turned off, reducing the supply current to 0.01 $\mu$ A (typ). When the MAX1577Y/MAX1577Z are shut down, C1 is discharged and the output is pulled to GND through an internal 5k $\Omega$  resistor.

M/IXI/M

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#### **Overvoltage Protection**

Output overvoltage protection (OVP) protects the IC and other components in case the output is open circuit. The MAX1577Z always limits the output voltage to 5.3V. With the MAX1577Y, the output OVP voltage varies slightly depending on the state of the EN\_ inputs (see Table 1).

#### **Output-Voltage Regulation Mode**

When the voltage across the sense resistor (V<sub>OUT</sub> - V<sub>CS</sub>) is less than 24mV, the MAX1577Y/MAX1577Z regulate the output voltage. For applications requiring a fixed output voltage instead of current regulation, simply connect CS directly to OUT as shown in Figure 3. The MAX1577Z regulates to a fixed 5.1V. The adjustable regulation voltage (3.4V, 4.6V, or 5.1V) of the MAX1577Y depends on the state of the EN\_ inputs (see Table 1).

#### **Thermal Protection**

Thermal protection prevents damage to the IC during overload conditions. When the die temperature exceeds +160°C, the MAX1577Y/MAX1577Z turn off until the die cools by 20°C. Once the die has cooled, the MAX1577Y/MAX1577Z turn on and initiate soft-start. During a continuous overload, this produces a pulsed output. Continous operation at high-output-current settings may induce thermal protection.

#### **Capacitor Selection**

Ceramic capacitors are recommended due to their low equivalent series resistance (ESR), small size, and low cost. Select capacitors that maintain their capacitance over temperature and DC bias. Typically X5R, X7R, or better ceramic capacitors perform well.

A  $4.7\mu$ F input capacitor and  $10\mu$ F output capacitor are recommended for most applications. Larger values can be used to reduce ripple.

For the transfer capacitor (C1), a 1 $\mu$ F capacitor is suitable for output currents up to 600mA. For output currents between 600mA and 1200mA, use a 2.2 $\mu$ F capacitor.

#### Input Ripple

In 1x LDO mode there is no charge-pump switching, and thus very little input ripple. The input ripple in 2x charge-pump mode depends on the input source impedance. Typical waveforms showing the ripple in 1x and 2x mode are in the *Typical Operating Characteristics*. In sensitive applications, increase the input capacitance to reduce input ripple.

### \_Applications Information

#### **Setting the Output Current**

For applications requiring a regulated output current, see the circuit of Figure 2. The output regulation current is set by  $R_{CS}$  as follows:

#### $R_{CS} = (V_{OUT} - V_{CS}) / I_{OUT}$

where  $I_{OUT}$  is the output current (up to 1200mA) and (V<sub>OUT</sub> - V<sub>CS</sub>) is the current-sense threshold (see Table 1).

For camera-flash applications with a short pulse duration, a 0.25W resistor is recommended.

#### **Setting the Output Voltage**

To regulate the output voltage instead of the current, connect CS directly to the output as shown in Figure 3. In this configuration the MAX1577Z regulates the output to 5.1V. The output voltage of the MAX1577Y is set to 5.1V, 4.6V, or 3.4V using the EN1 and EN2 inputs as shown in Table 1.

#### **Driving Camera Strobe/Flash**

In applications that do not require dimming, such as a camera strobe/flash, connect EN1 and EN2 together and drive as a single on/off control input. Pulse this input high to generate a strobe output.

#### **PWM Dimming**

To vary the brightness from 20% to 100% using pulsewidth modulation (PWM), hold EN2 high and apply the PWM signal (up to 5kHz) to EN1. Holding EN2 high keeps the part enabled and prevents repeated softstarts. The duty cycle of the PWM signal applied to EN1 controls the brightness, where 0% duty cycle corresponds to 20% brightness and 100% duty cycle corresponds to 100% brightness. A PWM signal with a logic-high voltage between 1.6V and 5.5V can be safely used regardless of the input voltage. The voltage of the PWM signal does not affect the brightness level.

For highest movie-mode efficiency, it is best to choose RCS so 0% PWM duty cycle provides the desired movie-mode current. Then, the flash current level is adjusted by the PWM duty cycle. When the movie-mode current is more than 33% of the flash current, it is more efficient to apply the PWM dimming signal to EN2 while holding EN1 high.

#### **Ballast Resistors for Multiple LEDs**

When using multiple parallel LEDs, add a ballast resistor in series with each LED to help balance the current between the LEDs. When using current regulation



(Figure 2), the ballast resistors must be small enough to prevent triggering the overvoltage protection:

RBALLAST < (5.1V - VF) / ILED

where I<sub>LED</sub> is the current through each LED and V<sub>F</sub> is the forward voltage of the LED at I<sub>LED</sub>. Within this limit, larger values of R<sub>BALLAST</sub> improve LED-to-LED matching but also increase the input voltage where the 2x charge pump makes the transition to the more efficient 1x LDO mode.

When using voltage regulation (Figure 3), select the ballast resistors based on the full brightness setting as follows:

With EN1 high and EN2 high:

The LED current at the dimmed settings are calculated as follows:

with EN1 high and EN2 low:

 $I_{LED} = (4.6V - V_F) / R_{BALLAST}$ 

with EN1 low and EN2 high:

 $I_{LED} = (3.4V - V_F) / R_{BALLAST}$ 

Note that V<sub>F</sub> varies as a function of  $I_{LED}$ . Typically, LED manufacturers provide a plot of forward voltage (V<sub>F</sub>) vs. forward current ( $I_{LED}$ ) in the LED data sheet.

#### **PC** Board Layout and Routing

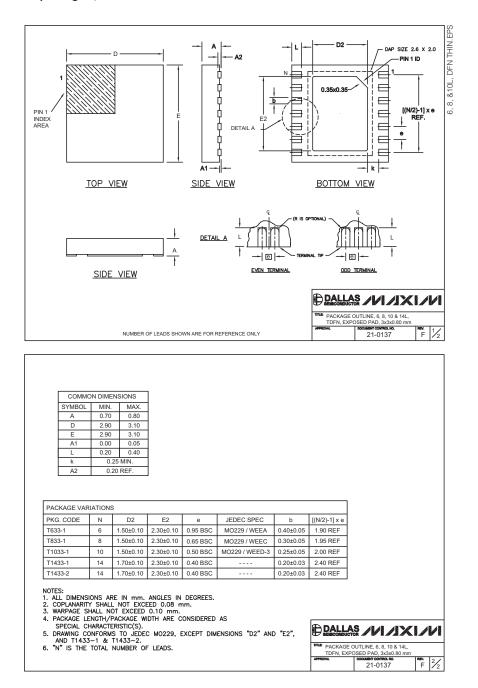
The MAX1577Y/MAX1577Z are high-frequency switched-capacitor regulators. For best circuit performance, use a solid ground plane and place C<sub>IN</sub>, C<sub>OUT</sub>, and C1 as close to the IC as possible. Connect the exposed pad to GND directly under the IC and allow sufficient copper area for cooling. Refer to the MAX1577Z evaluation kit for an example PC board layout.

### \_Chip Information

TRANSISTOR COUNT: 2482 PROCESS: BICMOS

### **Package Information**

(The package drawing(s) in this data sheet may not reflect the most current specifications. For the latest package outline information go to **www.maxim-ic.com/packages**.)



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